## **Description**

The 843N571I is a PLL based clock synthesizer for use in Ethernet applications. The device uses IDT's fourth generation FemtoClock<sup>®</sup> NG technology for optimal high clock frequency and low phase noise performance, combined with a low power consumption and high power supply noise rejection. Using IDT's latest FemtoClock NG PLL technology, the 843N571I achieves <0.3ps RMS phase jitter performance.

843N571I can synthesize 100MHz, 125MHz, 156.25MHz and a low frequency 33.33MHz CPU clock from a single device. Six LVCMOS outputs also serve as additional buffering of the 25MHz crystal reference.

### **Features**

- **•** Fourth generation FemtoClock® Next Generation (NG) technology
- Seven single-ended LVCMOS outputs, 30 $\Omega$  output impedance
- **•** Three LVPECL output pairs One differential LVPECL (QA, nQA) output pair: 156.25MHz Two selectable differential LVPECL output pairs (QB, nQB and QC, nQC): 100MHz and 125MHz
- **•** One single-ended LVCMOS (QD0) 33.33MHz CPU clock
- **•** Selectable external crystal or single-ended input source
- **•** Crystal oscillator interface designed for 25MHz, parallel resonant crystal
- **•** FemtoClock NG frequency multiplier provides low jitter, high frequency output
- **•** FemtoClock NG VCO frequency: 2.5GHz
- **•** RMS phase jitter @ 125MHz, using a 25MHz crystal (12kHz – 20MHz): 0.283ps (typical)
- **•** Power supply noise rejection PSNR: -80dB
- **•** 3.3V supply voltage
- **•** -40C to 85C ambient operating temperature
- **•** Lead-free (RoHS 6) packaging



## **Pin Assignment**

# **Block Diagram**



# **Pin Description and Pin Characteristic Tables**

### **Table 1. Pin Descriptions**



NOTE: *Pullup and Pulldown* refer to internal input resistors. See Table 2, *Pin Characteristics,* for typical values.

#### **Table 2. Pin Characteristics**



# **Function Tables**

**Table 3A. REFSEL Function Table**



### **Table 3B. FREQSEL Function Table**



### **Table 3C. FORCE\_LOW Function Table**



# **Absolute Maximum Ratings**

Exposure to absolute maximum rating conditions for extended periods may affect product reliability. Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics or AC Characteristics* is not implied.



## **Recommended Operating Conditions**



NOTE 1: It is the user's responsibility to ensure that device junction temperature remains below the maximum allowed.

NOTE 2: All conditions in the table must be met to guarantee device functionality.

NOTE 3: The device is verified to the maximum operating junction temperature through simulation.

## **DC Electrical Characteristics**

**Table 4A. Power Supply DC Characteristics,**  $V_{CC}$  = 3.3V  $\pm$  0.3V,  $V_{EE}$  = 0V,  $T_A$  = -40°C to 85°C



### **Table 4B. LVCMOS/LVTTL DC Characteristics,**  $V_{CC} = 3.3V \pm 0.3V$ ,  $T_A = -40^{\circ}C$  to 85°C





NOTE 1: Outputs terminated with 50Ω to V<sub>CC</sub>/2. See Parameter Measurement Information, *Output Load Test Circuit diagrams.* 

### **Table 4C. LVPECL DC Characteristics,**  $V_{CC} = 3.3V \pm 0.3V$ ,  $V_{EE} = 0V$ ,  $T_A = -40^{\circ}$ C to 85°C



NOTE 1: Outputs terminated with 50 $\Omega$  to V<sub>CC</sub> – 2V.

#### **Table 5. Crystal Characteristics**



### **AC Electrical Characteristics**

### **Table 6A. LVPECL AC Characteristics,**  $V_{CC} = 3.3V \pm 0.3V$ ,  $V_{EE} = 0V$ ,  $T_A = -40^{\circ}C$  to 85°C





NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions

NOTE 1: Refer to the Phase Noise Plot.

NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 3: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at the differential crosspoints.

### **Table 6B. AC Characteristics for Single Side Band Power Levels (LVPECL Outputs),**

 $V_{CC}$  = 3.3V ± 0.3V,  $V_{EE}$  = 0V, T<sub>A</sub> = 25<sup>°</sup>C





NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

### **Table 6C. LVCMOS AC Characteristics,**  $V_{CC} = 3.3V \pm 0.3V$ ,  $T_A = -40^{\circ}C$  to 85°C



NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions NOTE 1: Refer to the Phase Noise Plot.

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NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 3: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at V<sub>CC</sub>/2.

### **Table 6D. AC Characteristics for Single Side Band Power Levels (LVCMOS Outputs),**

 $V_{CC}$  = 3.3V ± 0.3V,  $V_{EE}$  = 0V, T<sub>A</sub> = 25°C



NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.



# **Typical Phase Noise at 25MHz (LVCMOS Output)**





# **Typical Phase Noise at 100MHz (LVPECL Output)**





## **Typical Phase Noise at 156.25MHz (LVPECL Output)**

## **Parameter Measurement Information**



**LVPECL Output Load Test Circuit**



**Phase Jitter**



**LVPECL Output Skew**



**LVCMOS Output Load Test Circuit**



**LVCMOS Output Skew**



**LVCMOS Output Rise/Fall Time**

# **Parameter Measurement Information, continued**





**LVCMOS Output Duty Cycle/Pulse Width/Period**





**LVPECL Output Duty Cycle/Pulse Width/Period**

# **Applications Information**

## **Overdriving the XTAL Interface**

The XTAL\_IN input can accept a single-ended LVCMOS signal through an AC coupling capacitor. A general interface diagram is shown in *Figure 1A.* The XTAL\_OUT pin can be left floating. The maximum amplitude of the input signal should not exceed 2V and the input edge rate can be as slow as 10ns. This configuration requires that the output impedance of the driver (Ro) plus the series resistance (Rs) equals the transmission line impedance. In addition,

3.3V 3.3V R1 100  $Zo = 50$  Ohm C1 7 Ohm  $\overline{\mathcal{W}}$ XTAL\_IN RS 43 0.1uF R2 100 Driv er\_LVCMOS  $\Box$ **XTAL\_OU** Cry stal Input Interf ace

**Figure 1A. General Diagram for LVCMOS Driver to XTAL Input Interface**



**Figure 1B. General Diagram for LVPECL Driver to XTAL Input Interface**

should equal the transmission line impedance. For most  $50\Omega$ applications, R1 and R2 can be 100 $\Omega$ . This can also be accomplished by removing R1 and making R2 50 $\Omega$ . By overdriving the crystal oscillator, the device will be functional, but note, the device performance is guaranteed by using a quartz crystal.

matched termination at the crystal input will attenuate the signal in half. This can be done in one of two ways. First, R1 and R2 in parallel

### **VFQFN EPAD Thermal Release Path**

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 2.* The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, please refer to the Application Note on the Surface Mount Assembly of Amkor's Thermally/ Electrically Enhance Leadframe Base Package, Amkor Technology.



**Figure 2. P.C. Assembly for Exposed Pad Thermal Release Path – Side View (drawing not to scale)**

### **Termination for 3.3V LVPECL Outputs**

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive  $50\Omega$ 



transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 3A and 3B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.



**Figure 3A. 3.3V LVPECL Output Termination Figure 3B. 3.3V LVPECL Output Termination**

### **Recommendations for Unused Input and Output Pins**

### **Inputs:**

### **REFCLK Input**

For applications not requiring the use of the reference clock, it can be left floating. Though not required, but for additional protection, a  $1k\Omega$  resistor can be tied from the REFCLK to ground.

### **Crystal Inputs**

For applications not requiring the use of the crystal oscillator input, both XTAL\_IN and XTAL\_OUT can be left floating. Though not required, but for additional protection, a  $1k\Omega$  resistor can be tied from XTAL\_IN to ground.

### **LVCMOS Control Pins**

All control pins have internal pull-ups or pull-downs; additional resistance is not required but can be added for additional protection. A 1 $k\Omega$  resistor can be used.

### **Outputs:**

### **LVCMOS Outputs**

All unused LVCMOS output can be left floating. There should be no trace attached.

### **LVPECL Outputs:**

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

### **Application Schematic Example**

*Figure 4* shows an example of 843N571I application schematic. In this example, the device is operated at  $V_{CC}$  = 3.3V. An 18pF parallel resonant 25MHz crystal is used. The load capacitance C1 = 15pF and C2 = 15pF are recommended for frequency accuracy. Depending on the parasitics of the printed circuit board layout, these values might required slight adjustment to optimize the frequency accuracy. Crystals with other load capacitance specifications can be used. This will require adjusting C1 and C2. For this device, the crystal load capacitors are required for proper operation.

As with any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. In order to achieve the best possible filtering, it is recommended that the placement of the filter components be on the device side of the PCB as close to the power pins as possible. If space is limited, the 0.1uF capacitor in each power pin filter should be placed on the device side. The other

components can be on the opposite side of the PCB.

Power supply filter recommendations are a general guideline to be used for reducing external noise from coupling into the devices. The filter performance is designed for wide range of noise frequency. This low-pass filter starts to attenuate noise at approximately 10kHz. If a specific frequency noise component with high amplitude interference is known, such as switching power supplies frequencies, it is recommended that component values be adjusted and if required, additional filtering be added. Additionally general design practice for power plane voltage stability suggests adding bulk capacitances in the general area of all devices.

The schematic example focuses on functional connections and is not configuration specific. Refer to the pin description and functional tables in the datasheet to ensure the logic control inputs are properly set.



**Figure 4. 843N571I Application Schematic**

## **Power Considerations**

This section provides information on power dissipation and junction temperature for the 843N571I. Equations and example calculations are also provided.

#### **1. Power Dissipation.**

The total power dissipation for the 843N571I is the sum of the core power plus the power dissipated due to loading. The following is the power dissipation for  $V_{CC}$  = 3.6V, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated due to loading.

#### **Core and LVPECL Output Power Dissipation**

- Power (core)<sub>MAX</sub> = V<sub>CC\_MAX</sub> \*  $I_{EE}$ <sub>MAX</sub> = 3.6V \* 250mA = **900mW**
- Power (outputs) $_{MAX}$  = 32mW/Loaded Output pair If all outputs are loaded, the total power is 3 \* 32mW = **96mW**

#### **Dynamic Power Dissipation at 33.3333MHz and 25MHz**

Power (33.33MHz) = C<sub>PD</sub> \* Frequency \* (V<sub>CC</sub>)<sup>2</sup> \* # of outputs = 6pF \* 33.3333MHz \* (3.6V)<sup>2</sup> \* 1= **2.592mW** Power (25MHz) =  $C_{PD}$  \* Frequency \* (V<sub>CC</sub>)<sup>2 \*</sup> # of outputs = 6pF \* 25MHz \* (3.6V)<sup>2</sup> \* 6 = **11.664mW** 

#### **Total Power Dissipation**

- **Total Power**
	- = Power (Core) + Power (Output) + Dynamic Power (33.3333MHz) + Dynamic Power (25MHz)
	- = 900mW + 96mW + 2.592mW + 11.66mW
	- **= 1010.252mW**

#### **2. Junction Temperature.**

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad, and directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj =  $\theta_{JA}$  \* Pd\_total + T<sub>A</sub>

Tj = Junction Temperature

 $\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

Pd\_total = Total Device Power Dissipation (example calculation is in section 1 above)

 $T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is 37.7°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

85°C + 1.010W \* 37.7°C/W = 123.1°C. This is below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

#### Table 7. Thermal Resistance  $\theta_{JA}$  for 40 Lead VFQFN Forced Convection



### **3. Calculations and Equations.**

The purpose of this section is to calculate the power dissipation for the LVPECL output pairs.

LVPECL output driver circuit and termination are shown in *Figure 5.*



**Figure 5. LVPECL Driver Circuit and Termination**

To calculate power dissipation due to loading, use the following equations which assume a 50 $\Omega$  load, and a termination voltage of V<sub>CC</sub> – 2V.

- $\bullet$   $\,$  For logic high, V<sub>OUT</sub> = V<sub>OH\_MAX</sub> = **V**<sub>CC\_MAX</sub> **–0.8V**  $(V_{\text{CC\_MAX}} - V_{\text{OH\_MAX}}) = 0.8V$
- For logic low,  $V_{\text{OUT}} = V_{\text{OL\_MAX}} = V_{\text{CC\_MAX}} 1.6V$  $(V_{CC~MAX} - V_{OL~MAX}) = 1.6V$

•

Pd H is power dissipation when the output drives high.

Pd\_L is the power dissipation when the output drives low.

 $Pd_H = [(V_{OH~MAX} - (V_{CC~MAX} - 2V))/R_L]$  \* ( $V_{CC~MAX} - V_{OH~MAX}$ ) = [(2V – ( $V_{CC~MAX} - V_{OH~MAX}$ ))/R<sub>L</sub>] \* ( $V_{CC~MAX} - V_{OH~MAX}$ ) =  $[(2V - 0.8V)/50\Omega]$  \* 0.8V = **19.2mW** 

 $Pd\_L = [(V_{OL\_MAX} - (V_{CC\_MAX} - 2V))/R_L] * (V_{CC\_MAX} - V_{OL\_MAX}) = [(2V - (V_{CC\_MAX} - V_{OL\_MAX}))/R_L] * (V_{CC\_MAX} - V_{OL\_MAX}) = [(2V - (V_{CC\_MAX} - V_{OL\_MAX})/R_L] * (V_{CC\_MAX} - V_{OL\_MAX})]$  $[(2V - 1.6V)/50 $\Omega$ ] * 1.6V = 12.8mW$ 

Total Power Dissipation per output pair = Pd\_H + Pd\_L = **32mW**

## **Reliability Information**

#### Table 8.  $\theta_{JA}$  vs. Air Flow Table for a 40 Lead VFQFN



### **Transistor Count**

The transistor count for 843N571I is: 22,466

## **Package Outline Drawings**

The package outline drawings are appended at the end of this document and are accessible from the link below. The package information is the most current data available.

www.idt.com/document/psc/nlnlg-40-package-outline-60-x-60-mm-body-epad-290-x-290-mm-qfn

### **Ordering Information**

#### **Table 10. Ordering Information**



## **Revision History**











1. DIMENSIONING & TOLERANCING CONFORM TO ASME Y14.5M. - 1994.

- 2. N IS THE NUMBER OF TERMINALS. Nd IS THE NUMBER OF TERMINALS IN X-DIRECTION & Ne IS THE NUMBER OF TERMINALS IN Y-DIRECTION.
- 3. ALL DIMENSIONS ARE IN MILLIMETERS.
- $\sqrt{4}$  dimension b applies to plated terminal and is measured BETWEEN 0.20 AND 0.30mm FROM TERMINAL TIP.

 $/$ 5. The Pin #1 identifier must exist on the top surface of the package BY USING INDENTATION MARK OR OTHER FEATURE OF PACKAGE BODY.  $/6$ . Exact shape and size of this feature is optional.

 $\sqrt{2}$  applied to exposed pad and terminals. Exclude embedded PART OF EXPOSED PAD FROM MEASURING.

 $\sqrt{8}$  applied only for terminals.

9. THIS OUTLINES CONFORMS TO JEDEC PUBLICATION 95 REGISTRATION MO-220, VARIATION VJJC-3 & VJJD-5 WITH THE EXCEPTION OF D2 & E2.











RECOMMENDED LAND PATTERN

### Inotes:

- 1. ALL DIMENSIONS ARE IN mm. ANGLES IN DEGREES.
- 2. TOP DOWN VIEW. AS VIEWED ON PCB.
- 3. COMPONENT OUTLINE SHOWS FOR REFERENCE IN GREEN.
- 4. LAND PATTERN IN BLUE. NSMD PATTERN ASSUMED.
- 5. LAND PATTERN RECOMMENDATION PER IPC-7351B GENERIC REQUIREMENT FOR SURFACE MOUNT DESIGN AND LAND PATTERN.

