

Description

The 9ZML1245/9ZML1255/9ZML1256 devices are third generation enhanced performance 2-input, 12-output clock multiplexers. Each input clock has software adjustable input-to-output delay when operating in Zero-Delay (ZDB) mode. The devices also implement an extensive set of features ensuring that clocks are well behaved with today's ever more complex power-up sequencing. The 9ZML1256 has an SMBus Write Lockout pin for increased device and system security.

PCIe Clocking Architectures

- Common Clocked (CC)
- Independent Reference (IR) with and without spread spectrum

Key Specifications

- Fanout Buffer Mode additive phase jitter:
 - PCIe Gen5 CC < 15fs RMS
 - DB2000Q additive jitter < 25fs RMS
 - QPI/UPI 11.4GB/s < 40fs RMS
 - IF-UPI additive jitter < 70fs RMS
- ZDB Mode phase jitter:
 - PCIe Gen5 CC < 24fs RMS
 - QPI/UPI 11.4GB/s < 110fs RMS
 - IF-UPI additive jitter < 130fs RMS
- Cycle-to-cycle jitter < 50ps
- Output-to-output skew < 50ps

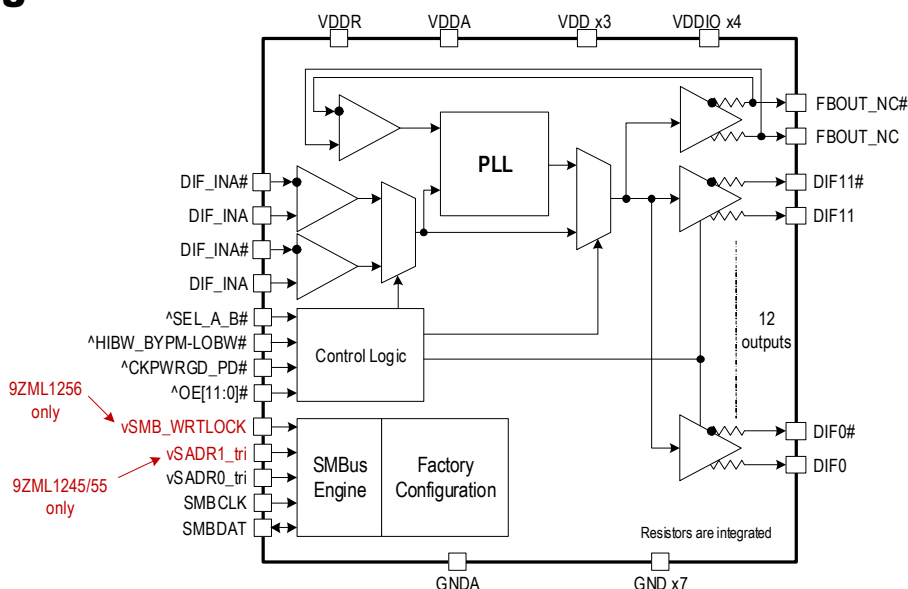
Features

- 12 Low-power HCSL (LP-HCSL) outputs
- Integrated terminations for 85Ω (9ZXL1255/56) and 100Ω (9ZXL1245) systems eliminate up to 4 resistors per output pair
- Flexible Power Sequencing (FPS) configuration via SMBus
- Power Down Tolerant input (PDT); inputs: SEL_A_B#, SADR[1:0]_tri, OE# pins
- Input-to-output delay: 0ps default. Programmable through 360°
- 2 software-configurable input-to-output delay lines
- Dedicated OE# pins support PCIe CLKREQ# function
- Up to 9 selectable SMBus addresses (9ZXL1245/55)
- SMBus Write Protect Pin (9ZXL1256)
- Selectable ZDB bandwidths minimizes jitter peaking in cascaded ZDB topologies
- Hardware/SMBus control of ZDB and FOB modes
- Spread-spectrum compatible
- Up to 400MHz FOB operation
- 100MHz ZDB mode operation
- 40°C to +85°C operating temperature range
- 10 × 10 mm 72-VFQFPN package

Typical Applications

- Servers/High-performance Computing
- nVME Storage
- Networking
- Accelerators
- Industrial Control

Block Diagram

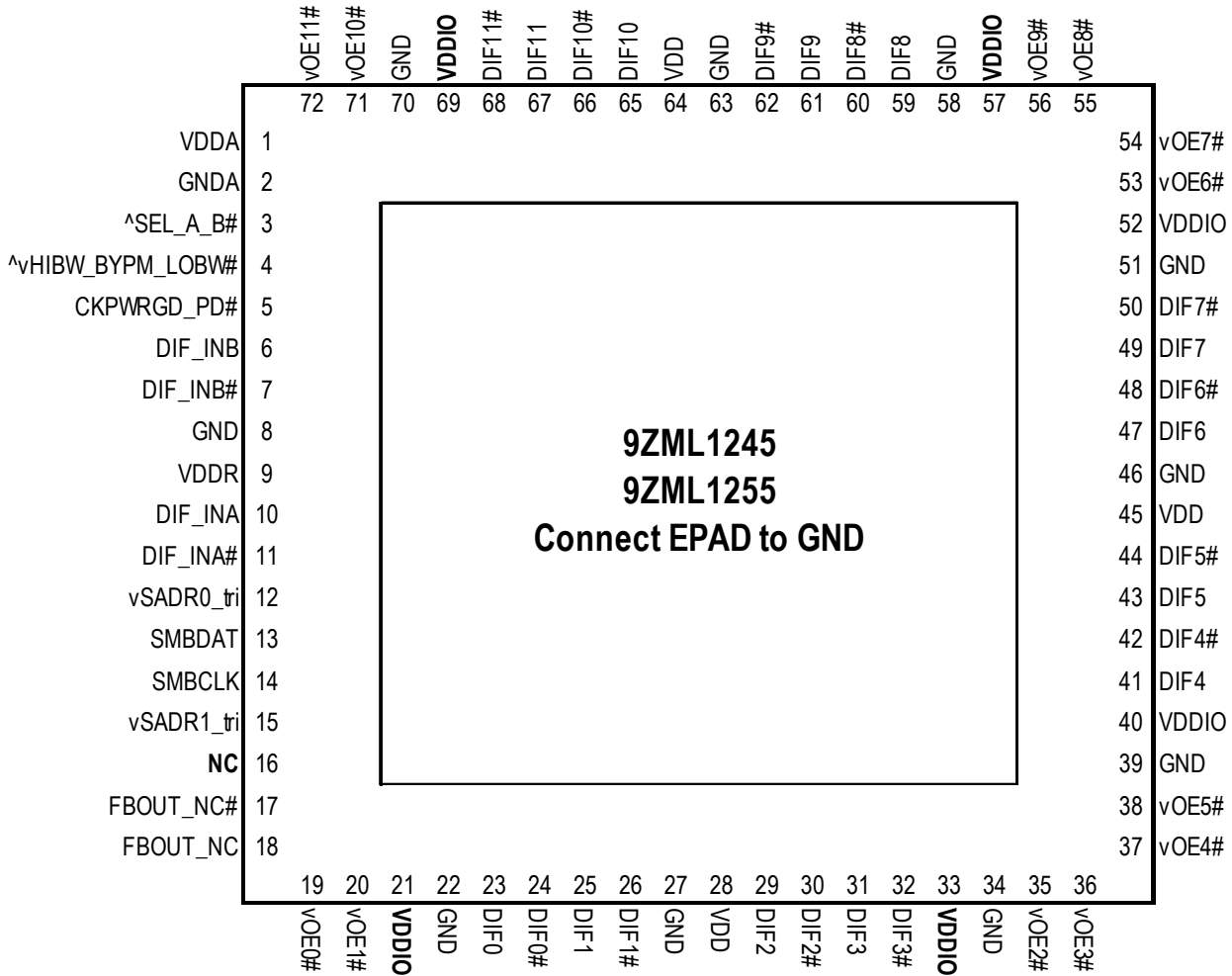


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Pin Assignments

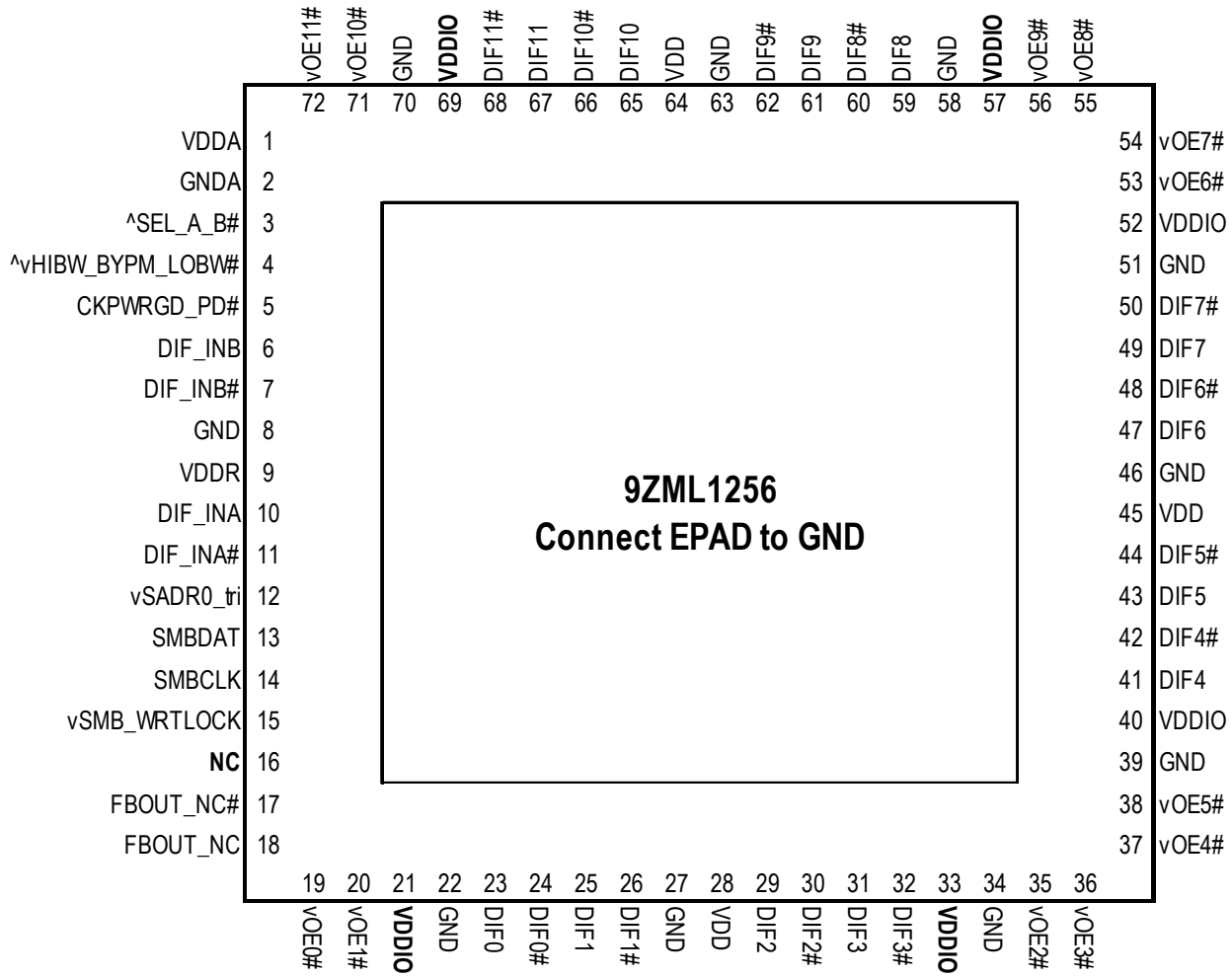
9ZML1245/9ZML1255



10 × 10 mm, 0.5mm pitch 72-VFQFPN

- ^ prefix indicates internal 120kOhm pull-up
- v prefix indicates internal 120kOhm pull-down
- ^v prefix indicates internal 120kOhm pull-down

9ZML1256



10 × 10 mm, 0.5mm pitch 72-VFQFPN

- ^ prefix indicates internal 120kOhm pull-up
- v prefix indicates internal 120kOhm pull-down
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Pin Descriptions

Table 1. Pin Descriptions

| Name | Type | Description | 9ZML1245/ 9ZML1255 Pin No. | 9ZML1256 Pin No. |
|-------------------|------------|--|----------------------------------|---------------------|
| ^SEL_A_B# | Input | Input to select differential input clock A or differential input clock B. This input has an internal pull-up resistor. 0 = Input B selected, 1 = Input A selected. | 3 | 3 |
| ^vHIBW_BYPM_LOBW# | Latched In | Tri-level input to select High BW, Bypass or Low BW mode. This pin is biased to VDD/2 (Bypass mode) with internal pull up/pull down resistors. See PLL Operating Mode table for Details. | 4 | 4 |
| CKPWRGD_PD# | Input | 3.3V input notifies device to sample latched inputs and start up on first high assertion, or exit Power Down Mode on subsequent assertions. Low enters Power Down Mode. | 5 | 5 |
| DIF_INA | Input | True input of differential clock. | 10 | 10 |
| DIF_INA# | Input | Complement input of differential clock. | 11 | 11 |
| DIF_INB | Input | True input of differential clock. | 6 | 6 |
| DIF_INB# | Input | Complement input of differential clock. | 7 | 7 |
| DIF0 | Output | Differential true clock output. | 23 | 23 |
| DIF0# | Output | Differential complementary clock output. | 24 | 24 |
| DIF1 | Output | Differential true clock output. | 25 | 25 |
| DIF1# | Output | Differential complementary clock output. | 26 | 26 |
| DIF10 | Output | Differential true clock output. | 65 | 65 |
| DIF10# | Output | Differential complementary clock output. | 66 | 66 |
| DIF11 | Output | Differential true clock output. | 67 | 67 |
| DIF11# | Output | Differential complementary clock output. | 68 | 68 |
| DIF2 | Output | Differential true clock output. | 29 | 29 |
| DIF2# | Output | Differential complementary clock output. | 30 | 30 |
| DIF3 | Output | Differential true clock output. | 31 | 31 |
| DIF3# | Output | Differential complementary clock output. | 32 | 32 |
| DIF4 | Output | Differential true clock output. | 41 | 41 |
| DIF4# | Output | Differential complementary clock output. | 42 | 42 |
| DIF5 | Output | Differential true clock output. | 43 | 43 |
| DIF5# | Output | Differential complementary clock output. | 44 | 44 |
| DIF6 | Output | Differential true clock output. | 47 | 47 |
| DIF6# | Output | Differential complementary clock output. | 48 | 48 |
| DIF7 | Output | Differential true clock output. | 49 | 49 |
| DIF7# | Output | Differential complementary clock output. | 50 | 50 |
| DIF8 | Output | Differential true clock output. | 59 | 59 |
| DIF8# | Output | Differential complementary clock output. | 60 | 60 |

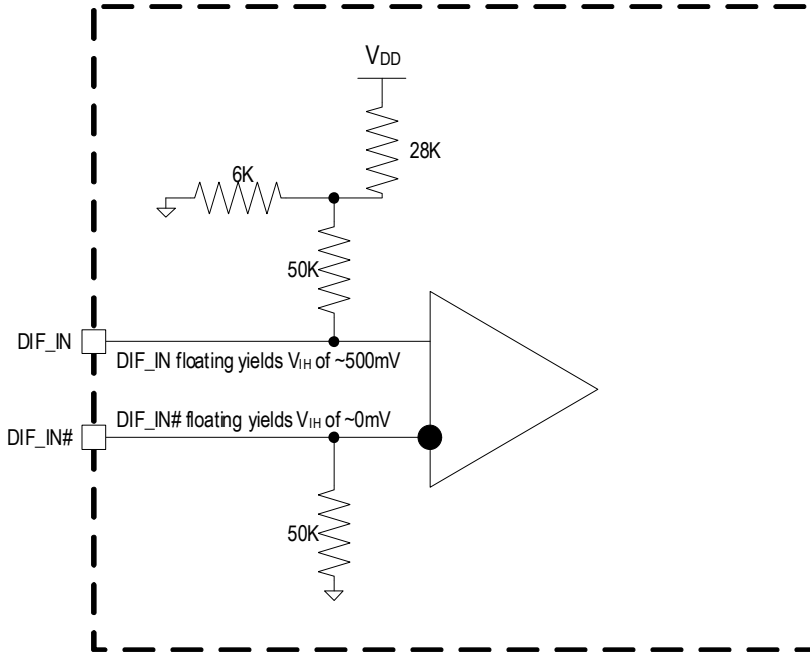
Table 1. Pin Descriptions (Cont.)

| Name | Type | Description | 9ZML1245/ 9ZML1255 Pin No. | 9ZML1256 Pin No. |
|-----------|--------|---|----------------------------------|---------------------|
| DIF9 | Output | Differential true clock output. | 61 | 61 |
| DIF9# | Output | Differential complementary clock output. | 62 | 62 |
| EPAD | GND | Connect to ground. | 73 | 73 |
| FBOUT_NC | Output | True half of differential feedback output. This pin should NOT be connected to anything outside the chip. It exists to provide delay path matching to get 0 propagation delay. | 18 | 18 |
| FBOUT_NC# | Output | Complementary half of differential feedback output. This pin should NOT be connected to anything outside the chip. It exists to provide delay path matching to get 0 propagation delay. | 17 | 17 |
| GND | GND | Ground pin. | 8 | 8 |
| GND | GND | Ground pin. | 22 | 22 |
| GND | GND | Ground pin. | 27 | 27 |
| GND | GND | Ground pin. | 34 | 34 |
| GND | GND | Ground pin. | 39 | 39 |
| GND | GND | Ground pin. | 46 | 46 |
| GND | GND | Ground pin. | 51 | 51 |
| GND | GND | Ground pin. | 58 | 58 |
| GND | GND | Ground pin. | 63 | 63 |
| GND | GND | Ground pin. | 70 | 70 |
| GND A | GND | Ground pin for the PLL core. | 2 | 2 |
| NC | — | No connection. | 16 | 16 |
| SMBCLK | Input | Clock pin of SMBUS circuitry. | 14 | 14 |
| SMBDAT | I/O | Data pin of SMBUS circuitry. | 13 | 13 |
| VDD | Power | Power supply, nominally 3.3V. | 28 | 28 |
| VDD | Power | Power supply, nominally 3.3V. | 45 | 45 |
| VDD | Power | Power supply, nominally 3.3V. | 64 | 64 |
| VDDA | Power | Power supply for PLL core. See Power Connections table for additional information. | 1 | 1 |
| VDDIO | Power | Power supply for differential outputs. | 21 | 21 |
| VDDIO | Power | Power supply for differential outputs. | 33 | 33 |
| VDDIO | Power | Power supply for differential outputs. | 40 | 40 |
| VDDIO | Power | Power supply for differential outputs. | 52 | 52 |
| VDDIO | Power | Power supply for differential outputs. | 57 | 57 |
| VDDIO | Power | Power supply for differential outputs. | 69 | 69 |
| VDDR | Power | Power supply for differential input clock (receiver). This VDD should be treated as an analog power rail and filtered appropriately. Nominally 3.3V. | 9 | 9 |

Table 1. Pin Descriptions (Cont.)

| Name | Type | Description | 9ZML1245/ 9ZML1255 Pin No. | 9ZML1256 Pin No. |
|--------------|-------|---|----------------------------------|---------------------|
| vOE0# | Input | Active low input for enabling output 0. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 19 | 19 |
| vOE1# | Input | Active low input for enabling output 1. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 20 | 20 |
| vOE10# | Input | Active low input for enabling output 10. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 71 | 71 |
| vOE11# | Input | Active low input for enabling output 11. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 72 | 72 |
| vOE2# | Input | Active low input for enabling output 2. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 35 | 35 |
| vOE3# | Input | Active low input for enabling output 3. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 36 | 36 |
| vOE4# | Input | Active low input for enabling output 4. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 37 | 37 |
| vOE5# | Input | Active low input for enabling output 5. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 38 | 38 |
| vOE6# | Input | Active low input for enabling output 6. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 53 | 53 |
| vOE7# | Input | Active low input for enabling output 7. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 54 | 54 |
| vOE8# | Input | Active low input for enabling output 8. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 55 | 55 |
| vOE9# | Input | Active low input for enabling output 9. This pin has an internal pull-down. 1 = disable output, 0 = enable output. | 56 | 56 |
| vSADR0_tri | Input | SMBus address bit. This is a tri-level input that works in conjunction with other SADR pins, if present, to decode SMBus Addresses. It has an internal pull-down resistor. See the SMBus Addresses table. | 12 | 12 |
| vSADR1_tri | Input | SMBus address bit. This is a tri-level input that works in conjunction with other SADR pins, if present, to decode SMBus Addresses. It has an internal pull-down resistor. See the SMBus Addresses table. | 15 | — |
| vSMB_WRTLOCK | Input | This pin prevents SMBus writes when asserted. SMBus reads are not affected. This pin has an internal pull-down. 0 = SMBus writes allows, 1 = SMBus writes blocked. | — | 15 |

Figure 1. Clock Input Bias Network



Absolute Maximum Ratings

Stresses above the ratings listed below can cause permanent damage to the 9ZML1245/9ZML1255/9ZML1256. These ratings, which are standard values for Renesas commercially rated parts, are stress ratings only. Functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods can affect product reliability. Electrical parameters are guaranteed only over the recommended operating temperature range.

Table 2. Absolute Maximum Ratings

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Units | Notes |
|----------------------|-------------|-----------------------------|-----------|---------|----------------|----------|-------|
| Supply Voltage | V_{DDX} | | | | 3.9 | V_{DD} | 1,2 |
| Input Low Voltage | V_{IL} | | GND - 0.5 | | | V | 1 |
| Input High Voltage | V_{IH} | Except for SMBus interface. | | | $V_{DD} + 0.5$ | V | 1,3 |
| Input High Voltage | V_{IHSMB} | SMBus clock and data pins. | | | 3.9 | V | 1 |
| Storage Temperature | T_s | | -65 | | 150 | °C | 1 |
| Junction Temperature | T_j | | | | 125 | °C | 1 |
| Input ESD Protection | ESD prot | Human Body Model. | 2500 | | | V | 1 |

¹ Guaranteed by design and characterization, not 100% tested in production.

² Operation under these conditions is neither implied nor guaranteed.

³ Not to exceed 3.9V.

Thermal Characteristics

Table 3. Thermal Characteristics

| Parameter | Symbol | Conditions | Package | Typical Values | Units | Notes |
|-----------------------------------|----------------|----------------------------------|---------|----------------|-------|-------|
| 9ZML12xx Thermal Resistance | θ_{JC} | Junction to case. | NLG72 | 19 | °C/W | 1 |
| | θ_{Jb} | Junction to base. | | 0.5 | °C/W | 1 |
| | θ_{JA0} | Junction to air, still air. | | 30 | °C/W | 1 |
| | θ_{JA1} | Junction to air, 1 m/s air flow. | | 23 | °C/W | 1 |
| | θ_{JA3} | Junction to air, 3 m/s air flow. | | 20 | °C/W | 1 |
| | θ_{JA5} | Junction to air, 5 m/s air flow. | | 19 | °C/W | 1 |

Electrical Characteristics

T_{AMB} = over the specified operating range. Supply voltages per normal operation conditions; see [Test Loads](#) for loading conditions.

Table 4. SMBus Parameters

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Units | Notes |
|---------------------------|--------------|---|---------|---------|-------------|-------|-------|
| SMBus Input Low Voltage | V_{ILSMB} | | | | 0.8 | V | |
| SMBus Input High Voltage | V_{IHSMB} | | 2.1 | | V_{DDSMB} | V | |
| SMBus Output Low Voltage | V_{OLSMB} | At I_{PULLUP} . | | | 0.4 | V | |
| SMBus Sink Current | I_{PULLUP} | At V_{OL} . | 4 | | | mA | |
| Nominal Bus Voltage | V_{DDSMB} | | 2.7 | | 3.6 | V | 1 |
| SCLK/SDATA Rise Time | t_{RSMB} | (Max $V_{IL} - 0.15V$) to (Min $V_{IH} + 0.15V$). | | | 1000 | ns | 1 |
| SCLK/SDATA Fall Time | t_{FSMB} | (Min $V_{IH} + 0.15V$) to (Max $V_{IL} - 0.15V$). | | | 300 | ns | 1 |
| SMBus Operating Frequency | f_{SMB} | SMBus operating frequency. | | | 400 | kHz | 5 |

¹ Guaranteed by design and characterization, not 100% tested in production.

² Control input must be monotonic from 20% to 80% of input swing.

³ Time from deassertion until outputs are > 200mV.

⁴ DIF_IN input.

⁵ The differential input clock must be running for the SMBus to be active.

Table 5. DIF_IN Clock Input Parameters

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Units | Notes |
|----------------------------------|--------------------|--|---------|---------|---------|-------|-------|
| Input Crossover Voltage – DIF_IN | V _{CROSS} | Crossover voltage. | 150 | | 900 | mV | 1 |
| Input Swing – DIF_IN | V _{SWING} | Differential value. | 300 | | | mV | 1 |
| Input Slew Rate – DIF_IN | dv/dt | Measured differentially. | 0.35 | | 8 | V/ns | 1,2 |
| Input Leakage Current | I _{IN} | V _{IN} = V _{DD} , V _{IN} = GND. | -5 | | 5 | μA | |
| Input Duty Cycle | d _{tin} | Measurement from differential waveform. | 45 | | 55 | % | 1 |
| Input Jitter – Cycle to Cycle | J _{DIFIn} | Differential measurement. | 0 | | 125 | ps | 1 |

¹ Guaranteed by design and characterization, not 100% tested in production.

² Slew rate measured through ±75mV window centered around differential zero.

Table 6. Input/Supply/Common Parameters

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Units | Notes |
|------------------------------------|------------------------|---|-----------|--------------------|-----------------------|-------|-------|
| Supply Voltage | V _{DDX} | Supply voltage for core and analog. | 3.135 | 3.3 | 3.465 | V | |
| Output Supply Voltage | V _{DDIO} | Supply voltage for DIF outputs, if present. | 0.95 | 1.05 | 3.465 | V | 5 |
| Ambient Operating Temperature | T _{AMB} | Industrial range (T _{IND}). | -40 | 25 | 85 | °C | |
| Input High Voltage | V _{IH} | Single-ended inputs, except SMBus, tri-level inputs. | 2 | | V _{DD} + 0.3 | V | |
| Input Low Voltage | V _{IL} | Single-ended inputs, except SMBus, tri-level inputs. | GND - 0.3 | | 0.8 | V | |
| Input High Voltage | V _{IH} | Tri-level inputs. | 2.2 | | V _{DD} + 0.3 | V | |
| Input Mid Voltage | V _{IM} | Tri-level inputs. | 1.2 | V _{DD} /2 | 1.8 | V | |
| Input Low Voltage | V _{IL} | Tri-level inputs. | GND - 0.3 | | 0.8 | V | |
| Input Current | I _{IN} | Single-ended inputs, V _{IN} = GND, V _{IN} = V _{DD} . | -5 | | 5 | μA | |
| | I _{INP} | Single-ended inputs. V _{IN} = 0 V; inputs with internal pull-up resistors. V _{IN} = V _{DD} ; inputs with internal pull-down resistors. | -50 | | 50 | μA | |
| Input Frequency | F _{ibyp} | Bypass/FOB Mode, Ck Detect Disabled (B4b6=0). | 1 | | 400 | MHz | 5 |
| | F _{ibypCkDet} | Bypass/FOB Mode, Ck Detect Enabled (B4b6=1). | 15 | | 400 | MHz | |
| | F _{izDB} | 100MHz ZDB Mode. | 98.5 | 100.00 | 102.5 | MHz | |
| ppm Error Contribution | ppm | ppm error contributed to input clock. | 0 | | | ppm | |
| Pin Inductance | L _{pin} | | | | 7 | nH | 1 |
| Capacitance | C _{IN} | Logic inputs, except DIF_IN. | 1.5 | | 5 | pF | 1 |
| | C _{INDIF_IN} | DIF_IN differential clock inputs. | 1.5 | | 2.7 | pF | 1,4 |
| | C _{OUT} | Output pin capacitance. | | | 6 | pF | 1 |
| Clk Stabilization | T _{STAB} | From V _{DD} power-up and after input clock stabilization or deassertion of PD# to 1st clock. | | 1 | 1.8 | ms | 1, 2 |
| Input SS Modulation Frequency PCIe | f _{MODINPCIe} | Allowable frequency for PCIe applications (Triangular modulation). | 30 | | 33 | kHz | |

Table 6. Input/Supply/Common Parameters (Cont.)

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Units | Notes |
|-------------|---------------|---|---------|---------|---------|---------|-------|
| OE# Latency | $t_{LATOE\#}$ | DIF start after OE# assertion. DIF stop after OE# deassertion. | 4 | 5 | 10 | clocks | 1,2,3 |
| Tdrive_PD# | t_{DRVPD} | DIF output enable after PD# deassertion. | | 57 | 300 | μ s | 1,3 |
| Tfall | t_F | Fall time of control inputs. | | | 5 | ns | 2 |
| Trise | t_R | Rise time of control inputs. | | | 5 | ns | 2 |

¹ Guaranteed by design and characterization, not 100% tested in production.

² Control input must be monotonic from 20% to 80% of input swing.

³ Time from deassertion until outputs are > 200mV, PLL mode.

⁴ DIF_IN input.

⁵ This is the default value.

Table 7. Current Consumption

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Units | Notes |
|--------------------------|-------------|--|---------|---------|---------|-------|-------|
| Operating Supply Current | I_{DDA} | V_{DDA} , PLL Mode at 100MHz. | | 55 | 66 | mA | 1 |
| | | V_{DDA} , Fan-out Buffer Mode at 100MHz. | | 10 | 13 | mA | 1 |
| | I_{DD} | All other V_{DD} pins. | | 16 | 20 | mA | |
| | I_{DDIO} | V_{DDIO} for LP-HCSL outputs, if applicable. | | 78 | 97 | mA | |
| Power Down Current | I_{DDAPD} | V_{DDA} , CKPWRGD_PD# = 0. | | 4 | 5 | mA | 1 |
| | I_{DDPD} | All other V_{DD} pins, CKPWRGD_PD# = 0. | | 0.5 | 1 | mA | |
| | I_{DDO} | V_{DDIO} pins, CKPWRGD_PD# = 0. | | 0.04 | 0.1 | mA | |

¹ Includes V_{DDR} .

Table 8. Skew and Differential Jitter Parameters

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Units | Notes |
|------------------------|------------------------|---|---------|---------|---------|----------|-------------|
| CLK_IN, DIF[x:0] | $t_{\text{SKEW_PLL}}$ | Input-to-output skew in PLL Mode at 100MHz, nominal temperature and voltage. | -100 | 9 | 100 | ps | 1,2,4,5,6,8 |
| CLK_IN, DIF[x:0] | $t_{\text{PD_BYP}}$ | Input-to-output skew in Bypass Mode at 100MHz, nominal temperature and voltage. | 2.3 | 3 | 3.8 | ns | 1,2,3,8 |
| CLK_IN, DIF[x:0] | $t_{\text{DSPO_PLL}}$ | Input-to-output skew variation in PLL Mode at 100MHz, across voltage and temperature. | -50 | 0 | 50 | ps | 1,2,3,8 |
| CLK_IN, DIF[x:0] | $t_{\text{DSPO_BYP}}$ | Input-to-output skew variation in Bypass Mode at 100MHz, across voltage and temperature, $T_{\text{AMB}} = 0$ to 70°C , default slew rate. | -250 | 0 | 250 | ps | 1,2,3,8 |
| | | Input-to-output skew variation in Bypass Mode at 100MHz, across voltage and temperature, $T_{\text{AMB}} = -40$ to $+85^{\circ}\text{C}$, default slew rate. | -350 | 0 | 350 | ps | 1,2,3,8 |
| CLK_IN, DIF[x:0] | t_{DTE} | Random differential tracking error between two 9ZX devices in Hi BW Mode. | | 3 | 5 | ps (RMS) | 1,2,3,5,8 |
| CLK_IN, DIF[x:0] | $t_{\text{DT SSE}}$ | Random differential spread spectrum tracking error between two 9ZX devices in Hi BW Mode. | | 23 | 50 | ps | 1,2,3,5,8 |
| DIF[x:0] | $t_{\text{SKEW_ALL}}$ | Output-to-output skew across all outputs, common to PLL and Bypass Mode, at 100MHz, default slew rate. | | 33 | 50 | ps | 1,2,3,8 |
| PLL Jitter Peaking | $j_{\text{peak-hibw}}$ | LOBW#_BYPASS_HIBW = 1. | | 1.3 | 2 | dB | 7,8 |
| PLL Jitter Peaking | $j_{\text{peak-lobw}}$ | LOBW#_BYPASS_HIBW = 0. | | 1.2 | 2 | dB | 7,8 |
| PLL Bandwidth | p_{llHIBW} | LOBW#_BYPASS_HIBW = 1. | 2 | 2.8 | 4 | MHz | 8,9 |
| PLL Bandwidth | p_{llLOBW} | LOBW#_BYPASS_HIBW = 0. | 0.7 | 1.1 | 1.4 | MHz | 8,9 |
| Duty Cycle | t_{DC} | Measured differentially, PLL Mode. | 45 | 50 | 55 | % | 1 |
| Duty Cycle Distortion | t_{DCD} | Measured differentially, Bypass Mode at 100MHz. | -0.5 | -0.1 | 0.5 | % | 1,10 |
| Jitter, Cycle to Cycle | $t_{\text{j cyc-cyc}}$ | PLL Mode. | | 17 | 50 | ps | 1,11 |

¹ Measured into fixed 2pF load cap. Input to output skew is measured at the first output edge following the corresponding input.

² Measured from differential cross-point to differential cross-point. This parameter can be tuned with external feedback path, if present.

³ All Bypass Mode input-to-output specs refer to the timing between an input edge and the specific output edge created by it.

⁴ This parameter is deterministic for a given device.

⁵ Measured with scope averaging ON to find mean value.

⁶ This value is programmable. See I2O programming table.

⁷ Measured as maximum pass band gain. At frequencies within the loop BW, highest point of magnification is called PLL jitter peaking.

⁸ Guaranteed by design and characterization, not 100% tested in production.

⁹ Measured at 3db down or half power point.

¹⁰ Duty cycle distortion is the difference in duty cycle between the output and the input clock when the device is operated in Bypass Mode.

¹¹ Measured from differential waveform. The device in bypass mode does not add cycle to cycle jitter.

Table 9. LP-HCSL Outputs

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Specification Limits | Units | Notes |
|------------------------|------------------|--|---------|---------|---------|----------------------|-------|-------|
| Slew Rate | dV/dt | Scope averaging on. | 2 | 2.8 | 4 | 1 – 4 | V/ns | 1,2,3 |
| Slew Rate Matching | $\Delta dV/dt$ | Slew rate matching. Scope averaging on. | | 5.3 | 20 | 20 | % | 1,4,7 |
| Maximum Voltage | Vmax | Measurement on single-ended signal using absolute value (scope averaging off). | 700 | 808 | 850 | 660 – 1150 | mV | 7,8 |
| Minimum Voltage | Vmin | | -150 | -35 | 150 | -300 | | 7,8 |
| Crossing Voltage (abs) | Vcross_abs | Scope averaging off. | 328 | 395 | 465.3 | 250 – 550 | mV | 1,5,7 |
| Crossing Voltage (var) | Δ -Vcross | Scope averaging off. | | 20 | 50 | 140 | mV | 1,6,7 |

¹ Guaranteed by design and characterization, not 100% tested in production.

² Measured from differential waveform.

³ Slew rate is measured through the Vswing voltage range centered around differential 0 V. This results in a ± 150 mV window around differential 0V.

⁴ Matching applies to rising edge rate for Clock and falling edge rate for Clock#. It is measured using a ± 75 mV window centered on the average cross point where Clock rising meets Clock# falling. The median cross point is used to calculate the voltage thresholds the oscilloscope is to use for the edge rate calculations.

⁵ Vcross is defined as voltage where Clock = Clock# measured on a component test board and only applies to the differential rising edge (i.e. Clock rising and Clock# falling).

⁶ The total variation of all Vcross measurements in any particular system. Note that this is a subset of Vcross_min/max (Vcross absolute) allowed. The intent is to limit Vcross induced modulation by setting Δ -Vcross to be smaller than Vcross absolute.

⁷ At default SMBus settings.

⁸ Includes previously separate values of +300mV overshoot and -300mV of undershoot.

Table 10. PCIe Phase Jitter Parameters for ZDB Mode

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Limits | Units | Notes |
|---|-----------------------------|-------------------------------|---------|---------|---------|--------|----------|---------|
| PCIe Phase Jitter, Low Bandwidth ZDB Mode (Common Clocked Architecture) | t _{jphPCIeG1-CC} | PCIe Gen 1 (2.5 GT/s) | | 3 | 7 | 86 | ps (p-p) | 1,2 |
| | t _{jphPCIeG2-CC} | PCIe Gen 2 Hi Band (5.0 GT/s) | | 0.09 | 0.2 | 3 | ps (RMS) | 1,2 |
| | | PCIe Gen 2 Lo Band (5.0 GT/s) | | 0.08 | 0.12 | 3.1 | ps (RMS) | 1,2 |
| | t _{jphPCIeG3-CC} | PCIe Gen 3 (8.0 GT/s) | | 0.05 | 0.07 | 1 | ps (RMS) | 1,2 |
| | t _{jphPCIeG4-CC} | PCIe Gen 4 (16.0 GT/s) | | 0.05 | 0.07 | 0.5 | ps (RMS) | 1,2,3,4 |
| | t _{jphPCIeG5-CC} | PCIe Gen 5 (32.0 GT/s) | | 0.018 | 0.022 | 0.15 | ps (RMS) | 1,2,3,5 |
| PCIe Phase Jitter, Low Bandwidth ZDB Mode (SRIS Architecture) | t _{jphPCIeG1-SRIS} | PCIe Gen 1 (2.5 GT/s) | | 8.71 | 8.73 | n/a | ps (RMS) | 1,2,6 |
| | t _{jphPCIeG2-SRIS} | PCIe Gen 2 (5.0 GT/s) | | 0.81 | 0.83 | n/a | ps (RMS) | 1,2,6 |
| | t _{jphPCIeG3-SRIS} | PCIe Gen 3 (8.0 GT/s) | | 0.329 | 0.335 | n/a | ps (RMS) | 1,2,6 |
| | t _{jphPCIeG4-SRIS} | PCIe Gen 4 (16.0 GT/s) | | 0.222 | 0.235 | n/a | ps (RMS) | 1,2,6 |
| | t _{jphPCIeG5-SRIS} | PCIe Gen 5 (32.0 GT/s) | | 0.084 | 0.091 | n/a | ps (RMS) | 1,2,6 |

Table 10. PCIe Phase Jitter Parameters for ZDB Mode (Cont.)

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Limits | Units | Notes |
|--|------------------------|-------------------------------|---------|---------|---------|----------|----------|---------|
| PCIe Phase Jitter, High Bandwidth ZDB Mode (Common Clocked Architecture) | $t_{jphPCIeG1-CC}$ | PCIe Gen 1 (2.5 GT/s) | | 5 | 7 | 86 | ps (p-p) | 1,2 |
| | $t_{jphPCIeG2-CC}$ | PCIe Gen 2 Hi Band (5.0 GT/s) | | 0.19 | 0.25 | 3 | ps (RMS) | 1,2 |
| | | PCIe Gen 2 Lo Band (5.0 GT/s) | | 0.09 | 0.13 | 3.1 | ps (RMS) | 1,2 |
| | $t_{jphPCIeG3-CC}$ | PCIe Gen 3 (8.0 GT/s) | | 0.10 | 0.13 | 1 | ps (RMS) | 1,2 |
| | $t_{jphPCIeG4-CC}$ | PCIe Gen 4 (16.0 GT/s) | | 0.10 | 0.13 | 0.5 | ps (RMS) | 1,2,3,4 |
| $t_{jphPCIeG5-CC}$ | PCIe Gen 5 (32.0 GT/s) | | 0.032 | 0.042 | 0.15 | ps (RMS) | 1,2,3,5 | |
| PCIe Phase Jitter, High Bandwidth ZDB Mode (SRIS Architecture) | $t_{jphPCIeG1-SRIS}$ | PCIe Gen 1 (2.5 GT/s) | | 8.61 | 8.63 | N/A | ps (RMS) | 1,2,6 |
| | $t_{jphPCIeG2-SRIS}$ | PCIe Gen 2 (5.0 GT/s) | | 0.88 | 0.96 | | ps (RMS) | 1,2,6 |
| | $t_{jphPCIeG3-SRIS}$ | PCIe Gen 3 (8.0 GT/s) | | 0.354 | 0.375 | | ps (RMS) | 1,2,6 |
| | $t_{jphPCIeG4-SRIS}$ | PCIe Gen 4 (16.0 GT/s) | | 0.271 | 0.305 | | ps (RMS) | 1,2,6 |
| | $t_{jphPCIeG5-SRIS}$ | PCIe Gen 5 (32.0 GT/s) | | 0.097 | 0.109 | | ps (RMS) | 1,2,6 |

¹ The Refclk jitter is measured after applying the filter functions found in PCI Express Base Specification 5.0, Revision 1.0. See the [Test Loads](#) section of the data sheet for the exact measurement setup. The worst case results for each data rate are summarized in this table. Equipment noise is removed from all results.

² Jitter measurements shall be made with a capture of at least 100,000 clock cycles captured by a real-time oscilloscope (RTO) with a sample rate of 20 GS/s or greater. Broadband oscilloscope noise must be minimized in the measurement. The measured PP jitter is used (no extrapolation) for RTO measurements. Alternately, jitter measurements may be used with a Phase Noise Analyzer (PNA) extending (flat) and integrating and folding the frequency content up to an offset from the carrier frequency of at least 200MHz (at 300MHz absolute frequency) below the Nyquist frequency. For PNA measurements for the 2.5 GT/s data rate, the RMS jitter is converted to peak-to-peak jitter using a multiplication factor of 8.83. In the case where real-time oscilloscope and PNA measurements have both been done and produce different results, the RTO result must be used.

³ SSC spurs from the fundamental and harmonics are removed up to a cutoff frequency of 2 MHz taking care to minimize removal of any non-SSC content.

⁴ Note that 0.7ps RMS is to be used in channel simulations to account for additional noise in a real system.

⁵ Note that 0.25ps RMS is to be used in channel simulations to account for additional noise in a real system.

⁶ The PCI Express Base Specification 5.0, Revision 1.0 provides the filters necessary to calculate SRIS jitter values, however, it does not provide specification limits, hence the n/a in the Limit column. SRIS values are informative only. In general, a clock operating in an SRIS system must be twice as good as a clock operating in a Common Clock system. For RMS values, twice as good is equivalent to dividing the CC value by $\sqrt{2}$. An additional consideration is the value for which to divide by $\sqrt{2}$. The conservative approach is to divide the ref clock jitter limit, and the case can be made for dividing the channel simulation values by $\sqrt{2}$, if the ref clock is close to the Tx clock input. An example for Gen4 is as follows. A “rule-of-thumb” SRIS limit would be either $0.5ps\ RMS/\sqrt{2} = 0.35ps\ RMS$ if the clock chip is far from the clock input, or $0.7ps\ RMS/\sqrt{2} = 0.5ps\ RMS$ if the clock chip is near the clock input.

Table 11. Additive PCIe Phase Jitter for Fanout Buffer Mode

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Limits | Units | Notes |
|--|------------------------|-------------------------------|---------|---------|---------|----------|----------|---------|
| Additive PCIe Phase Jitter, Fanout Buffer Mode ⁷ (Common Clocked Architecture) | $t_{jphPCIeG1-CC}$ | PCIe Gen 1 (2.5 GT/s) | | 1.3 | 1.9 | 86 | ps (p-p) | 1,2 |
| | $t_{jphPCIeG2-CC}$ | PCIe Gen 2 Hi Band (5.0 GT/s) | | 0.089 | 0.126 | 3 | ps (RMS) | 1,2 |
| | | PCIe Gen 2 Lo Band (5.0 GT/s) | | 0.023 | 0.034 | 3.1 | ps (RMS) | 1,2 |
| | $t_{jphPCIeG3-CC}$ | PCIe Gen 3 (8.0 GT/s) | | 0.044 | 0.062 | 1 | ps (RMS) | 1,2 |
| | $t_{jphPCIeG4-CC}$ | PCIe Gen 4 (16.0 GT/s) | | 0.044 | 0.062 | 0.5 | ps (RMS) | 1,2,3,4 |
| $t_{jphPCIeG5-CC}$ | PCIe Gen 5 (32.0 GT/s) | | 0.017 | 0.024 | 0.15 | ps (RMS) | 1,2,3,5 | |
| Additive PCIe Phase Jitter, Fanout Buffer Mode ⁷ (SRIS Architecture) | $t_{jphPCIeG1-SRIS}$ | PCIe Gen 1 (2.5 GT/s) | | 0.127 | 0.181 | N/A | ps (RMS) | 1,2,6 |
| | $t_{jphPCIeG2-SRIS}$ | PCIe Gen 2 (5.0 GT/s) | | 0.112 | 0.159 | | ps (RMS) | 1,2,6 |
| | $t_{jphPCIeG3-SRIS}$ | PCIe Gen 3 (8.0 GT/s) | | 0.029 | 0.042 | | ps (RMS) | 1,2,6 |
| | $t_{jphPCIeG4-SRIS}$ | PCIe Gen 4 (16.0 GT/s) | | 0.031 | 0.043 | | ps (RMS) | 1,2,6 |
| | $t_{jphPCIeG5-SRIS}$ | PCIe Gen 5 (32.0 GT/s) | | 0.027 | 0.038 | | ps (RMS) | 1,2,6 |

¹ The Refclk jitter is measured after applying the filter functions found in PCI Express Base Specification 5.0, Revision 1.0. See the [Test Loads](#) section of the data sheet for the exact measurement setup. The total Ref Clk jitter limits for each data rate are listed for convenience. The worst case results for each data rate are summarized in this table. Equipment noise is removed from all results.

² Jitter measurements shall be made with a capture of at least 100,000 clock cycles captured by a real-time oscilloscope (RTO) with a sample rate of 20 GS/s or greater. Broadband oscilloscope noise must be minimized in the measurement. The measured PP jitter is used (no extrapolation) for RTO measurements. Alternately, jitter measurements may be used with a Phase Noise Analyzer (PNA) extending (flat) and integrating and folding the frequency content up to an offset from the carrier frequency of at least 200MHz (at 300MHz absolute frequency) below the Nyquist frequency. For PNA measurements for the 2.5 GT/s data rate, the RMS jitter is converted to peak-to-peak jitter using a multiplication factor of 8.83. In the case where real-time oscilloscope and PNA measurements have both been done and produce different results, the RTO result must be used.

³ SSC spurs from the fundamental and harmonics are removed up to a cutoff frequency of 2 MHz taking care to minimize removal of any non-SSC content.

⁴ Note that 0.7ps RMS is to be used in channel simulations to account for additional noise in a real system.

⁵ Note that 0.25ps RMS is to be used in channel simulations to account for additional noise in a real system.

⁶ The PCI Express Base Specification 5.0, Revision 1.0 provides the filters necessary to calculate SRIS jitter values, however, it does not provide specification limits, hence the n/a in the Limit column. SRIS values are informative only. In general, a clock operating in an SRIS system must be twice as good as a clock operating in a Common Clock system. For RMS values, twice as good is equivalent to dividing the CC value by $\sqrt{2}$. An additional consideration is the value for which to divide by $\sqrt{2}$. The conservative approach is to divide the ref clock jitter limit, and the case can be made for dividing the channel simulation values by $\sqrt{2}$, if the ref clock is close to the Tx clock input. An example for Gen4 is as follows. A “rule-of-thumb” SRIS limit would be either $0.5ps\ RMS/\sqrt{2} = 0.35ps\ RMS$ if the clock chip is far from the clock input, or $0.7ps\ RMS/\sqrt{2} = 0.5ps\ RMS$ if the clock chip is near the clock input.

⁷ Additive jitter for RMS values is calculated by solving for “b” where $b = \sqrt{c^2 - a^2}$ and “a” is rms input jitter and “c” is rms output jitter.

Table 12. Filtered Phase Jitter Parameters – QPI/UPI, IF-UPI and DB2000Q

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Specification Limits | Units | Notes |
|------------------------------------|-------------------|--|---------|---------|---------|----------------------|----------|-------|
| Phase Jitter, ZDB Mode | t_{jphQPI_UPI} | QPI and UPI (100MHz or 133MHz, 4.8Gb/s, 6.4Gb/s 12UI) | | 0.19 | 0.39 | 0.5 | ps (RMS) | 1,2 |
| | | QPI and UPI (100MHz, 8.0Gb/s, 12UI) | | 0.10 | 0.15 | 0.3 | ps (RMS) | 1,2 |
| | | QPI and UPI (100MHz, ≤ 11.4 Gb/s, 12UI) | | 0.08 | 0.12 | 0.2 | ps (RMS) | 1,2 |
| Additive Phase Jitter, Fanout Mode | t_{jphQPI_UPI} | QPI and UPI (100MHz or 133MHz, 4.8Gb/s, 6.4Gb/s 12UI) | | 0.08 | 0.22 | N/A | ps (RMS) | 1,2,3 |
| | | QPI and UPI (100MHz, 8.0Gb/s, 12UI) | | 0.04 | 0.08 | | ps (RMS) | 1,2,3 |
| | | QPI and UPI (100MHz, ≤ 11.4 Gb/s, 12UI) | | 0.03 | 0.06 | | ps (RMS) | 1,2,3 |
| | t_{jphIF_UPI} | IF-UPI, Lo-BW ZDB Mode | | 0.10 | 0.13 | 1 | ps (RMS) | 1,4,5 |
| | | IF-UPI, Hi-BW ZDB Mode | | 0.17 | 0.20 | 1 | ps (RMS) | 1,4,5 |
| | | IF-UPI, Fanout Mode | | 0.06 | 0.07 | 1 | ps (RMS) | 1,4 |
| | $t_{jphDB2000Q}$ | DB2000Q, Fanout Mode | | 28 | 39 | 80 | fs (RMS) | 1,4,5 |

¹ Applies to all differential outputs, guaranteed by design and characterization. See [Test Loads](#) for measurement setup details.

² Calculated from Intel™-supplied clock jitter tool.

³ For RMS values, additive jitter is calculated by solving for “b” where $b = \sqrt{c^2 - a^2}$, “a” is rms input jitter and “c” is rms total jitter.

⁴ Calculated from phase noise analyzer with Intel-specified brick-wall filter applied. This is an additive jitter specification regardless of buffer operating mode.

⁵ The IF-UPI specification is an additive specification, regardless of the buffer operating mode. The enhanced 9ZML devices meet this specification in all operating modes.

Table 13. Phase Jitter Parameters – 12kHz to 20MHz

| Parameter | Symbol | Conditions | Minimum | Typical | Maximum | Specification Limits | Units | Notes |
|---|---------------------|-------------------------------------|---------|---------|---------|----------------------|----------|-------|
| 12kHz–20MHz Additive Phase Jitter, Fanout Buffer Mode | $t_{jph12k-20MFOB}$ | Fanout Buffer Mode, SSC OFF, 100MHz | | 130 | | N/A | fs (RMS) | 1,2,3 |

¹ Applies to all differential outputs, guaranteed by design and characterization. See [Test Loads](#) for measurement setup details.

² 12kHz to 20MHz brick wall filter.

³ For RMS values, additive jitter is calculated by solving for “b” where $b = \sqrt{c^2 - a^2}$, “a” is rms input jitter and “c” is rms total jitter.

Power Management

Table 14. Power Management

| Inputs | | Control Bits | Outputs | | PLL State |
|-------------|---------|--------------|---------|----------|-----------|
| CKPWRGD_PD# | DIF_IN | SMBus EN bit | DIF_x | FBOUT_NC | |
| 0 | X | X | Low/Low | Low/Low | Off |
| 1 | Running | 0 | Low/Low | Running | On |
| | | 1 | Running | Running | On |

Table 15. Power Connections

| Pin Number | | | Description |
|-----------------|------------------------|--|--------------|
| V _{DD} | V _{DDIO} | GND | |
| 1 | | 2 | Analog PLL |
| 9 | | 8 | Analog input |
| 28, 45, 64 | 21, 33, 40, 52, 57, 69 | 16, 22, 27, 34, 39, 46, 51, 58, 63, 70 | DIF clocks |

Table 16. PLL Operating Mode

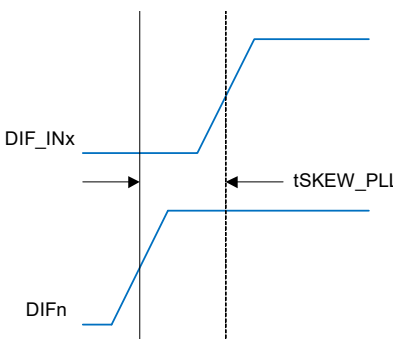
| HIBW_BYPM_LOBW# | Byte0[7:6] |
|--------------------|------------|
| Low (PLL Low BW) | 00 |
| Mid (Bypass) | 01 |
| High (PLL High BW) | 11 |

Note: PLL is OFF in Bypass Mode.

Skew Programming

Table 17. Skew Programming

| Skew[2:0] | Skew Steps | Skew (ps) |
|-----------|------------|-----------|
| 000 | 0 | 0.00 |
| 001 | 1 | -416.67 |
| 010 | 2 | -833.33 |
| 011 | 3 | -1250.00 |
| 100 | 4 | -1666.67 |
| 101 | 5 | -2083.33 |
| 110 | 6 | -2500.00 |
| 111 | 7 | -2916.67 |



Test Loads

Figure 2. Test Load for AC/DC Measurements

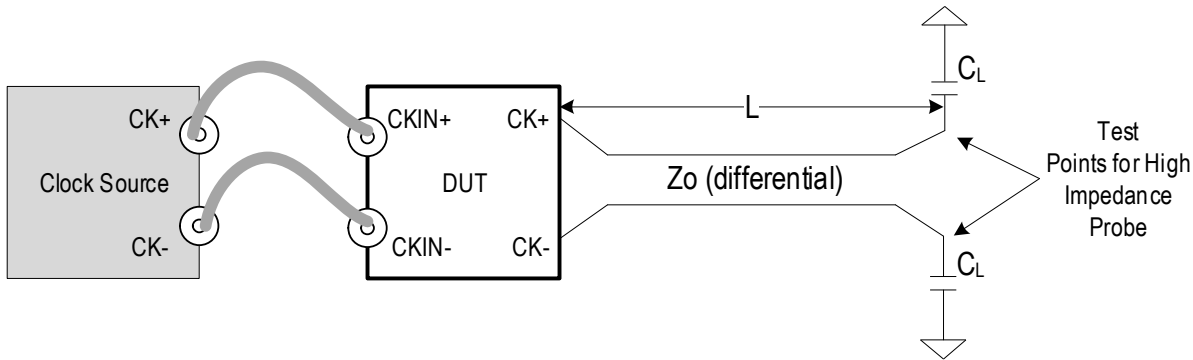


Table 18. Parameters for AC/DC Measurements

| Clock Source | Device Under Test (DUT) | Rs (Ω) | Differential Zo (Ω) | L (cm) | CL (pF) | Parameters Measured |
|--------------|-------------------------|-----------------|------------------------------|--------|---------|---------------------|
| SMA100B | 9ZML124x | Internal | 100 | 25.4 | 2 | AC/DC parameters |
| SMA100B | 9ZML125x | Internal | 85 | 25.4 | 2 | |

Figure 3. Test Load for Phase Jitter Measurements using Phase Noise Analyzer

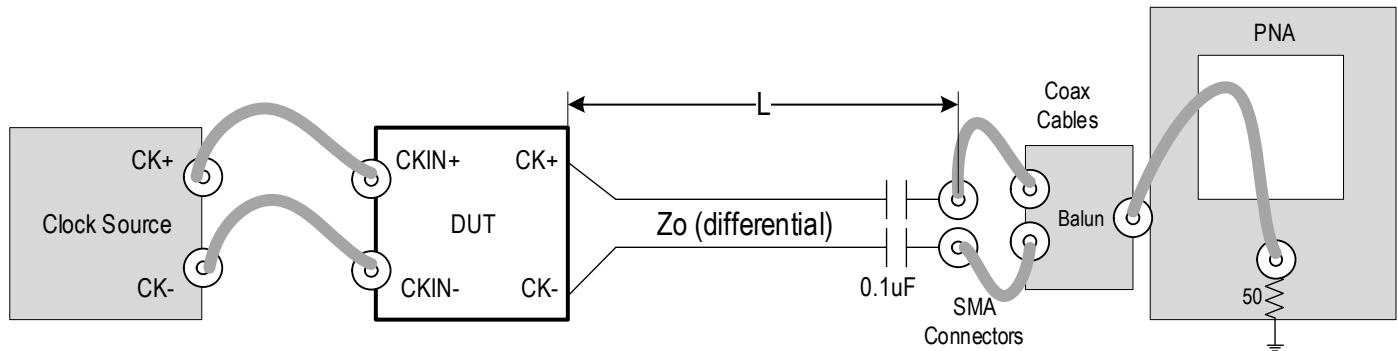


Table 19. Parameters for Phase Jitter Measurements using Phase Noise Analyzer

| Clock Source | Device Under Test (DUT) | Rs (Ω) | Differential Zo (Ω) | L (cm) | CL (pF) | Notes | Parameters Measured |
|--------------|-------------------------|-----------------|------------------------------|--------|---------|-------------|-----------------------|
| SMA100B | 9ZML124x | Internal | 100 | 25.4 | N/A | Fanout Mode | PCIe, IF-UPI, DB2000Q |
| 9FGV1006 | 9ZML124x | Internal | 100 | 25.4 | | ZDB Mode | |
| SMA100B | 9ZML125x | Internal | 85 | 25.4 | | Fanout Mode | |
| 9FGV1006 | 9ZML125x | Internal | 85 | 25.4 | | ZDB Mode | |

Figure 4. Test Load for Phase Jitter Measurements using Oscilloscope

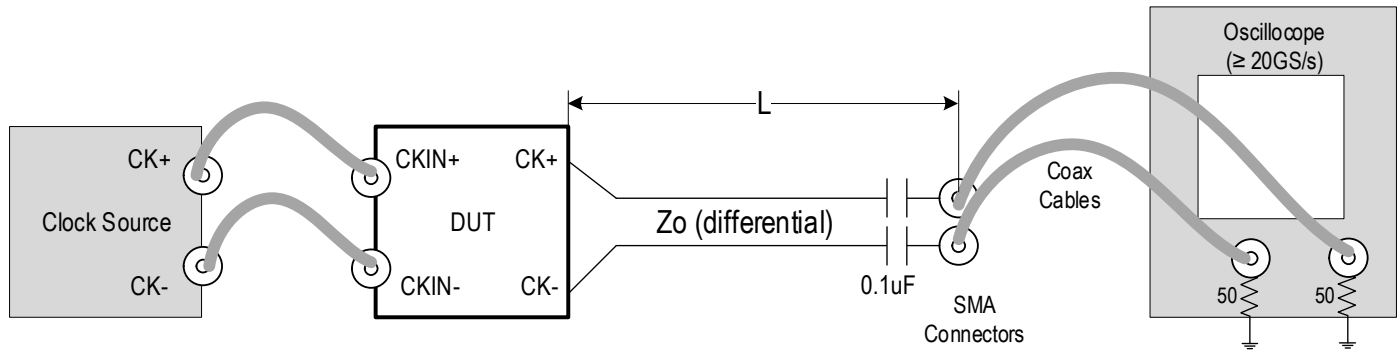


Table 20. Parameters for Phase Jitter Measurements using Oscilloscope

| Clock Source | Device Under Test (DUT) | Rs (Ω) | Differential Zo (Ω) | L (cm) | CL (pF) | Notes | Parameters Measured |
|--------------|-------------------------|----------|---------------------|--------|---------|-------------|---------------------|
| SMA100B | 9ZML124x | Internal | 100 | 25.4 | N/A | Fanout Mode | QPI/UPI |
| 9FGV1006 | 9ZML124x | Internal | 100 | 25.4 | | ZDB Mode | |
| SMA100B | 9ZML125x | Internal | 85 | 25.4 | | Fanout Mode | |
| 9FGV1006 | 9ZML125x | Internal | 85 | 25.4 | | ZDB Mode | |

Alternate Terminations

The LP-HCSL output can easily drive other logic families. See [“AN-891 Driving LVPECL, LVDS, and CML Logic with “Universal” Low-Power HCSL Outputs”](#) for termination schemes for LVPECL, LVDS, CML and SSTL.

General SMBus Serial Interface Information

How to Write

- Controller (host) sends a start bit
- Controller (host) sends the write address
- Renesas clock will **acknowledge**
- Controller (host) sends the beginning byte location = N
- Renesas clock will **acknowledge**
- Controller (host) sends the byte count = X
- Renesas clock will **acknowledge**
- Controller (host) starts sending Byte N through Byte N+X-1
- Renesas clock will **acknowledge** each byte **one at a time**
- Controller (host) sends a stop bit

| Index Block Write Operation | | |
|-----------------------------|-----------|--------------------------|
| Controller (Host) | | Renesas (Slave/Receiver) |
| T | starT bit | |
| Slave Address | | |
| WR | WRite | |
| Beginning Byte = N | | ACK |
| | | ACK |
| Data Byte Count = X | | ACK |
| Beginning Byte N | | ACK |
| O | | O |
| O | | O |
| O | | O |
| Byte N + X - 1 | | ACK |
| P | stoP bit | |

How to Read

- Controller (host) will send a start bit
- Controller (host) sends the write address
- Renesas clock will **acknowledge**
- Controller (host) sends the beginning byte location = N
- Renesas clock will **acknowledge**
- Controller (host) will send a separate start bit
- Controller (host) sends the read address
- Renesas clock will **acknowledge**
- Renesas clock will send the data byte count = X
- Renesas clock sends Byte N+X-1
- Renesas clock sends **Byte 0 through Byte X (if X_(H) was written to Byte 8)**
- Controller (host) will need to acknowledge each byte
- Controller (host) will send a not acknowledge bit
- Controller (host) will send a stop bit

| Index Block Read Operation | | |
|----------------------------|-----------------|-------------------|
| Controller (Host) | | Renesas |
| T | starT bit | |
| Slave Address | | |
| WR | WRite | |
| Beginning Byte = N | | ACK |
| | | ACK |
| RT | Repeat starT | |
| Slave Address | | |
| RD | ReaD | |
| | | ACK |
| | | Data Byte Count=X |
| ACK | | Beginning Byte N |
| ACK | | O |
| O | | O |
| O | | O |
| O | | |
| | | Byte N + X - 1 |
| N | Not acknowledge | |
| P | stoP bit | |

Table 21. SMBus Addresses

| Pin | | SMBus Address | |
|-----------|-----------|-------------------|----------|
| SADR1_tri | SADR0_tri | 9ZML1245/9ZML1255 | 9ZML1256 |
| 0 | 0 | D8 | D8 |
| 0 | M | DA | DA |
| 0 | 1 | DE | DE |
| M | 0 | C2 | — |
| M | M | C4 | — |
| M | 1 | C6 | — |
| 1 | 0 | CA | — |
| 1 | M | CC | — |
| 1 | 1 | CE | — |

Table 22. Byte 0: PLL Mode and Frequency Select Register

| Byte 0 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|-------------------------|-------------------------------|-------------------------------|-----------------------|--------------------------------------|---|----------------------|-----------------------------|--------------------------|
| Control Function | PLL Operating Mode Readback 1 | PLL Operating Mode Readback 0 | Input Select Readback | SMB_WRTLOCK_Readback (9ZML1256 only) | Enable S/W control of PLL BW and Input select | PLL Operating Mode 1 | PLL Operating Mode 0 | Input Select control bit |
| Type | R | R | R | R | RW | RW | RW | RW |
| 0 | 00 = Low BW ZDB Mode | 01 = Bypass (Fanout Buffer) | DIF_INB is selected | Pin is Low | HW Latch | 00 = Low BW ZDB Mode | 01 = Bypass (Fanout Buffer) | DIF_INB is selected |
| 1 | 10 = Reserved | 11 = High BW ZDB Mode | DIF_INA is selected | Pin is High | SMBus Control | 10 = Reserved | 11 = High BW ZDB Mode | DIF_INA is selected |
| Name | PLL Mode bit [1] | PLL Mode bit [0] | SEL_A_B# | SMB_WRTLOCK_RB | PLL_SEL_SW_EN | PLL Mode bit [1] | PLL Mode bit [0] | SEL_A_B# |
| Default | Latch | Latch | Pin | 0 | 0 | 1 | 1 | 1 |

Note: Setting bit 3 to '1' allows the user to override the latch value from pin 5 via use of bits 2 and 1. The system may require a warm system reset if the user changes these bits. The clock itself does not require a reset. Setting bit 3 to a '1' also allows the user to use bit 0 to control the input select.

Table 23. Byte 1: Output Control Register 1

| Byte 1 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|------------------|-----------------|---------|---------|---------|---------|---------|---------|---------|
| Control Function | Output Enable | | | | | | | |
| Type | RW | | | | | | | |
| 0 | Low/Low | | | | | | | |
| 1 | OE# Pin Control | | | | | | | |
| Name | DIF7_en | DIF6_en | DIF5_en | DIF4_en | DIF3_en | DIF2_en | DIF1_en | DIF0_en |
| Default | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Table 24. Byte 2: Output Control Register 2

| Byte 2 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|------------------|----------|----------|----------|----------|-----------------|----------|---------|---------|
| Control Function | Reserved | Reserved | Reserved | Reserved | Output_enable | | | |
| Type | | | | | RW | | | |
| 0 | | | | | Low/Low | | | |
| 1 | | | | | OE# Pin Control | | | |
| Name | | | | | DIF11_en | DIF10_en | DIF9_en | DIF8_en |
| Default | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |

Bytes 3 is Reserved

Table 25. Byte 4: Input Detect Readback Register

| Byte 4 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|------------------|-------------------------------|--------------------------------|----------|----------|----------|----------|----------|----------|
| Control Function | Input clock presence detect | Enable or disable Input Detect | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| Type | R | RW | | | | | | |
| 0 | Selected Input is not present | Input Detect is Disabled | | | | | | |
| 1 | Selected input is present | Input Detect is Enabled | | | | | | |
| Name | Input_Detect_RB | Input_Detect_En | | | | | | |
| Default | See Note | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Notes on Byte 4:

- Clock detect circuit monitors selected input clock (A or B)
- Parks the output clocks in a low/low state ~150ns after input clock disappears.
- When enabled (Byte 4, bit 6 set to '1'):
 - Bypass Mode minimum operating frequency is 25MHz

- ZDB mode frequency range is unchanged
- Realtime absence or presence of selected input clock may be read back from Byte 4, bit 7

When disabled (Byte 4, bit 6 set to '0' - default):

- Bypass Mode minimum operating frequency is unchanged
- ZDB mode frequency range is unchanged
- Byte 4, bit 7 reads 0
- Part behaves like existing devices

Table 26. Byte 5: Revision and Vendor ID Register

| Byte 5 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|------------------|--------------|------|------|------|--------------------|------|------|------|
| Control Function | Revision ID | | | | Vendor ID | | | |
| Type | R | R | R | R | R | R | R | R |
| 0 | A rev = 0010 | | | | IDT/Renesas = 0001 | | | |
| 1 | | | | | | | | |
| Name | RID3 | RID2 | RID1 | RID0 | VID3 | VID2 | VID1 | VID0 |
| Default | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 |

Table 27. Byte 6: Device ID Register

| Byte 6 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|----------------------|-----------------|---------|---------|---------|---------|---------|---------|---------|
| Control Function | Device ID | | | | | | | |
| Type | R | R | R | R | R | R | R | R |
| 0 | See table below | | | | | | | |
| 1 | | | | | | | | |
| Name | DevID 7 | DevID 6 | DevID 5 | DevID 4 | DevID 3 | DevID 2 | DevID 1 | DevID 0 |
| 9ZML1245 9ZML1255 | 0hC5 | | | | | | | |
| 9ZML1256 | 0hC5 | | | | | | | |

Table 28. Byte 7: Byte Count Register

| Byte 7 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|------------------|----------|----------|----------|---|------|------|------|------|
| Control Function | Reserved | Reserved | Reserved | Writing to this register configures how many bytes will be read back on a block read. | | | | |
| Type | | | | RW | RW | RW | RW | RW |
| 0 | | | | Default value is 8. | | | | |
| 1 | | | | | | | | |
| Name | | | | BC4 | BC3 | BC2 | BC1 | BC0 |
| Default | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |

Table 29. Byte 8: Output Skew Register A (when Input Clock A is selected)

| Byte 8 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|------------------|----------|----------|----------|----------|----------|---|----------------|----------------|
| Control Function | Reserved | Reserved | Reserved | Reserved | Reserved | Channel A Output delay programming (early) | | |
| Type | | | | | | RW | RW | RW |
| 0 | | | | | | Binary value of number of VCO periods that outputs will be pulled earlier than input. | | |
| 1 | | | | | | | | |
| Name | | | | | | I2O_FB_A Skew2 | I2O_FB_A Skew1 | I2O_FB_A Skew0 |
| Default | 0 | 0 | 0 | 0 | 0 | 0 | 0 | |

Note: For example, at 2.4GHz, each VCO period is 416.7ps and there are 24 VCO periods in a 100MHz output. B10[6] determines the accumulative mode of the delay steps. By default, accumulation is disabled and the part operates in absolute mode. This means that the maximum number of skew tuning steps is '111' or 7. In absolute mode, each skew tuning value must be higher than the value previously written. For example, writing '010' pulls the output a total 2 VCO periods earlier. Writing '000' or '001' is not allowed. Writing '011' will pull the output one additional cycle earlier for a total of 3. In accumulative mode, each write to bits [2:0] will pull the output a early by that number of VCO periods. Writing '110' 4 times would pull the output back in phase with the input (360 degrees). Writing '001' twice will accomplish the same result as writing '010' once - pulling the output 2 VCO periods earlier.

Table 30. Byte 9: Output Skew Register B (when Input Clock B is selected)

| Byte 9 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|------------------|----------|----------|----------|----------|----------|---|----------------|----------------|
| Control Function | Reserved | Reserved | Reserved | Reserved | Reserved | Channel A Output delay programming (early) | | |
| Type | | | | | | RW | RW | RW |
| 0 | | | | | | Binary value of number of VCO periods that outputs will be pulled earlier than input. | | |
| 1 | | | | | | | | |
| Name | | | | | | I2O_FB_A Skew2 | I2O_FB_A Skew1 | I2O_FB_A Skew0 |
| Default | 0 | 0 | 0 | 0 | 0 | 0 | 0 | |

The note for Byte 8 also applies to Byte 9.

Table 31. Byte 10: Global Amplitude and Delay Step Control Register

| Byte 10 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|------------------|--------------------------------|--------------------------------|----------|----------|----------|--|--------|--------|
| Control Function | Bypass the skew tuning circuit | Enable accumulative step delay | Reserved | Reserved | Reserved | Global Differential Amplitude Control | | |
| Type | RW | RW | | | | RW | RW | RW |
| 0 | Skew Circuit Enabled | Disable | | | | 0xx = Reserved, 100 = 750mV 101 = 850mV, 110 = 950mV, 111 = Reserved | | |
| 1 | Skew Circuit Bypassed | Enable | | | | | | |
| Name | Skew Bypass | Accumulate Mode | | | | AMP[2] | AMP[1] | AMP[0] |
| Default | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 1 |

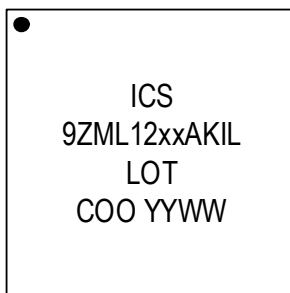
Table 32. Byte 11: Output Configuration Register

| Byte 11 | Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|------------------|--------------------------|----------------------------|--|-----------|--|----------|-------------------------|---------------|
| Name | SR_OUT | SR_FB | ZO_OUT[1] | ZO_OUT[0] | ZO_FB[1] | ZO_FB[0] | STP[1] | STP[0] |
| Control Function | Output slew rate control | Feedback slew rate control | Output differential impedance | | Feedback differential impedance | | Differential Stop State | |
| Type | RW | RW | RW | RW | RW | RW | RW | RW |
| 0 | Slow | Slow | 00 = 33ohm 01 = 85ohm (9ZML125x) 10 = 100ohm (9ZML124x) 11 = Reserved | | 00 = 33ohm 01 = 85ohm (9ZML125x) 10 = 100ohm (9ZML124x) 11 = Reserved | | 00 = Low/Low | 01 = HiZ/HiZ |
| 1 | Fast | Fast | | | | | 10 = High/Low | 11 = Low/High |
| Default | 1 | 1 | X | X | X | X | 0 | 0 |

Package Outline Drawings

The package outline drawings are located at the end of this document and are accessible from the Renesas website (see [Ordering Information](#) for POD links). The package information is the most current data available and is subject to change without revision of this document.

Marking Diagram



- Lines 1 and 2: truncated part number
 - “xx” denotes 45, 55, or 56
- Line 3: “LOT” denotes the lot number.
- Line 4:
 - “COO” denotes country of origin.
 - “YYWW” denotes the last two digits of the year and work week that the part was assembled.

Ordering Information

Table 33. Ordering Information

| Orderable Part Number | Output Impedance | Description | Number of Clock Outputs | Package | Temperature Range | Part Number Suffix and Shipping Method |
|-----------------------|------------------|--|-------------------------|-------------------------------|-------------------|--|
| 9ZML1245AKILF | 100 | 2:12 ZDB/FOB | 12 | 10 × 10 × 0.9 mm 72-VFQFPN | -40°C to +85°C | None = Trays “T” = Tape and Reel, Pin 1 Orientation: EIA-481C (see Table 34 for more details) |
| 9ZML1245AKILFT | | | | | | |
| 9ZML1255AKILF | 85 | 2:12 ZDB/FOB | | | | |
| 9ZML1255AKILFT | | | | | | |
| 9ZML1256AKILF | 85 | 2:12 ZDB/FOB with SMBus Write Protection | | | | |
| 9ZML1256AKILFT | | | | | | |

“A” is the device revision designator (will not correlate with the datasheet revision).

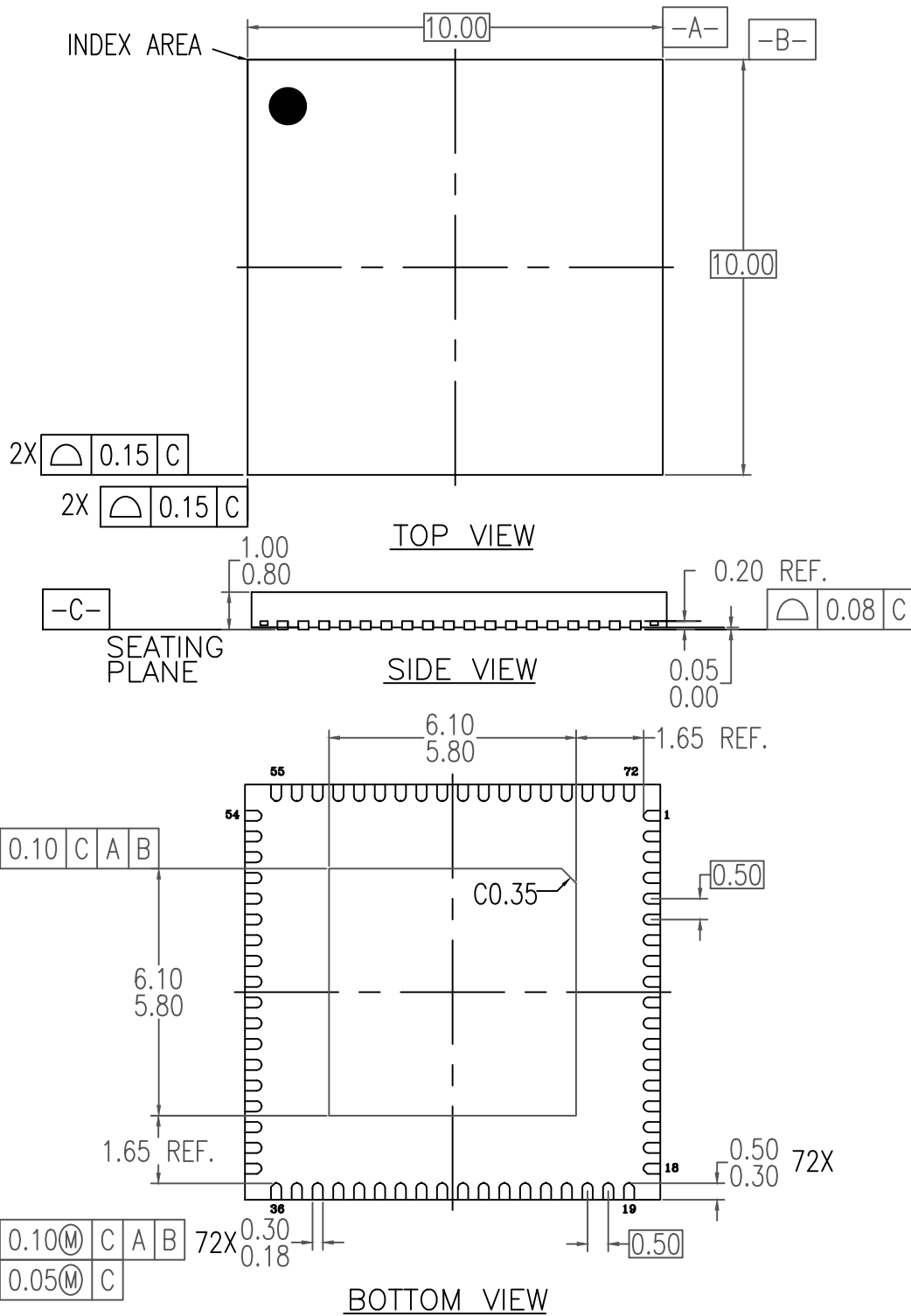
“LF” denotes Pb-free configuration, RoHS compliant; “T” is the orderable suffix for Tape and Reel packaging.

Table 34. Pin 1 Orientation in Tape and Reel Packaging

| Part Number Suffix | Pin 1 Orientation | Illustration |
|--------------------|------------------------|--------------|
| T | Quadrant 1 (EIA-481-C) | |

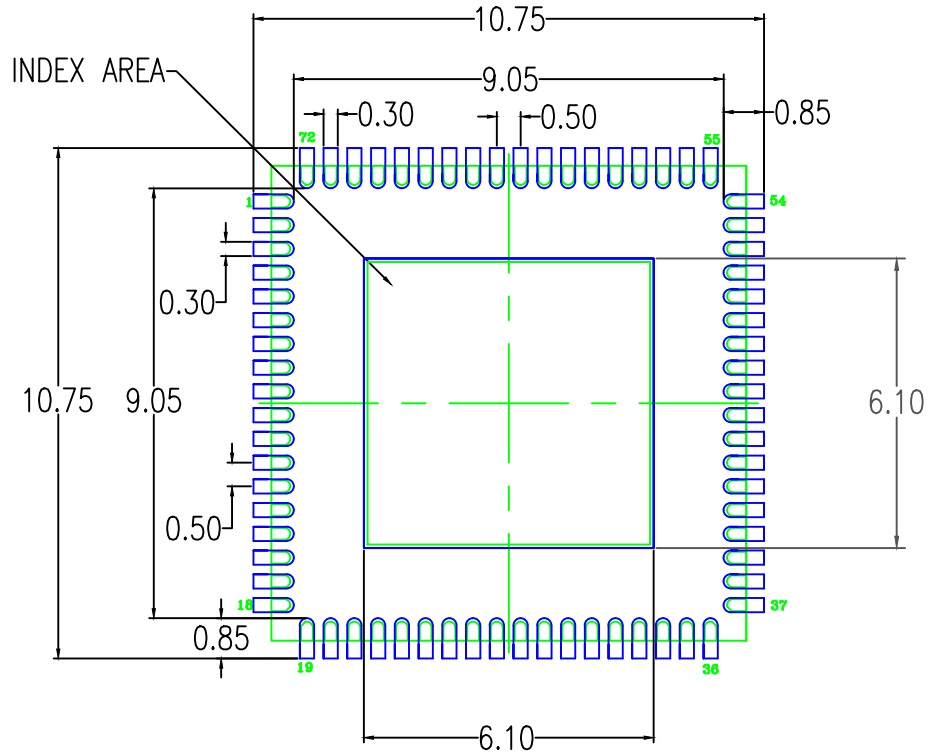
Revision History

| Revision Date | Description of Change |
|-------------------|--|
| May 12, 2021 | Updates to Byte 0, bit 0 and bit 5 defaults. |
| April 29, 2021 | Corrected typo in Byte 0, bit 0. Removed reference to 9ZXL devices from footnote for Byte 0. |
| March 17, 2021 | Added Byte 11. |
| October 9, 2020 | Updated Byte 0, bit Control Function. Updated the footnote for this register table. |
| May 19, 2020 | Rebranded document. |
| November 14, 2019 | Combined 9ZML1245, 9ZML1255, and 9ZML1256 datasheets into one single document. |



NOTES:

1. ALL DIMENSIONING AND TOLERANCING CONFORM TO ANSI Y14.5M-1994
2. ALL DIMENSIONS ARE IN MILLIMETERS.
3. INDEX AREA (PIN1 IDENTIFIER)



RECOMMENDED LAND PATTERN DIMENSION

NOTES:

1. ALL DIMENSIONS ARE IN MM. ANGLES IN DEGREES.
2. TOP DOWN VIEW. AS VIEWED ON PCB.
3. COMPONENT OUTLINE SHOWS FOR REFERENCE IN GREEN.
4. LAND PATTERN IN BLUE. NSMD PATTERN ASSUMED.
5. LAND PATTERN RECOMMENDATION PER IPC-7351B GENERIC REQUIREMENT FOR SURFACE MOUNT DESIGN AND LAND PATTERN.

| Package Revision History | | |
|--------------------------|---------|--|
| Date Created | Rev No. | Description |
| Sept 3, 2019 | Rev 03 | Update P1 Dimension from 5.8 to 5.95 mm SQ |
| May 8, 2017 | Rev 02 | Change Package Code QFN to VFQFPN |