ACNW3190 5.0 Amp High Output Current IGBT Gate Drive Optocoupler

Data Sheet

Description

The ACNW3190 contains an AlGaAs LED, which is optically coupled to an integrated circuit with a power output stage. This optocoupler is ideally suited for driving power IGBTs and MOSFETs used in motor control inverter applications. The high operating voltage range of the output stage provides the drive voltages required by gate controlled devices. The voltage and high peak output current supplied by this optocoupler make it ideally suited for direct driving IGBTs with ratings up to 1200V/200 A, 600 V/300 A. For IGBTs with higher ratings, the ACNW3190 can be used to drive a discrete power stage which drives the IGBT gate. The ACNW3190 has the highest insulation voltage of V_{IORM} =1414 Vpeak in the IEC/ EN/DIN EN 60747-5-5.

Functional Diagram

TRUTH TABLE

A 0.1 μF bypass capacitor must be connected between pins 5 and 8.

Features

- 5.0 A Maximum Peak Output Current
- 15 kV/μs Minimum Common Mode Rejection (CMR) at V_{CM} = 1500 V
- \bullet 0.5 V Maximum Low Level Output Voltage (V_{OL}) Eliminates Need for Negative Gate Drive
- $I_{CC} = 5$ mA Maximum Supply Current
- Under Voltage Lock-Out Protection (UVLO) with Hysteresis
- Wide Operating V_{CC} Range: 15 to 30 Volts
- 500 ns Maximum Switching Speeds
- Industrial Temperature Range: -40°C to 100°C
- Safety Approval

UL Recognized 5000 Vrms for 1 min.

CSA Approval IEC/EN/DIN EN 60747-5-5 Approved $V_{IORM} = 1414 V_{peak}$

Applications

- IGBT/MOSFET Gate Drive
- AC/Brushless DC Motor Drives
- Industrial Inverters
- Switch Mode Power Supplies

CAUTION: It is advised that normal static precautions be taken in handling and assembly of this component to prevent damage and/or degradation which may be induced by ESD.

Ordering Information

ACNW3190 is UL Recognized with 5000Vrms for 1 minute per UL1577.

To order, choose a part number from the part number column and combine with the desired option from the option column to form an order entry.

Example 1:

 ACNW3190-500E to order product of 400mil DIP Gull Wing Surface Mount package in Tape and Reel packaging with IEC/EN/DIN EN 60747-5-5 Safety Approval in RoHS compliant.

Example 2:

ACNW3190-000E to order product of 400mil DIP package in tube packaging and RoHS compliant.

Option datasheets are available. Contact your Avago sales representative or authorized distributor for information.

Package Outline Drawings

ACNW3190 Gull Wing Surface Mount Option 300 Outline Drawing

Dimensions in inches (millimeters)

Note: Floating Lead Protrusion is 0.25 mm (10 mils) max.

Recommended Pb-Free IR Profi le

Recommended reflow condition as per JEDEC Standard, J-STD-020 (latest revision). Non-Halide Flux should be used.

Dependence of Safety Limiting Values on Temperature

Note:

 The Thermal Derating Graph above is in relation to Figure 30 and Figure 31 and $S = 2$ cm.

Table 1. IEC/EN/DIN EN 60747-5-5 Insulation Characteristics*

Regulatory Information

The ACNW3190 is approved by the following organizations:

IEC/EN/DIN EN 60747-5-5

Maximum Working Insulation Voltage $V_{\text{IORM}} = 1414V_{\text{PEAK}}$

UL

Approval under UL 1577, component recognition program up to $V_{\text{ISO}} = 5000 V_{\text{RMS}}$ expected prior to product release. File E55361.

CSA

Approval under CSA Component Acceptance Notice #5, File CA 88324 expected prior to product release.

Isolation characteristics are guaranteed only within the safety maximum ratings which must be ensured by protective circuits in application. Surface mount classification is class A in accordance with CECCOO802.

** Refer to the optocoupler section of the Isolation and Control Components Designer's Catalog, under Product Safety Regulations section IEC/EN/ DIN EN 60747-5-5, for a detailed description of Method a and Method b partial discharge test profiles.

Table 2. Insulation and Safety Related Specifications

Notes:

All Avago data sheets report the creepage and clearance inherent to the optocoupler component itself. These dimensions are needed as a starting point for the equipment designer when determining the circuit insulation requirements. However, once mounted on a printed circuit board, minimum creepage and clearance requirements must be met as specifi ed for individual equipment standards. For creepage, the shortest distance path along the surface of a printed circuit board between the solder fi llets of the input and output leads must be considered (the recommended Land Pattern does not necessarily meet the minimum creepage of the device). There are recommended techniques such as grooves and ribs which may be used on a printed circuit board to achieve desired creepage and clearances. Creepage and clearance distances will also change depending on factors such as pollution degree and insulation level.

Table 3. Absolute Maximum Ratings

Table 4. Recommended Operating Conditions

Table 5. Electrical Specifications (DC)

Unless otherwise noted, all typical values are at T_A = 25°C, V_{CC} - V_{EE} = 30 V, V_{EE} = Ground; all Minimum/Maximum specifications are at Recommended Operating Conditions (T_A = -40 to 100°C, I_{F(ON)} = 10 to 16 mA, V_{F(OFF)} = -3.6 to 0.8 V, V_{CC} $= 15$ to 30 V, V_{EE} $=$ Ground)

Table 6. Switching Specifications (AC)

Unless otherwise noted, all typical values are at $T_A = 25^{\circ}$ C, V_{CC} - V_{EE} = 30 V, V_{EE} = Ground; all Minimum/Maximum specifications are at Recommended Operating Conditions (T_A = -40 to 100°C, I_{F(ON)} = 10 to 16 mA, V_{F(OFF)} = -3.6 to 0.8 V, $V_{CC} = 15$ to 30 V, $V_{EE} =$ Ground).

Table 7. Package Characteristics

Unless otherwise noted, all typical values are at $T_A = 25^{\circ}C$; all Minimum/Maximum specifications are at Recommended Operating Conditions.

The Input-Output Momentary Withstand Voltage is a dielectric voltage rating that should not be interpreted as an input-output continuous voltage rating. For the continuous voltage rating, refer to your equipment level safety specification or Avago Technologies Application Note 1074 entitled "Optocoupler Input-Output Endurance Voltage."

Notes:

- 1. Derate linearly above 70 $^{\circ}$ C free-air temperature at a rate of 0.3 mA/ $^{\circ}$ C.
- 2. Maximum pulse width = 10 µs. This value is intended to allow for component tolerances for designs with I_0 peak minimum = 4.0 A. See Applications section for additional details on limiting I_{OH} peak.
- 3. Derate linearly above 70 $^{\circ}$ C free-air temperature at a rate of 4.8 mA/ $^{\circ}$ C.
- 4. Derate linearly above 70°C free-air temperature at a rate of 5.4 mA/ °C.
- 5. Maximum pulse width = $50 \mu s$.
- 6. In this test V_{OH} is measured with a dc load current. When driving capacitive loads V_{OH} will approach V_{CC} as I_{OH} approaches zero amps.
- 7. Maximum pulse width $= 1$ ms.
- 8. In accordance with UL1577, each optocoupler is proof tested by applying an insulation test voltage ≥6000 Vrms for 1 second (leakage detection current limit, $I_{I-O} \leq 5 \mu A$).
- 9. Device considered a two-terminal device: pins 1, 2, 3, and 4 shorted together and pins 5, 6, 7, and 8 shorted together.
- 10. The difference between tPHL and t_{PLH} between any two ACNW3190 parts under the same test condition.
- 11. Pins 1 and 4 need to be connected to LED common.
- 12. Common mode transient immunity in the high state is the maximum tolerable dVCM/dt of the common mode pulse, VCM, to assure that the output will remain in the high state (i.e., $V_O > 15.0 V$).
- 13. Common mode transient immunity in a low state is the maximum tolerable dVCM/dt of the common mode pulse, VCM, to assure that the output will remain in a low state (i.e., V_Q < 1.0 V).
- 14. This load condition approximates the gate load of a 1200 V/100A IGBT.
- 15. Pulse Width Distortion (PWD) is defined as |tPHL-tPLH| for any given device.

Figure 1. V_{OH} vs. Temperature **Figure 2. Io_H vs. Temperature**

Figure 5. I_{OL} vs. Temperature **Figure 6. V_{OL} vs.** I_{OL}.

Figure 3. V_{OH} vs. 1_{OH} Figure 4. V_{OL} vs. Temperature

Figure 7. I_{CC} vs. Temperature **Figure 8.** I_{CC} vs. V_{CC}

Figure 9. I_{FLH} vs. Temperature Figure 10. Propagation delay vs. V_{CC}

Figure 11. Propagation delay vs. I_F \blacksquare **Figure 12. Propagation delay vs. Temperature**

Figure 13. Propagation Delay vs. Rg **Figure 14. Propagation Delay vs. Cg**

Figure 17. I_{OH} Test Circuit **Figure 18. I_{OL} Test Circuit**

 \vert 1 8 \downarrow 100 mA 2 7 **V_{OL}** 0.1 μF \leftarrow $\left(\frac{1}{2}\right)^{V_{CC}}$ = 15
to 30 V + – 6 3 z 4 5 \bigtriangledown

Figure 19. V_{OH} Test Circuit **Figure 20. V_{OL} Test Circuit**

Figure 21. IFLH Test Circuit Figure 22. UVLO Test Circuit

Figure 23. t_{PLH}, t_{PHL}, t_r, and t_f Test Circuit and Waveforms

Application Notes

Figure 25 and 26 show two recommended application circuits. Figure 25 show a single power supply gate driver using the driver's maximum V_{OL} value of 0.5V. Figure 26 show a dual power supply gate driver circuit which is applicable for higher power IGBT driving due to the existence of higher miller capacitance in these IGBTs.

Figure 25. Recommended LED drive and application circuit

For high side bootstrap driving, note that the bypass capacitor of 0.1μF in parallel with 10μF or above to be connected across VCC and VEE is important to deliver high peak output current.

Figure 26. ACNW3190typical application circuit with negative IGBT gate drive

Selecting the Gate Resistor (Rg) to minimize IGBT Switching Losses.

Step 1: Calculate Rg Minimum from the IOL Peak Specification

The IGBT and Rg in Figure 26 can be analyzed as a simple RC circuit with a voltage supplied by the ACNW3190. The operating temperature is 100°C.

$$
Rg \ge \frac{(V_{CC} - V_{EE} - V_{OL})}{I_{OLPEAK}}
$$

$$
= \frac{(15V + 5V - 3.5V)}{4A}
$$

$$
\approx 4.3\Omega
$$

The VOL value of 3.5V in the previous equation is a conservative value of VOL at the peak current of 4.0A (see Figure 6). At lower Rg values the voltage supplied by the ACNW3190 is not an ideal voltage step. This results in lower peak currents (more margin) than predicted by this analysis. When negative gate drive is not used, V_{EE} in the previous equation is equal to zero volts.

Step 2: Check the ACNW3190 Power Dissipation and Increase Rg if Necessary.

The ACNW3190 total power dissipation (PT) is equal to the sum of the emitter power (PE) and the output power (PO):

 $P_T = P_F + P_O$

 $P_E = I_F * V_F * Duty$ Cycle

 $P_O = P_{O(B|AS)} + P_{O(SWITCHING)}$

 $=$ Icc $*$ (Vcc - V_{FF}) + Esw(R_G, Q_G) $*$ f

For the circuit in Figure 26 with I_F (worst case) = 16 mA, $Rg = 4.3Ω$, Max Duty Cycle = 80%, Qg = 1000 nC, f = 15 kHz and T_A max = 85 \degree C:

 $P_E = 16$ mA $*$ 1.95V $*$ 0.8 = 25 mW

 $P_O = 3.25$ mA $*$ 20 V + 13 µJ $*$ 15 kHz

- $= 65$ mW + 195 mW
- $= 260$ mW
- $<$ 728 mW (P_{O(MAX)} @ 85°C = 800 mW-15C*4.8 mW/C)

The value of 3.25 mA for ICC in the previous equation was obtained by derating the ICC max of 5 mA to ICC max at 100C (see Figure 7).

The above computation shows that the power dissipation is within the specified limits. However, designers should verify that the thermal limits have not been violated by using the thermal model provided in this datasheet. This thermal model obtained based on JEDEC specification.

Under Voltage Lockout Feature

The ACNW3190 contains an under voltage lockout (UVLO) feature that is designed to protect the IGBT under fault conditions which cause the ACNW3190 supply voltage (equivalent to the fully-charged IGBT gate voltage) to drop below a level necessary to keep the IGBT in a low resistance state. When the ACNW3190 output is in the high state and the supply voltage drops below the ACNW3190 V_{UVLO-threshold} (9.5 < V_{UVLO-} < 12.0) the optocoupler output will go into the low state with a typical delay, UVLO Turn Off Delay, of 0.6 μs. When the ACNW3190 output is in the low state and the supply voltage rises above the ACNW3190 V_{UVLO+} threshold (11.0) V_{UVLO+} < 13.5) the optocoupler output will go into the high state (assumes LED is "ON") with a typical delay, UVLO Turn On Delay of 0.8 μs.

Figure 28. Under Voltage Lock Out

Thermal Model

Introduction

For application which requires an output gate current more than 2A, adequate PCB pad heat-sink must be provided to dissipate the power loss in the package. Failure to provide proper heat dissipation will potentially damage the gate drive after pro-long usage. This thermal model allows designer to compute the temperature of the LED and detector.

Defi nitions

θ1:Thermal impedance from LED junction to ambient

- θ2:Thermal impedance from LED to detector (output IC)
- θ3:Thermal impedance from detector (output IC) junction to ambient

Ambient Temperature: Measured approximately 1.25 cm above the optocoupler, with no forced air.

Description

This thermal model assumes that an 8-pin single-channel plastic package optocoupler is soldered into a 7.62 cm x 7.62 cm printed circuit board (PCB). The temperature at the LED and Detector junctions of the optocoupler can be calculated using the equations below.

 Δ TEA = A11*PE + A12*PD

 \triangle TDA = A21*PE + A22*PD

where,

 Δ TEA = Temperature difference between ambient and LED

 ΔTDA = Temperature difference between ambient and detector

- PE = Power dissipation from LED
- PD = Power dissipation from detector

A11, A12, A21, A22 thermal coefficients (units in °C/W) are functions of the thermal impedances θ1, θ2, θ3 (See Note 2).

