

High Speed, Low Power Monolithic Op Amp

FEATURES

725 MHz Gain Bandwidth – AD849 175 MHz Gain Bandwidth – AD848 4.8 mA Supply Current 300 V/μs Slew Rate 80 ns Settling Time to 0.1% for a 10 V Step – AD849 Differential Gain: AD848 = 0.07%, AD849 = 0.08% Differential Phase: AD848 = 0.08°, AD849 = 0.04° Drives Capacitive Loads

DC PERFORMANCE

3 nV/√Hz Input Voltage Noise – AD849 85 V/mV Open Loop Gain into a 1 kΩ Load – AD849 1 mV max Input Offset Voltage Performance Specified for ±5 V and ±15 V Operation Available in Plastic, Hermetic Cerdip and Small Outline Packages. Chips and MIL-STD-883B Parts Available. Available in Tape and Reel in Accordance with EIA-481A Standard

APPLICATIONS Cable Drivers 8- and 10-Bit Data Acquisition Systems Video and R_F Amplification Signal Generators

PRODUCT DESCRIPTION

The AD848 and AD849 are high speed, low power monolithic operational amplifiers. The AD848 is internally compensated so that it is stable for closed loop gains of 5 or greater. The AD849 is fully decompensated and is stable at gains greater than 24. The AD848 and AD849 achieve their combination of fast ac and good dc performance by utilizing Analog Devices' junction isolated complementary bipolar (CB) process. This process enables these op amps to achieve their high speed while only requiring 4.8 mA of current from the power supplies.

The AD848 and AD849 are members of Analog Devices' family of high speed op amps. This family includes, among others, the AD847 which is unity gain stable, with a gain bandwidth of 50 MHz. For more demanding applications, the AD840, AD841 and AD842 offer even greater precision and greater output current drive.

The AD848 and AD849 have good dc performance. When operating with ± 5 V supplies, they offer open loop gains of 13 V/mV (AD848 with a 500 Ω load) and low input offset voltage of 1 mV maximum. Common-mode rejection is a minimum of 92 dB. Output voltage swing is ± 3 V even into loads as low as 150 Ω .

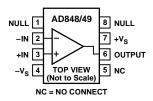
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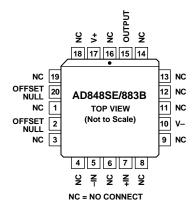
AD848/AD849

CONNECTION DIAGRAMS

Plastic (N), Small Outline (R) and Cerdip (Q) Packages



20-Terminal LCC Pinout



APPLICATIONS HIGHLIGHTS

- 1. The high slew rate and fast settling time of the AD848 and AD849 make them ideal for video instrumentation circuitry, low noise pre-amps and line drivers.
- 2. In order to meet the needs of both video and data acquisition applications, the AD848 and AD849 are optimized and tested for ± 5 V and ± 15 V power supply operation.
- 3. Both amplifiers offer full power bandwidth greater than 20 MHz (for 2 V p-p with ± 5 V supplies).
- 4. The AD848 and AD849 remain stable when driving any capacitive load.
- 5. Laser wafer trimming reduces the input offset voltage to 1 mV maximum on all grades, thus eliminating the need for external offset nulling in many applications.
- 6. The AD848 is an enhanced replacement for the LM6164 series and can function as a pin-for-pin replacement for many high speed amplifiers such as the HA2520/2/5 and EL2020 in applications where the gain is 5 or greater.

AD848/AD849—SPECIFICATIONS (@ T_A = +25°C, unless otherwise noted)

				AD848			D848 /		
Model	Conditions	Vs	Min	Тур	Max	Min	Тур	Max	Units
INPUT OFFSET VOLTAGE ¹		$\pm 5 V$		0.2	1		0.2	1	mV
	T _{MIN} to T _{MAX}	±15 V ±5 V		0.5	2.3 1.5		0.5	2.3 2	mV mV
	I MIN to I MAX	±15 V			3.0			3 .5	mV
Offset Drift		±5 V, ±15 V		7			7		µV/°C
INPUT BIAS CURRENT	T _{MIN} to T _{MAX}	±5 V, ±15 V ±5 V, ±15 V		3.3	6.6 7.2		3.3	6.6/5 7.5	μΑ μΑ
INPUT OFFSET CURRENT	T _{MIN} to T _{MAX}	±5 V, ±15 V ±5 V, ±15 V		50	300 400		50	300 400	nA nA
Offset Current Drift	I MIN TO I MAX	$\pm 5 V, \pm 15 V$ $\pm 5 V, \pm 15 V$		0.3	400		0.3	400	nA/°C
OPEN LOOP GAIN	$V_{O} = \pm 2.5 V$	±5 V							
	$R_{LOAD} = 500 \ \Omega$		9	13		9	13		V/mV
	T_{MIN} to T_{MAX}		7	0		7/5	0		V/mV
	$R_{LOAD} = 150 \Omega$	±15 V		8			8		V/mV
	$V_{OUT} = \pm 10 V$ $R_{LOAD} = 1 k\Omega$	±13 v	12	20		12	20		V/mV
	T_{MIN} to T_{MAX}		8	20		8/6	20		V/mV
DYNAMIC PERFORMANCE									
Gain Bandwidth	$A_{VCL} \ge 5$	$\pm 5 V$		125			125		MHz
Full Power Bandwidth ²	$V_{\rm O} = 2 V p-p,$	±15 V		175			175		MHz
	$R_{L} = 500 \Omega$ $V_{O} = 20 V p-p,$	±5 V		24			24		MHz
	$v_{\rm O} = 20 \text{ v } \text{p-p},$ $R_{\rm L} = 1 \text{ k}\Omega$	±15 V		4.7			4.7		MHz
Slew Rate	10L 1 100	±5 V		200			200		V/µs
	$R_{LOAD} = 1 k\Omega$	±15 V	225	300		225	300		V/µs
Settling Time to 0.1%	-2.5 V to +2.5 V	±5 V		65			65		ns
	10 V Step, $A_V = -4$	±15 V		100			100		ns
Phase Margin	$\begin{array}{l} C_{LOAD} = 10 \ pF \\ R_{LOAD} = 1 \ k\Omega \end{array}$	±15 V		60			60		Degree
DIFFERENTIAL GAIN	f = 4.4 MHz	±15 V		0.07			0.07		%
DIFFERENTIAL PHASE	f = 4.4 MHz	±15 V		0.08			0.08		Degree
COMMON-MODE REJECTION	$V_{CM} = \pm 2.5 V$	±5 V	92	105		92	105		dB
	$V_{CM} = \pm 12 V$	±15 V	92	105		92	105		dB
	T _{MIN} to T _{MAX}		88			88			dB
POWER SUPPLY REJECTION	$V_{S} = \pm 4.5 \text{ V to } \pm 18 \text{ V}$ T_{MIN} to T_{MAX}		85 80	98		85 80	98		dB dB
INPUT VOLTAGE NOISE	f = 10 kHz	±15 V		5			5		nV/√H
NPUT CURRENT NOISE	f = 10 kHz	±15 V		1.5			1.5		pA/√H
INPUT COMMON-MODE									1
VOLTAGE RANGE		±5 V		+4.3			+4.3		V
				-3.4			-3.4		V
		±15 V		+14.3			+14.3 -13.4		V V
OUTPUT VOLTAGE SWING	$R_{LOAD} = 500 \Omega$	±5 V	3.0	3.6		3.0	3.6		±V
	$R_{LOAD} = 300 \Omega$ $R_{LOAD} = 150 \Omega$	$\pm 5 V$ $\pm 5 V$	2.5	3		2.5	3		±V ±V
	$R_{LOAD} = 50 \Omega$	±5 V		1.4			1.4		±V
	$R_{LOAD} = 1 \ k\Omega$	±15 V	12			12			±V
	$R_{LOAD} = 500 \Omega$	±15 V	10			10			±V
SHORT CIRCUIT CURRENT		±15 V		32			32		mA
INPUT RESISTANCE				70			70		kΩ
INPUT CAPACITANCE				1.5			1.5		pF
OUTPUT RESISTANCE	Open Loop			15			15		Ω
POWER SUPPLY					. 10			. 10	V
Operating Range		+5 V	±4.5	4.8	±18	±4.5	10	±18 6.0	V m A
Quiescent Current	T _{MIN} to T _{MAX}	±5 V		4.ð	6.0 7.4		4.8	6.0 7.4/8.3	mA mA
	MIN CO MAX	±15 V		5.1	6.8		5.1	6.8	mA
	T _{MIN} to T _{MAX}	1	1		8.0	1		8.0/9.0	mA

NOTES ¹Input offset voltage specifications are guaranteed after 5 minutes at $T_A = +25^{\circ}C$. ²Full power bandwidth = slew rate/2 π V_{PEAK}. Refer to Figure 1. All min and max specifications are guaranteed. Specifications in **boldface** are tested on all production units at final electrical test. All others are guaranteed but not necessarily tested. Caracteristic methods to achieve rate in a state of the st Specifications subject to change without notice.

Model	Conditions	Vs	AD849J Min Typ Max	AD849A/S Min Typ Max	Units
INPUT OFFSET VOLTAGE ¹	Conditions	±5 V	0.3 1	0.1 0.75	mV
		±15 V	0.3 1	0.1 0.75	mV
	T_{MIN} to T_{MAX}	±5 V	1.3	1.0	mV
Offset Drift		±15 V ±5 V, ±15 V	1.3	1.0	mV µV/°C
INPUT BIAS CURRENT		$\pm 5 V, \pm 15 V$ $\pm 5 V, \pm 15 V$	3.3 6.6	3.3 6.6/5	μΑ
	$T_{\mbox{\scriptsize MIN}}$ to $T_{\mbox{\scriptsize MAX}}$	$\pm 5 V, \pm 15 V$ $\pm 5 V, \pm 15 V$	7.2	7.5	μA
INPUT OFFSET CURRENT		±5 V, ±15 V	50 300	50 300	nA
Offset Current Drift	T_{MIN} to T_{MAX}	±5 V, ±15 V ±5 V, ±15 V	400 0.3	400 0.3	nA nA/°C
OPEN LOOP GAIN	$V_{\rm O} = \pm 2.5 \rm V$	±5 V			
OPEN LOOP GAIN	$R_{LOAD} = 500 \Omega$	±0 V	30 50	30 50	V/mV
	T_{MIN} to T_{MAX}		20	20/15	V/mV
	$R_{LOAD} = 150 \Omega$		32	32	V/mV
	$V_{OUT} = \pm 10 V$	±15 V	45 85	45 85	V/mV
	$\begin{aligned} R_{LOAD} &= 1 \ k\Omega \\ T_{MIN} \ to \ T_{MAX} \end{aligned}$		45 85 30	45 85 30/25	V/mV
DYNAMIC PERFORMANCE					
Gain Bandwidth	$A_{\rm VCL} \geq 25$	±5 V	520	520	MHz
Full Power Bandwidth ²	$V_{O} = 2 V p - p,$	±15 V	725	725	MHz
run rower Bunamatin	$R_L = 500 \Omega$	±5 V	20	20	MHz
	$V_{\rm O} = 20 \text{ V p-p},$ $R_{\rm L} = 1 \text{ k}\Omega$	±15 V	4.7	4.7	MHz
Slew Rate	10L - 1 K22	$\pm 10^{\circ} V$ $\pm 5^{\circ} V$	200	200	V/µs
	$R_{LOAD} = 1 \ k\Omega$	±15 V	225 300	225 300	V/µs
Settling Time to 0.1%	-2.5 V to +2.5 V	±5 V	65	65	ns
	10 V Step, $A_V = -24$	±15 V	80	80	ns
Phase Margin	$C_{\text{LOAD}} = 10 \text{ pF}$ $R_{\text{LOAD}} = 1 \text{ k}\Omega$	±15 V	60	60	Degrees
DIFFERENTIAL GAIN	f = 4.4 MHz	±15 V	0.08	0.08	%
DIFFERENTIAL PHASE	f = 4.4 MHz	±15 V	0.04	0.04	Degrees
COMMON-MODE REJECTION	$V_{CM} = \pm 2.5 V$	±5 V	100 115	100 115	dB
	$V_{CM} = \pm 12 V$ T_{MIN} to T_{MAX}	±15 V	100 115 96	100 115 96	dB dB
POWER SUPPLY REJECTION	$V_{\rm S} = \pm 4.5 \text{ V to } \pm 18 \text{ V}$		98 120	98 120	dB
	T_{MIN} to T_{MAX}		94	94	dB
INPUT VOLTAGE NOISE	f = 10 kHz	±15 V	3	3	nV/√Hz
INPUT CURRENT NOISE	f = 10 kHz	±15 V	1.5	1.5	pA/√Hz
INPUT COMMON-MODE				1.0	
VOLTAGE RANGE		±5 V	+4.3 -3.4	+4.3 -3.4	
		±15 V	+14.3	+14.3	V
			-13.4	-13.4	V
OUTPUT VOLTAGE SWING	$R_{LOAD} = 500 \Omega$	±5 V	3.0 3.6	3.0 3.6	±V
	$R_{LOAD} = 150 \Omega$	±5 V	2.5 3	2.5 3	±V
	$R_{LOAD} = 50 \Omega$	$\pm 5 V$	1.4	1.4 12	$\pm V$
	$\begin{aligned} R_{\text{LOAD}} &= 1 \text{ k}\Omega \\ R_{\text{LOAD}} &= 500 \ \Omega \end{aligned}$	±15 V ±15 V	12 10	12	$\begin{array}{c} \pm V \\ \pm V \end{array}$
SHORT CIRCUIT CURRENT		±15 V	32	32	mA
INPUT RESISTANCE			25	25	kΩ
INPUT CAPACITANCE			1.5	1.5	pF
OUTPUT RESISTANCE	Open Loop		15	15	Ω
POWER SUPPLY					
Operating Range			±4.5 ±18	± 4.5 ± 18	V .
Quiescent Current	T _{MIN} to T _{MAX}	±5 V	4.8 6.0 7.4	4.8 6.0 7.4/8.3	mA mA
	I MIN CO I MAX	±15 V	5.1 6.8	5.1 6.8	mA
	T_{MIN} to T_{MAX}		8.0	8.0/9.0	mA

NOTES ¹Input offset voltage specifications are guaranteed after 5 minutes at $T_A = +25$ °C. ²Full power bandwidth = slew rate/2 π V_{PEAK}. Refer to Figure 1.

All min and max specifications are guaranteed. Specifications in **boldface** are tested on all production units at final electrical test. All others are guaranteed but not necessarily tested. Specifications subject to change without notice.

ABSOLUTE MAXIMUM RATINGS¹

Supply Voltage ±18 V
Internal Power Dissipation ²
Plastic (N) 1.1 Watts
Small Outline (R) 0.9 Watts
Cerdip (Q) 1.1 Watts
LCC (E) 0.8 Watts
Input Voltage $\pm V_S$
Differential Input Voltage ±6 V
Storage Temperature Range (Q) $\dots -65^{\circ}$ C to $+150^{\circ}$ C
(N, R) $-65^{\circ}C$ to $+125^{\circ}C$
Junction Temperature +175°C
Lead Temperature Range (Soldering 60 sec) +300°C

NOTES

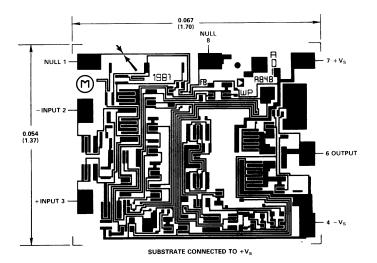
¹Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

²LCC: $\theta_{JA} = 150^{\circ}$ C/Watt Mini-DIP Package: $\theta_{JA} = 110^{\circ}$ C/Watt Cerdip Package: $\theta_{JA} = 110^{\circ}$ C/Watt

Small Outline Package: $\theta_{JA} = 155^{\circ}C/Watt.$

METALIZATION PHOTOGRAPH

Contact factory for latest dimensions. (AD848 and AD849 are identical except for the part number in the upper right.) Dimensions shown in inches and (mm).



ORDERING GUIDE

ORDERING GOIDE								
Model	Gain Bandwidth MHz	Min Stable Gain	Max Offset Voltage mV	Temperature Range – °C	Package Option ¹			
AD848JN	175	5	1	0 to +70	N-8			
AD848JR ²	175	5	1	0 to +70	R-8			
AD848JCHIPS	175	5	1	0 to +70	Die Form			
AD848AQ	175	5	1	-40 to +85	Q-8			
AD848SQ	175	5	1	-55 to +125	Q-8			
AD848SQ/883B	175	5	1	-55 to +125	Q-8			
AD848SE/883B	175	5	1	-55 to +125	E-20A			
AD849JN	725	25	1	0 to +70	N-8			
AD849JR ²	725	25	1	0 to +70	R-8			
AD849AQ	725	25	0.75	-40 to +85	Q-8			
AD849SQ	725	25	0.75	-55 to +125	Q-8			
AD849SQ/883B	725	25	0.75	-55 to +125	Q-8			
AD847J/A/S	50	1	1	See AD847 Data	a Sheet			

NOTES

¹E = LCC; N = Plastic DIP; Q = Cerdip; R = Small Outline IC (SOIC).

²Plastic SOIC (R) available in tape and reel. AD848 available in S grade chips. AD849 available in J and S grade chips.

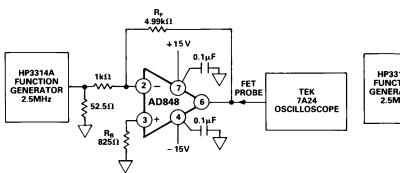


Figure 1. AD848 Inverting Amplifier Configuration

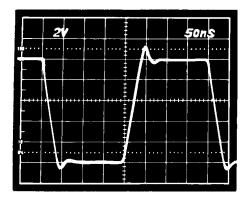


Figure 1a. AD848 Large Signal Pulse Response

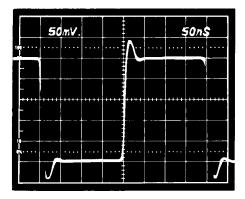


Figure 1b. AD848 Small Signal Pulse Response

OFFSET NULLING

The input voltage of the AD848 and AD849 are very low for high speed op amps, but if additional nulling is required, the circuit shown in Figure 3 can be used.

For high performance circuits it is recommended that a resistor (R_B in Figures 1 and 2) be used to reduce bias current errors by matching the impedance at each input. The offset voltage error caused by the input currents is decreased by more than an order of magnitude.

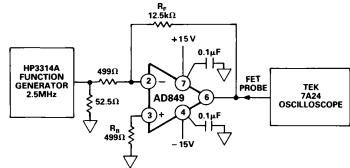


Figure 2. AD849 Inverting Amplifier Configuration

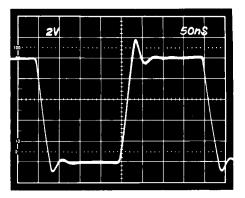


Figure 2a. AD849 Large Signal Pulse Response

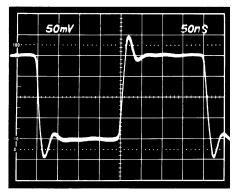


Figure 2b. AD849 Small Signal Pulse Response

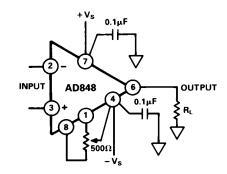


Figure 3. Offset Nulling

AD848/AD849–Typical Characteristics (@ $T_A = +25^{\circ}C$ and $V_S = \pm 15 V$, unless otherwise noted)

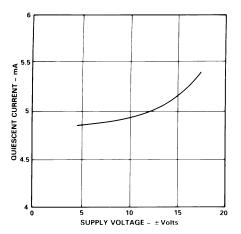


Figure 4. Quiescent Current vs. Supply Voltage (AD848 and AD849)

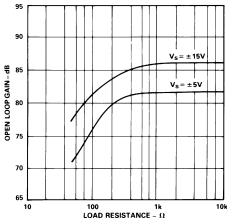


Figure 7. Open Loop Gain vs. Load Resistance (AD848)

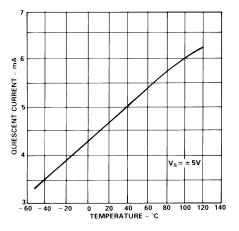


Figure 10. Quiescent Current vs. Temperature (AD848 and AD849)

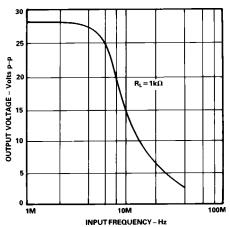


Figure 5. Large Signal Frequency Response (AD848 and AD849)

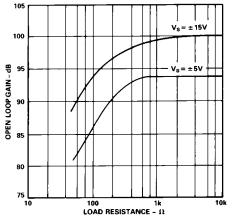


Figure 8. Open Loop Gain vs. Load Resistance (AD849)

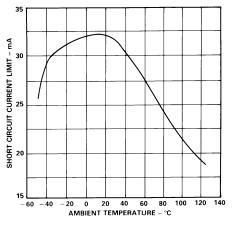


Figure 11. Short Circuit Current Limit vs. Temperature (AD848 and AD849)

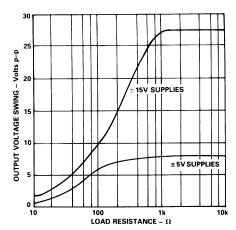


Figure 6. Output Voltage Swing vs. Load Resistance (AD848 and AD849)

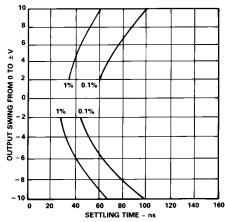


Figure 9. Output Swing and Error vs. Settling Time (AD848)

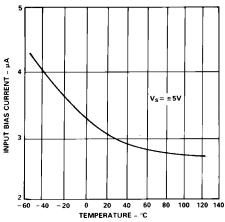
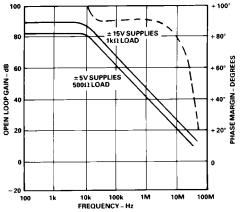


Figure 12. Input Bias Current vs. Temperature (AD848 and AD849)



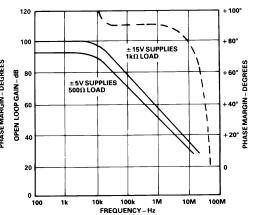


Figure 13. Open Loop Gain and Phase Margin vs. Frequency (AD848)

 $3V RMS R_L = 1k\Omega$

Figure 14. Open Loop Gain and Phase Margin vs. Frequency (AD849)

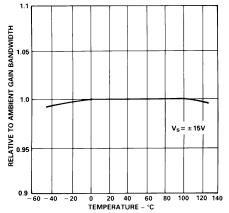
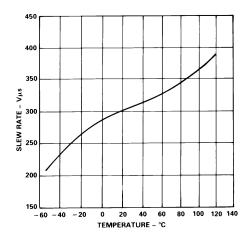


Figure 15. Normalized Gain Bandwidth Product vs. Temperature (AD848 and AD849)



2ND HARMONIC -110 -115 **3RD HARMONI** -120 100k 10k 100 1k FREQUENCY - Hz

Frequency (AD848)

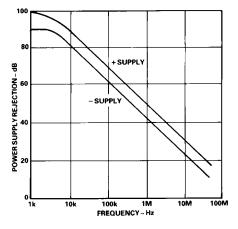


Figure 19. Power Supply Rejection vs. Frequency (AD848)

 $3V RMS R_L = 1k\Omega$ 95 뜅 HARMONIC DISTORTION -- 100 105 - 110 2ND HARMON - 115 RD HARMONIC - 120 100 10k 100 1k FREQUENCY - Hz

Figure 17. Harmonic Distortion vs. Frequency (AD849)

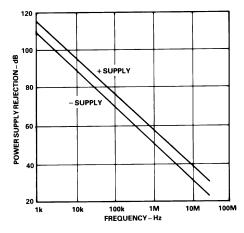


Figure 20. Power Supply Rejection vs. Frequency (AD849)

Figure 18. Slew Rate vs. Temperature (AD848 and AD849)

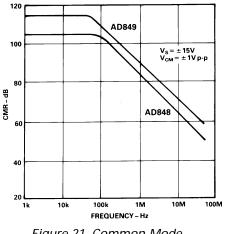
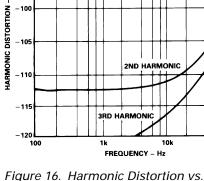


Figure 21. Common-Mode Rejection vs. Frequency



뜅