ANALOG 10 GHz to 40 GHz, 4-Channel Rx Mixer with
DEVICES 4× LO Multiplier/Filter

Data Sheet **[ADAR2004](https://www.analog.com/ADAR2004?doc=ADAR2004.pdf)**

FEATURES

Quad LNA, mixer, IF VGA 4× LO multiplier with programmable harmonic filter RF input frequency range: 10 GHz to 40 GHz IF output frequency range: 0 MHz to 800 MHz LO input frequency range: 2.4 GHz to 10.1 GHz Gain range: 21 dB to 41 dB Input P1dB: −20 dBm typical (at minimum gain) Noise figure: 9 dB typical (at maximum gain) 3-wire or 4-wire SPI control On-chip programmable state machines for fast multiplier/filter and receiver switching and control On-chip temperature sensor and ADC DC power: 910 mW (2.5 V supply)

7 mm × 7 [mm, 48-terminal LGA](#page--1-0) package

APPLICATIONS

Millimeter wave imaging Security Medical Industrial Multichannel receivers

GENERAL DESCRIPTION

The ADAR2004 is a 4-channel receiver IC that is optimized for millimeter wave body scanning applications. Accepting differential input signals from 10 GHz to 40 GHz, the ADAR2004 provides a low intermediate frequency (IF) output up to 800 MHz. Each receive channel also has independent gain control.

The mixer local oscillator (LO) path includes a 4× multiplier requiring an applied LO frequency between 2.4 GHz and 10.1 GHz. The 4× multiplier block includes a programmable filter that helps keep harmonics down before reaching the mixer.

Two programmable state machines are included to facilitate simple configuration, control, and fast switching of the frequency multiplier, filter, and receiver sections. These sequencers are programmed through the serial peripheral interface (SPI) and are then operated by pulsed inputs (reset and advance).

The ADAR2004 requires only a single 2.5 V supply with power consumption of 910 mW with all channels turned on.

The ADAR2004 is available in a 7×7 mm, [48-terminal LGA](#page--1-0) [package](#page--1-0) and is specified from −40°C to +85°C.

FUNCTIONAL BLOCK DIAGRAM

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REVISION HISTORY

8/2020-Revision 0: Initial Version

SPECIFICATIONS

V_{POS1}, V_{POS2}, V_{POS4} = 2.5 V, V_{POS3} = VREG, and T_A = -40°C to +85°C, unless otherwise noted.

Table 1.

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TIMING SPECIFICATIONS

 V_{POS1} , V_{POS2} , V_{POS4} = 2.5 V, V_{POS3} = VREG, and T_A = -40°C to +85°C, unless otherwise noted. See [Figure](#page-4-1) 2 to [Figure 4](#page-4-2) for the timing diagrams.

Table 2. SPI Timing

Timing Diagrams

Figure 4. SPI 4-Wire Read Timing Diagram

SPI Block Write Mode

Data can be written to the SPI registers using the block write mode where the register address automatically increments and data for consecutive registers can be written without sending new address bits. Data writing can be continued indefinitely until CS is raised, ending the transaction (see [Figure 5\)](#page-5-0)*.*

ABSOLUTE MAXIMUM RATINGS

Table 3.

¹ GND is the common ground to which all GNDx pins are connected.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Careful attention to PCB thermal design is required.

 θ_{JA} is the natural convection junction to ambient thermal resistance measured in a one cubic foot sealed enclosure. $θ$ _{JC} is the junction to case thermal resistance.

Table 4. Thermal Resistance

¹ Pad soldered.

ELECTROSTATIC DISCHARGE (ESD) RATINGS

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

Human body model (HBM) per ANSI/ESDA/JEDEC JS-001.

Charged device model (CDM) per ANSI/ESDA/JEDEC JS-002.

ESD Ratings for ADAR2004

Table 5. ADAR2004, 48-Terminal LGA

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

Figure 6. Pin Configuration (Top View, Not to Scale)

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Table 6. Pin Function Descriptions

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Figure 36. Receiver Sleep to Active Switching Time

Figure 37. RF Input Return Loss vs. RF Input Frequency

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Figure 41. ADC Code vs. Temperature

THEORY OF OPERATION **OVERVIEW**

The main elements of the ADAR2004 are a quad LNA, a mixer, an IF variable gain amplifier (VGA), a 4× LO frequency multiplier with integrated switchable harmonic filter, and a 1:4 signal splitter.

The four differential RF inputs are intended to be connected directly to a differential antenna structure, such as a dipole. Apply a single-tone LO input signal in the 2.4 GHz to 10.1 GHz frequency range with a power level of approximately −20 dBm to the LOIN port. This signal is frequency multiplied by 4 and filtered before driving the LO inputs of the four mixers.

The operation of these subcircuits can be controlled from the SPI port as well as two programmable state machines, one focused on LO multiplier/filter control and the other focused on quad receiver control.

The ADAR2004 also includes an Analog Devices, Inc., SPI port that is used for device configuration and readback. Although the state machines provide fast switching between states, all functions can also be controlled directly through the SPI port.

LO INPUT BUFFER, 4× MULTIPLIER, AND BAND-PASS FILTER

The LO input buffer provides approximately 17 dB of gain and provides an optimal drive signal to the 4× multiplier circuits for LO input power levels down to −25 dBm. The bias levels of the input and output stages of the input buffer are independently adjustable through the SPI via Register 0x013. See the [Bias Points](#page-23-0) [and Voltages](#page-23-0) section for more information.

The broadband frequency multiplier consists of three parallel subcircuits. Each subcircuit (low band, mid band, high band) is optimized to multiply and filter a segment of the total frequency range (2.4 GHz to 10.1 GHz input, 9.6 GHz to 40.4 GHz output). Recommended ranges and register settings for each band are

shown in [Table 7.](#page-15-4) SP3T switches at the input and output of the multiplier block are used to select the desired subcircuit.

Each subcircuit consists of a 4× multiplier and an adjustable band-pass filter (BPF). The bias levels of the 4× multipliers are adjustable through the SPI via Register 0x011 and Register 0x012. See the Bias Points [and Voltages](#page-23-0) section for more information.

When the LO input frequency is in the low end of the subcircuit band, the BPF corner frequency must be set to its low state. Set the associated bit high to set the BPF corner frequency to its low state. See [Table 7.](#page-15-4)

To complete a full 9.6 GHz to 40.4 GHz frequency sweep, adjust the multiplier/filter block settings six times to ensure optimum harmonic rejection and output power. These six settings are shown in [Table 7.](#page-15-4)

In addition to having sleep and active modes, the 4× multipliers can be set to ready mode. Ready mode is a hybrid state between sleep and active mode. Current consumption in ready mode is higher than sleep mode but lower than active mode. The switching time between ready mode and active mode is significantly faster than from sleep mode to active mode.

1:4 SIGNAL SPLITTER NETWORK

The output of the multiplier/filter block is then applied to a 1:4 active power splitting network that is composed of two stages. The first stage is a 1:2 active splitter, which then feeds the second stage, two 1:2 active splitters. Each output path from the second stage drives a single downconverting mixer, which results in a single input signal being split into four independently controlled channels. The bias levels of each splitter stage are adjustable through the SPI via Register 0x014. See the [Bias Points](#page-23-0) and [Voltages](#page-23-0) section for more information.

Table 7. Multiplier/Filter Settings for Optimal LO Harmonic Rejection

¹ MULT_x refers to the MULT_EN_MODE_x and MULT_SPI registers.

RECEIVERS

There are four independent receive channels, each with fully differential inputs and outputs. Each channel includes an RF LNA front end, a downconverting mixer, and a dedicated IF VGA. The inputs operate from 10 GHz to 40 GHz and are intended to be connected to dipole antennas, whereas the outputs operate from low frequency to 800 MHz and are meant to be directly connected to an ADC. The bias levels of the LNAs, mixers, and IF VGAs are adjustable through the SPI. There is one setting for all the LNAs (LNA_BIAS in Register 0x015), one setting for the mixers (MIX_BIAS in Register 0x015), and one setting for the IF VGAs (IFAMP_BIAS in Register 0x016). The IF VGAs also have a setting for V_{OCM} (IFAMP_CM in Register 0x017). See the Bias Points [and Voltages](#page-23-0) section for more information.

Although normal operation envisages all four receive channels operating at one time, the programmability allows any combination of receive channels to be turned on or off simultaneously.

TEMPERATURE SENSOR

The ADAR2004 has an on-chip temperature sensor that feeds into a dedicated 8-bit ADC for monitoring the temperature of the chip. Use the following equation to calculate the approximate temperature in Celsius from the ADC output:

TA = (1.05 × *ADC_OUTPUT*) – 80

where *ADC_OUTPUT* is the ADC output word in Register 0x031.

ADC AND ADC CLOCK

The ADAR2004 has an on-chip, 8-bit ADC and a variable clock input, each with their own enable control bits.

To take a measurement from the ADC, first write to the ADC_CTRL register, Register 0x030. This register contains the following bits:

- Bit 0: ADC_EOC (read only). This bit is a flag for when the ADC conversion is complete.
- Bit 4: ST_CONV (read/write). This bit is set to start an ADC conversion cycle.
- Bit 5: CLK_EN (read/write). This bit enables the ADC clock.
- Bit 6: ADC_EN (read/write). This bit enables the ADC.
- Bit 7: ADC_CLKFREQ_SEL (read/write). This bit is used to set the clock frequency. A low sets the clock to 2 MHz, whereas a high sets the clock to 250 kHz.

After the ADC_CTRL register is written, it must be polled to wait for the ADC_EOC bit to go high. When the ADC_EOC bit goes high, the measured value can be read out from the ADC_OUTPUT register, Register 0x031.

APPLICATIONS INFORMATION

SPI CONTROL

The ADAR2004 is designed to operate as part of a larger array. The built in state machines help to ease the control of many chips in parallel and to ensure that the fastest switching speeds are achieved. However, it is possible to operate every aspect of the ADAR2004 using the SPI port alone. When the state machines are disabled by setting MULT_SEQ_EN (Register 0x019, Bit 7) and RX_SEQ_EN (Register 0x018, Bit 7) low, the multiplier/filter and receiver blocks respond to the SPI controlled registers (Register 0x02B to Register 0x02F), rather than stepping through the programmed states.

MULT_SPI (Register 0x02F) controls the multiplier/filter block when in SPI mode and has all the same controls as a typical multiplier/filter mode.

Register 0x02B to Register 0x02E control the receiver block when in SPI mode and have all the same controls as a typical receiver mode, as well as individual enables for the LNAs, mixers, and IF VGAs. Note that when the ADAR2004 receiver block is in SPI mode, the channel enables do not turn on the desired channel unless each piece of the receiver signal chain (LNA, mixer, IF VGA) is enabled as well, which contrasts with the sequencer modes, where a channel enable turns on the entire channel. The sequencer does not have access to the individual enables.

Operating the ADAR2004 in this manner can be thought of as a manual, rather than an automatic, approach. With the sequencers disabled, any changes to the configuration of the chip must be made through an SPI write, because pulsing any of the sequencer control pins has no effect.

STATE MACHINE MODES vs. STATES

Both the multiplier/filter state machine and the receiver state machine have 16 modes available to set the configuration of their respective subcircuitry. The sequencers also have 16 states available to cycle through.

Within each mode of the multiplier/filter state machine, the user can define the following:

- The enabled status of the RF input buffer (on or off, one bit).
- The sleep, ready, or active state of each $4\times$ multiplier band (two bits for each band, six bits in total). The two bits control the ready and active status, and if neither bit is high, the multiplier band is set to sleep. Both bits must be high to be fully active.
- The BPF status (on or off, 1 bit for all bands).

Within each mode of the receiver state machine, the user can define the following:

The enabled status of each receive channel (one bit for each band, four bits in total). Each bit enables the entire channel, including the respective LNA, mixer, and IF VGA.

- The enabled status of the first 1:2 signal splitter (on or off, one bit).
- The enabled status of the 1:2 signal splitter feeding the mixers on Channel 1 and Channel 2 (on or off, one bit).
- The enabled status of the 1:2 signal splitter feeding the mixers on Channel 3 and Channel 4 (on or off, one bit).
- The gain of each channel (four bits for each, 16 total).

Each multiplier/filter state is used to select an operating mode. Each state bit field contains four bits, allowing selection of any mode between 0 and 15 (Register 0x070 to Register 0x07F). There are 16 multiplier/filter states available (Register 0x022 to Register 0x029). When the multiplier/filter state machine is enabled and the sequencer depth is set, the multiplier/filter state machine cycles through the states in order, up to the defined state machine depth.

Similarly, each receiver state is used to select an operating mode. Each state bit field has four bits, allowing the selection of any mode between 0 and 15 (Register 0x040 to Register 0x06F). There are 16 receiver states available (Register 0x01A to Register 0x021). When the receiver state machine is enabled and the sequencer depth is set, the receiver state machine cycles through the states in order, up to the defined state machine depth.

[Figure](#page-17-4) 42 shows how the state machine pointer moves through a loop. In this diagram, n is the total number of states inside the loop. Because the sequencer depth bit field is 0 indexed, n is equal to one more than the value in the sequencer depth bit field.

n = *MULT_STATES* + 1

where: $n = 1$ to 16.

MULT_STATES is the multiplier sequencer depth.

n = *RX_STATES* + 1

where:

 $n = 1$ to 16.

RX_STATES is the receiver sequence depth.

Figure 42. State Machine Position Loop

STATE MACHINE SETUP

Both state machines in the ADAR2004 have configuration registers that control various aspects of the state machine.

For the multiplier/filter sequencer, this register is Register 0x019 and contains the following bits:

Bits 0 to Bit 3: MULT_STATES. These bits set the number of states in the loop (se[e Figure](#page-17-4) 42).

- Bit 4: MULT_CTL_LATCH_BYP. This bit bypasses the latch on MADV and MRST. If the latch is enabled, the next state is loaded up on the rising edge of a MRST or MADV pulse. The state is then latched to the device on the falling edge of the same pulse. If the latch is bypassed, everything is applied as soon as possible after the rising edge of the pulse, with no latching.
- Bit 5: MULT_SLP_HOLD. This bit prevents the multiplier/filter block from advancing when forced into a sleep state by the receiver block. This bit is used in conjunction with RX_SLP_CTRL (Register 0x018, Bit 6). See the [Sequencer Sleep Hold](#page-20-0) section for more information.
- Bit 6: MULT_SLP_CTRL. This bit forces the multiplier/filter block to sleep whenever the receiver block is sleeping. See th[e Sequencer Sleep Control](#page-19-1) section for more information.
- Bit 7: MULT_SEQ_EN. This bit enables the multiplier/filter block. This bit must be set high for the block to operate with the external pins.

For the receiver sequencer, the control register is Register 0x018 and contains the following bits:

- Bit 0 to Bit 3: RX_STATES. These bits set the number of states in the loop (se[e Figure](#page-17-4) 42).
- Bit 4: RX_CTL_LATCH_BYP. This bit bypasses the latch on RxADV and RxRST. If the latch is enabled, the next state is loaded up on the rising edge of an RxRST or RxADV pulse. The state is then latched to the device on the falling edge of the same pulse. If the latch is bypassed, everything is applied as soon as possible after the rising edge of the pulse, with no latching.
- Bit 5: RX_SLP_HOLD. This bit prevents the receiver block from advancing when forced into a sleep state by the multiplier/filter block. This bit is used in conjunction with MULT_SLP_CTRL (Register 0x019, Bit 6). See the Sequencer [Sleep Hold](#page-20-0) section for more information.
- Bit 6: RX_SLP_CTRL. This bit forces the receiver block to sleep whenever the multiplier/filter block is sleeping. See the [Sequencer Sleep Control](#page-19-1) section for more information.
- Bit 7: RX_SEQ_EN. This bit enables the receiver block. This bit must be set high for the block to operate with the external pins.

MULTIPLIER/FILTER STATE MACHINE

A programmable state machine provides a convenient and fast control mechanism for the multiplier/filter block and avoids the need for SPI writes each time the block must be reconfigured.

To enable the state machine, set the MULT_SEQ_EN bit (Register 0x019, Bit 7) high.

Although only six multiplier/filter modes are required for a complete 9.6 GHz to 40.4 GHz sweep, as described i[n Table 7,](#page-15-4) a maximum state machine depth of 16 is provided for optimum flexibility.

Eight default modes can be assigned to any of the 16 states. These eight modes consist of a sleep mode, a ready mode, and the six modes required to complete a 9.6 GHz to 40.4 GHz sweep, as described in [Table 7.](#page-15-4) It is possible to overwrite any of the multiplier/filter modes with a custom set of operating conditions by changing the bits in Register 0x070 to Register 0x07F.

After the modes are defined, set the order in which the sequencer moves through the desired modes by filling the state bits in Register 0x022 to Register 0x029 in order with the modes of interest. Any state can point to any mode, except State 0, which always points to Mode 0. Note that the sequencer moves through the states in order, up to the state machine depth.

Finally, the user must define how many states are used by setting the state machine depth (MULT_STATES, Register 0x019, Bits[3:0]). MULT_STATES is 0 indexed. Therefore, setting the depth to 0 leaves MULT_STATE_1 (Register 0x022, Bits[7:4]) as the only state in the loop.

After the multiplier/filter state machine is programmed and enabled, operation is controlled by the MRST and MADV pins. Alternatively, operation can be controlled through the SPI using the MULT_RST_SPI and MULT_ADV_SPI bits (Register 0x02A, Bit 3 and Bit 2, respectively). Note that using the SPI is slower than pulsing the sequencer pins directly.

MRST moves the pointer on the multiplier/filter state machine to State 0 regardless of the current position of the pointer and can be asserted at any time. State 0 always refers to Mode 0 and cannot be set to another mode. However, Mode 0 can be overwritten with any multiplier/filter configuration. Mode 0 is defined in Register 0x070.

MADV pulses advance the multiplier/filter state machine pointer one state at a time until the defined sequencer depth is cycled through. At that point, an additional MADV pulse moves the pointer back to State 1, which is normally set to a ready mode (however, State 1 can be set to any mode). State 1 applies the mode defined in the MULT_STATE_1 bits (Register 0x022, Bits[7:4]).

RECEIVER STATE MACHINE

Like the multiplier/filter state machine, the receiver state machine can be used to quickly cycle through receiver states without using the comparatively slower SPI.

To enable the state machine, set the RX_SEQ_EN bit (Register 0x018, Bit 7) high.

The receiver state machine controls the status of the four receive channels (each with an LNA, mixer, and IF VGA) and the status of the 1:4 splitter network by defining the desired modes of operation in Register 0x040 to Register 0x06F. Each mode outlines a custom set of operating conditions.

Although only one active state is required to enable all receive channels, a state machine depth of 16 is provided for optimum flexibility of the receiver sequencer and to lower the total number of control lines required to operate multiple ADAR2004 chips in parallel. It is possible to control up to 16 ADAR2004 ICs using the

same four sequencer lines (MADV, MRST, RxADV, RxRST). See the [Parallel Chip Control](#page-21-0) section for more information.

Following the mode definitions, the user must fill the state bits in Register 0x01A to Register 0x021 with the modes of interest. Any state can point to any mode, except State 0, which always points to Mode 0. Note that the sequencer moves through the states in order, up to the state machine depth.

After defining the states, the user must set the number of states to be used by the sequencer by changing the RX_STATES bits (Register 0x018, Bits[3:0]). RX_STATES is 0 indexed. Therefore, setting the depth to 0 leaves RX_STATE_1 as the only state in the loop.

After the multiplier/filter state machine is programmed and enabled, operation is controlled by the RxRST and RxADV pins. Alternatively, operation can be controlled through the SPI using the RX_RST_SPI and RX_ADV_SPI bits (Register 0x02A, Bits[1:0]). Note that using the SPI is slower than pulsing the sequencer pins directly.

RxRST moves the pointer on the receiver state machine to State 0 regardless of the current position of the pointer and can be asserted at any time. State 0 always refers to Mode 0 and cannot be set to another mode. However, Mode 0 can be overwritten with any receiver configuration. Mode 0 is defined in Register 0x040, Register 0x041, and Register 0x042.

RxADV pulses advance the receiver state machine pointer one state at a time until the defined sequencer depth is cycled through. At that point, an additional RxADV pulse moves the pointer back to State 1. State 1 applies the mode defined in the RX_STATE_1 bits (Register 0x01A, Bits[7:4]).

FREQUENCY SWEEP ALL CHANNELS

[Figure 43](#page-20-2) shows a method of operation that can be used during a 20 GHz to 40 GHz frequency sweep of all receive channels. As described in [Table 7,](#page-15-4) three multiplier/filter states are required during a 20 GHz to 40 GHz sweep. In this example, the defined state machine depth, MULT_STATES (Register 0x019, Bits[3:0]), is 3 because there are four states inside the loop, and MULT_STATES is 0 indexed.

As shown i[n Figure 43,](#page-20-2)

- Multiplier/Filter State $0 = \text{sleep}$ (outside the loop)
- Multiplier/Filter State 1 = mid band multiplier ready
- Multiplier/Filter State 2 = output 20 GHz to 25 GHz to mixers
- Multiplier/Filter State $3 =$ output 25 GHz to 30 GHz to mixers
- Multiplier/Filter State 4 = output 30 GHz to 40 GHz to mixers

The initial state is the sleep state where power consumption is at a minimum. A pulse on MADV advances the state machine to the second state, which is defined as a ready state. By partially powering up the mid band multiplier with its BPF set high, an additional pulse on MADV makes this subcircuit path active in under 10 ns. By making use of the ready mode throughout the sweep, the settling time can be kept under 10 ns.

After the appropriate number of pulses is applied to MADV (4, in this case), the state machine automatically returns to the second state (ready mode).

SEQUENCER SLEEP CONTROL

To further simplify the control of the ADAR2004, it is possible to link the sleep states of the two state machines so that one sequencer going to sleep forces the other sequencer to sleep as well. This link helps to limit the total number of required states to achieve the desired type of operation. To use this feature, one of the two sleep control bits must be set, but not both.

For example, when the ADAR2004 is configured for a frequency sweep (as shown i[n Figure 43\)](#page-20-2), if the RX_SLP_CTRL bit (Register 0x018, Bit 6) is set, when the multiplier/filter sequencer is reset, the receiver state machine is forced to sleep as well. This means that the receiver state machine does not require a state dedicated to sleep if it only needs to sleep when the multiplier/filter sleeps. Furthermore, because the multiplier/filter sleep state is controlling the sleep state of the receiver, bringing the multiplier/filter out of sleep also brings the receiver out of sleep. This routine is controlled with either the SPI or one external line (MADV).

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Figure 43. Multiplier State Machine Operating Example for aFrequency Sweep of All ReceiverChannels Simultaneously From 20GHz to 40GHz (N/AMeans Not Applicable)

SEQUENCER SLEEP HOLD

By default, when one of the sequencers is forced asleep using one of the sleep control bits (MULT_SLP_CTRL or RX_SLP_CTRL), the counter for the sequencer being controlled can still be advanced. Because of this behavior, it is possible for a state machine to be put to sleep in one condition and brought out of sleep in another, depending on whether the sequencer advance or reset signals were exercised while the sequencer was sleeping.

If this behavior is undesired, the sleep hold bits (MULT_SLP_HOLD and RX_SLP_HOLD) can be asserted to force the associated state machine counter to ignore any inputs on the sequencer advance line. The counter also ignores advance signals coming from the SPI.

Note that the counter always responds to a reset signal, even when the sleep hold bit is high.

When sleep hold is used, take care when bringing the state machines out of sleep mode to ensure that the desired modes are reached. If the advance pins for both sequencers are pulsed too closely together under this condition, it is possible for the sequencer being controlled to not move into the expected state. To prevent this issue, stagger the advance pulses so that the rising edges are separated by a minimum of 3 ns with the pulse of the controlled sequencer coming second. Se[e Figure 44](#page-20-3) for an example of how to pulse the sequencers under this condition.

Figure 44. Example of How to Pulse the Sequencer Advance Pinsto Ensure Advancement With Receiver State Machine Sleep Hold Enabled

SEQUENCER CONTROL LATCH BYPASS

Typically, when a sequencer control line is pulsed, the upcoming state is loaded on the rising edge of the control pulse and latched to the various signal blocks on the falling edge of the same pulse. The latching helps to line up all the internal control signals so that the changes take place simultaneously.

It is possible to bypass the latching of the internal control signals by setting the bypass bits (RX_CTL_LATCH_BYP and MULT_CTL_LATCH_BYP) in the sequencer setup registers (Register 0x018, Bit 4 and Register 0x019, Bit 4).

Bypassing the latch results in the new state taking effect as soon as possible after the rising edge. Because the internal control signals are not aligned, it is possible that the overall switching time between states increases when compared to using the latch. Also, glitches are more likely to occur in the internal control signals, resulting in undesired transients in the RF blocks.

Note that this latch is the last check before any new data is sent to the various individual blocks. Therefore, when using the ADAR2004 in manual or SPI mode (sequencers disabled), the latching must be bypassed. If the latching is not bypassed, the blocks do not receive the new instructions unless the external sequencer pins are pulsed. However, this issue is uncommon because the sequencers are disabled in this mode of operation.

PARALLEL CHIP CONTROL

Up to 16 devices (a total of 64 channels) can be driven by a single set of four state machine control lines, three common SPI lines, and a CS line for each chip. Using this method, the total number of digital control lines is $7 + N$, where N is the number of ADAR2004 ICs(se[e Figure](#page-21-2) 45 for a basic diagram). Parallel chip control can be used to minimize the total number of digital control lines. The SPI lines can be reduced to two common lines if 3-wire mode is selected by setting the SDOACTIVE and SDOACTIVE_ bits (Register 0x000, Bit 4 and Bit 3, respectively) low. If 3-wire SPI mode is used, the total number of digital lines is $6 + N$.

Figure 45. SPI and State Machine Digital Lines For Addressing and Controlling Up to 16 ADAR2004 Devices

MULTICHIP FREQUENCY SWEEP

[Figure](#page-22-0) 46 shows an example of how the two state machines can be used to complete a multichip frequency sweep from 10 GHz to 16 GHz (that is, receive on all four channels on a single chip while at a fixed frequency range, move to the next chip and receive on all channels, moving through 16 chips in total, and then move to the next frequency and repeat the process). This example assumes that the state machine control lines are connected in parallel for up to 16 devices (64 channels, see [Figure](#page-21-2) 45).

Initially, pulses on RxRST and MRST put both state machines in State 0, which in this case, is a sleep mode.

Next, pulses on MADV and RxADV advance both state machines to their first active state (receiving on all channels of ADAR2004 IC 1).

After ADAR2004 IC 1 receives the signal, an additional pulse on RxADV activates all channels on ADAR2004 IC 2 while putting the multiplier/filter of the first chip into a ready mode and the receivers of that chip into a sleep mode to prevent disrupting the multiplier/filter signal before the receiver turns off. This sequence continues until all 16 ADAR2004 devices receive at the first frequency or range.

At that point, a pulse is applied to both MADV and RxADV to advance the multiplier/filter to the next frequency range of interest and set the ADAR2004 IC 1 back into an active mode. Another series of RxADV pulses follow until ADAR2004 IC 16 is receiving the new frequency range.

[Table](#page-22-1) 8 describes how the receiver state machine for each ADAR2004 can be set up to work in sequence. Each device is turned fully on for only one state, but these states are all offset from each other. To run this sequence, where up to 16 devices are swept with all state machines driven in parallel, 16 receive states are used inside the loop, with the sleep state (State 0) used as a reset condition.

If there were more tiles of 16 chips in the array that need to receive after the tile described in [Table](#page-22-1) 8, this tile can have a reset pulse sent to put the sequencers into the initial sleep mode to wait for their turn to receive again.

Figure 46. State Machine Loop Example For a 16-Chip Frequency Sweep (N/A Means Not Applicable)

¹ SLP is the sleep state (State 0)

BIAS POINTS AND VOLTAGES

Table 9. Default Bias Points

REGISTER SUMMARY

Table 10. ADAR2004 Register Summary

Address	Name	Bits	Bit Name	Description		Reset Access
0x000	INTERFACE_CONFIG_A	7	SOFTRESET	Soft Reset	0x0	R/W
		6	LSB_FIRST	LSB First	0x0	R/W
		5	ADDR_ASCN	Address Ascension	0x0	R/W
		4	SDOACTIVE	SDO Active	0x1	R/W
		3	SDOACTIVE_	SDO Active	0x1	R/W
		2	ADDR_ASCN_	Address Ascension	0x0	R/W
		1	LSB_FIRST_	LSB First	0x0	R/W
		0	SOFTRESET_	Soft Reset	0x0	R/W
0x001	INTERFACE_CONFIG_B	7	SINGLE_INSTRUCTION	Single Instruction	0x0	R/W
		6	$\overline{\text{CS}}$ STALL	\overline{CS} Stall	0x0	R/W
		5	MASTER_SLAVE_RB	Master Slave Readback	0x0	R/W
		4	SLOW_INTERFACE_CTRL	Slow Interface Control	0x0	R/W
		3	RESERVED	Reserved	0x0	R
			[2:1] SOFT_RESET	Soft Reset	0x0	R/W
		0	RESERVED	Reserved	0x0	R
0x002	DEV_CONFIG		[7:4] DEV_STATUS	Device Status	0x1	R/W
			[3:2] CUST_OPERATING_MODE	Custom Operating Modes	0x0	R/W
			[1:0] NORM_OPERATING_MODE	Normal Operating Modes	0x0	R/W
0x003	CHIP_TYPE		$[7:0]$ CHIP_TYPE	Chip Type	0x0	R
0x004	PRODUCT_ID_H		$[7:0]$ PRODUCT_ID[15:8]	Product ID High	0x0	R
0x005	PRODUCT_ID_L		$[7:0]$ PRODUCT_ID[7:0]	Product ID Low	0x0	R
0x00A	SCRATCH_PAD		[7:0] SCRATCHPAD	Scratch Pad	0x0	R/W
0x00B	SPI_REV		$[7:0]$ SPI_REV	SPI Revision	0x0	R
0x00C	VENDOR_ID_H		$[7:0]$ VENDOR_ID[15:8]	Vendor ID High	0x0	R
0x00D	VENDOR_ID_L		[7:0] VENDOR_ID[7:0]	Vendor ID Low	0x0	R
0x00F	TRANSFER_REG		$[7:1]$ RESERVED	Reserved	0x0	R
		0	MASTER_SLAVE_XFER	Master Slave Transfer	0x0	R/W
0x010	PWRON		$[7:1]$ RESERVED	Reserved	0x0	R
		0	PWRON	Chip Power-Up	0x1	R/W
0x011	BIAS_CURRENT_MULT1		[7:4] MULT_MID_BIAS	Mid Band 4x Bias Current Setting	0x5	R/W
			[3:0] MULT_LOW_BIAS	Low Band 4x Bias Current Setting	0x5	R/W
0x012	BIAS_CURRENT_MULT2		$[7:4]$ RESERVED	Reserved	0x0	R/W
		[3:0]	MULT_HIGH_BIAS	High Band 4x Bias Current Setting	0x7	R/W
0x013	BIAS_CURRENT_LOAMP		$[7:4]$ LO_AMP2_BIAS	LO Amp Output Stage Bias Current Setting	0x7	R/W
			$[3:0]$ LO_AMP1_BIAS	LO Amp Input Stage Bias Current Setting	0x8	R/W
0x014	BIAS_CURRENT_SPLT		$[7:4]$ SPLT2_BIAS	Second Active Splitter Stages Bias Current Setting	0x7	R/W
			$[3:0]$ SPLT1_BIAS	First Active Splitter Stage Bias Current Setting	0xA	R/W
0x015	BIAS_CURRENT_LNAMIX		$[7:4]$ MIX_BIAS	Mixer Bias Current Setting	0x2	R/W
			$[3:0]$ LNA_BIAS	LNA Bias Current Setting	0xA	R/W
0x016	BIAS_CURRENT_IFAMP		$[7:4]$ IFAMP_BIAS	IF Output Amp Bias Current Setting	0xC	R/W
			$[3:0]$ RESERVED	Reserved	0x0	R/W
0x017	IFAMP_CM		$[7:4]$ RESERVED	Reserved	0x0	R
		[3:0]	IFAMP_CM	IF Output Amp Common-Mode Voltage Setting	0x4	R/W
0x018	RX_SEQUENCER_SETUP	7	RX_SEQ_EN	Enables Receiver Sequencer	0x0	R/W
		6	RX_SLP_CTRL	Sets Receiver Sleep Mode Control	0x0	R/W
		5	RX_SLP_HOLD	Holds the Receiver Sequencer State When Multiplier Is in Sleep Mode	0x0	R/W
		4	RX_CTL_LATCH_BYP	Bypasses the Control Latch for Receiver Controls	0x1	R/W
			$[3:0]$ RX_STATES	Sets Number of Receiver Sequencer States	0x0	R/W

