

ATS627LSG

True Zero Speed, Low Jitter, High Accuracy Position Sensor IC

FEATURES AND BENEFITS

- Highly accurate in presence of:
	- \Box Anomalous target geometry (tooth-tooth variation)
	- \Box Signature teeth or valleys
	- □ Target runout
- Highly repeatable output edges (low jitter)
- True zero-speed operation
- Undervoltage lockout
- Air gap independent switch points
- Defined power-on state
- High operating temperature
- Single-chip sensing IC for high reliability
- Enhanced quality through Scan Path and IDDQ measurement
- Enhanced EMC performance

PACKAGE: 4-pin SIP (suffix SG)

DESCRIPTION

The ATS627 is a true zero-speed gear tooth sensor IC consisting of an optimized Hall IC and rare earth pellet configuration in a single overmolded package. The integrated circuit provides a manufacturer-friendly solution for digital gear tooth sensing applications. This small package can be easily assembled and used in conjunction with gears of various shapes and sizes.

The dual-element Hall IC switches in response to differential magnetic signals created by a ferrous target. Digital processing of the analog signal provides zero-speed performance independent of air gap as well as dynamic adaptation of device performance to the typical operating conditions found in automotive applications.

High-resolution peak detecting DACs are used to set the adaptive switching thresholds of the device. Bounded tracking and switchpoint hysteresis reduce the negative effects of any anomalies in the magnetic signal associated with the targets used in many automotive applications. This sensor IC system is optimized for engine crank applications that use targets possessing signature regions.

This device is available in a lead (Pb) free 4-pin SIP package (SG) with a 100% matte tin-plated leadframe.

Functional Block Diagram

SELECTION GUIDE

*Contact Allegro™ for additional packing options

ABSOLUTE MAXIMUM RATINGS

Pinout Diagram

Terminal List Table

OPERATING CHARACTERISTICS: Valid through full operating supply voltage and ambient temperature ranges, using Reference Target 60+2; unless otherwise specified

Continued on the next page…

OPERATING CHARACTERISTICS (continued): Valid through full operating supply voltage and ambient temperature ranges, using Reference Target 60+2; unless otherwise specified

[1] Typical values are at $T_A = 25^{\circ}$ C and $V_{CC} = 12$ V.

[2] Maximum voltage must be adjusted for power dissipation and junction temperature; see Power Derating section.

[3] Negative current is defined as current coming out of (sourced from) the specified device terminal.

[4] Sustained voltages beyond the clamp voltage may cause permanent damage to the IC.

 $[5]$ 1 G (gauss) = 0.1 mT (millitesla).

[6] The repeatability specification is based on statistical evaluation of a sample population, evaluated at 1000 Hz.

[7] Power-on frequencies ≤ 200 Hz. Higher power-on frequencies may result in a delay of full output accuracy or undetected target edges. [8] Excludes effects caused by signature region.

Figure 1. Definition of Output Delay Time, $t_{d(OUT)}$, Output Fall Time, $t_{f(OUT)}$, and Output Rise Time, $t_{r(OUT)}$.

Figure 2. Definition of T_{TARGET}

Supply Current (Off) versus Supply Voltage

Output Voltage (On) versus Output Current

Supply Current (On) versus Temperature

Supply Current (Off) versus Temperature

Output Voltage (On) versus Temperature

Allegro MicroSystems 955 Perimeter Road Manchester, NH 03103-3353 U.S.A. www.allegromicro.com

Timing Accuracy versus Operational Speed

AG = 0.5 mm; relative to T_A = 25°C, S_{ROT} = 1000 rpm

Sequential Features, Mechanical Rising Edge

Signature Feature, Mechanical Falling Edge

Sequential Features, Mechanical Falling Edge

1.5 1.0 0.5 0

-0.5 -1.0

Relative Timing Accuracy (°)

Relative Timing Accuracy (°)

 $-1.5\frac{1}{0}$

True Zero Speed, Low Jitter, High Accuracy Position Sensor IC ATS627LSG

Timing Accuracy versus Operational Speed

AG = 2.5 mm; relative to T_A = 25°C, S_{ROT} = 1000 rpm

Signature Feature, Mechanical Rising Edge

Sequential Features, Mechanical Rising Edge

0 500 1000 1500 2000 2500 Operational Speed, S_{ROT} (rpm)

Signature Feature, Mechanical Falling Edge

Sequential Features, Mechanical Falling Edge

-0.5 -1.0

 $-1.5\frac{1}{0}$

Air Gap, AG (rpm)

0 0.5 1.0 1.5 2.0 2.5 3.0

True Zero Speed, Low Jitter, High Accuracy Position Sensor IC ATS627LSG

Timing Accuracy versus Air Gap

 T_A = 25°C; relative to AG = 1.5 mm, S_{ROT} = 1000 rpm

Operational Speed, S_{ROT} (rpm)

 500 -1000

-0.5 -1.0

 $-1.5\frac{1}{0}$

 1500 \rightarrow 2000

Signature Feature, Mechanical Falling Edge

Air Gap, AG (rpm)

0 0.5 1.0 1.5 2.0 2.5 3.0

Timing Accuracy versus Ambient Temperature

AG = 0.5 mm; relative to T_A = 25°C, S_{ROT} = 1000 rpm

Sequential Features, Mechanical Rising Edge

Signature Feature, Mechanical Falling Edge

Sequential Features, Mechanical Falling Edge

1.5 Relative Timing Accuracy (°) Relative Timing Accuracy (°) 1.0 0.5 $\overline{0}$ -0.5 -1.0 -1.5 –
-50 -50 -25 0 25 50 75 100 125 150 175 Ambient Temperature, TA (°C)

microsvstems

1.5 1.0 0.5 $\overline{0}$

-0.5 -1.0

Relative Timing Accuracy (°)

Relative Timing Accuracy (°)

–ا 1.5−
50–

True Zero Speed, Low Jitter, High Accuracy Position Sensor IC ATS627LSG

Timing Accuracy versus Ambient Temperature

AG = 2.5 mm; relative to T_A = 25°C, S_{ROT} = 1000 rpm

Signature Feature, Mechanical Falling Edge

Sequential Features, Mechanical Rising Edge

Ambient Temperature, T_A (°C) -50 -25 0 25 50 75 100 125 150 175

Sequential Features, Mechanical Falling Edge

microsvstems

Timing Accuracy versus Operational Speed

 T_A = 25°C; relative to AG = 1.5 mm, S_{ROT} = 1000 rpm

Signature Feature, Mechanical Falling Edge

Sequential Features, Mechanical Rising Edge

Sequential Features, Mechanical Falling Edge

THERMAL CHARACTERISTICS: May require derating at maximum conditions; see Power Derating section

*Additional thermal information available on the Allegro website

Reference Target Characteristics

Reference Target 60+2

FUNCTIONAL DESCRIPTION

Sensing Technology

The ATS627 contains a single-chip differential Hall-effect sensor IC, a samarium cobalt pellet, and a flat ferrous pole piece (concentrator). As shown in figure 5, the Hall IC supports two Hall elements, which sense the magnetic profile of the ferrous gear target simultaneously, but at different points (spaced at a 2.2 mm pitch), generating a differential internal analog voltage, V_{PROC} , that is processed for precise switching of the digital output signal.

The Hall IC is self-calibrating and also possesses a temperaturecompensated amplifier and offset cancellation circuitry. The built-in voltage regulator provides supply noise rejection throughout the operating voltage range. Changes in temperature do not greatly affect this device due to the stable amplifier design and the offset compensation circuitry. The Hall transducers and signal processing electronics are integrated on the same silicon substrate, using a proprietary BiCMOS process.

Target Profiling During Operation

An operating device is capable of providing digital information that is representative of the mechanical features of a rotating gear. The waveform diagram in figure 7 presents the automatic translation of the mechanical profile, through the magnetic profile that it induces, to the digital output signal of the ATS627. No additional optimization is needed and minimal processing circuitry is required. This ease of use reduces design time and incremental assembly costs for most applications.

Figure 5. Relative motion of the target is detected by the dual Hall elements in the Hall IC.

Figure 6. This left-to-right (pin 1 to pin 4) direction of target rotation results in a high output state when a tooth of the target gear is nearest the package face (see figure 3). A right-to-left (pin 4 to pin 1) rotation inverts the output signal polarity.

Determining Output Signal Polarity

In figure 7, the top panel, labeled Mechanical Position, represents the mechanical features of the target gear and orientation to the device. The bottom panel, labeled Device Output Signal, displays the square waveform corresponding to the digital output signal that results from a rotating gear configured as shown in figure 6, and electrically connected as in figure 9. That direction of rotation (of the gear side adjacent to the package face) is: perpendicular to the leads, across the face of the device, from the pin 1 side to the pin 4 side. This results in the IC output switching from low state to high state as the leading edge of a tooth (a rising mechanical edge, as detected by the IC) passes the package face. In this configuration, the device output switches to its high polarity when a tooth is the target feature nearest to the package. If the direction of rotation is reversed so that the gear rotates from the pin 4 side to the pin 1 side, the output polarity inverts; that is, the output signal goes high when a falling edge is detected and a valley is nearest to the package.

Figure 7: The magnetic profile reflects the geometry of the target, allowing the ATS627 to present an accurate digital output response.

Undervoltage Lockout

When the supply voltage falls below the undervoltage lockout voltage, $V_{CC(UV)}$, the device enters Reset, where the output state returns to the Power-On State (POS) until sufficient V_{CC} is supplied. This lockout feature prevents false signals, caused by undervoltage conditions, from propagating to the output of the IC.

Power Supply Protection

The device contains an on-chip regulator and can operate over a wide V_{CC} range. For devices that must operate from an unregulated power supply, transient protection must be added externally. For applications using a regulated line, EMI/RFI protection may still be required. Contact Allegro for information on the circuitry needed for compliance with various EMC specifications. Refer to figure 9 for an example of a basic application circuit.

Automatic Gain Control (AGC)

This feature allows the device to operate with an optimal internal electrical signal, regardless of the air gap (within the AG specification). At power-on, the device determines the peak-to-peak amplitude of the signal generated by the target.

This feature is also active in Running mode, though very conservatively invoked, to optimize the signal amplitude in the scenario where signal amplitude during the initial calibration period is not representative of the Running mode signal.

Automatic Offset Adjust (AOA)

The AOA circuitry automatically compensates for the effects of chip, magnet, and installation offsets. This circuitry is continuously active, including during both Power-on mode and Running mode, compensating for any offset drift (within the Allowable User-Induced Differential Offset). Continuous operation also allows it to compensate for offsets induced by temperature variations over time. This circuitry works with the AGC during calibration to adjust V_{PROC} in the internal range to allow the DACs to acquire the signal peaks.

Bounded Update

The ATS627 continuously updates its switchpoints based on the actual signal being received from the target. When the output switches, the sensor resets the tracking DACs so that each proper magnetic signal peak can be acquired. To prevent establishing switchpoints on outlier signal maxima, tracking is limited, or bounded, in magnitude. If such limiting were not applied, then anomalous target features, such as bent, broken, or misformed teeth, could create significant output accuracy errors (see figure 8).

Running Mode Lockout

The ATS627 has a Running mode lockout feature to prevent switching in response to small amplitude input signals that are characteristic of vibration signals. The internal logic of the chip interprets small signal amplitudes below a certain level to be the result of target vibration. The output is held to the state present prior to lockout, until the amplitude of the signal returns to normal operational levels.

Figure 8. Operation of Bounded Update method (for illustrative purposes only, values may not be to scale)

- Two DACs track the V_{PROC} signal: PDAC tracks positive (high) peaks, and NDAC tracks negative (low) peaks.
- The DACs track the V_{PROC} signal until a peak is reached or the bounding limit is reached. Successive Pp and Np values are used to establish the next switchpoint.

APPLICATION INFORMATION

Power Derating

The device must be operated below the maximum junction temperature of the device, $T_I(max)$. Under certain combinations of peak conditions, reliable operation may require derating supplied power or improving the heat dissipation properties of the application. This section presents a procedure for correlating factors affecting operating T_I . (Thermal data is also available on the Allegro MicroSystems website.)

The Package Thermal Resistance, $R_{\theta JA}$, is a figure of merit summarizing the ability of the application and the device to dissipate heat from the junction (die), through all paths to the ambient air. Its primary component is the Effective Thermal Conductivity, K, of the printed circuit board, including adjacent devices and traces. Radiation from the die through the device case, R_{BIC} , is a relatively small component of $R_{\theta JA}$. Ambient air temperature, T_A , and air motion are significant external factors, damped by overmolding.

The effect of varying power levels (Power Dissipation, P_D), can be estimated. The following formulas represent the fundamental relationships used to estimate T_J , at P_D .

$$
P_D = V_{IN} \times I_{IN} \tag{1}
$$

$$
\Delta T = P_D \times R_{\theta J A} \tag{2}
$$

$$
T_J = T_A + \varDelta T \tag{3}
$$

For example, given common conditions such as: $T_A = 25^{\circ}C$, V_{CC} = 12 V, I_{CC} = 7 mA, and R_{θ JA} = 126 °C/W, then:

$$
P_D = V_{CC} \times I_{CC} = 12 \text{ V} \times 7 \text{ mA} = 84 \text{ mW}
$$

$$
\Delta T = P_D \times R_{\theta J A} = 84 \text{ mW} \times 126 \text{ °C/W} = 10.6 \text{ °C}
$$

$$
T_J = T_A + \Delta T = 25 \text{ °C} + 10.6 \text{ °C} = 35.6 \text{ °C}
$$

A worst-case estimate, $P_D(max)$, represents the maximum allowable power level ($V_{CC}(max)$, $I_{CC}(max)$), without exceeding T_J (max), at a selected $R_{\theta JA}$ and T_A .

Example: Reliability for V_{CC} at $T_A = 150^{\circ}C$, package SG, using single layer PCB.

Observe the worst-case ratings for the device, specifically: $R_{\theta JA} = 126$ °C/W, T_J(max) = 165°C, V_{CC(absmax)} = 24 V, and $I_{CC} = 12$ mA.

Calculate the maximum allowable power level, $P_D(max)$. First, invert equation 3:

 ΔT (max) = T_J(max) – T_A = 165°C – 150°C = 15°C

This provides the allowable increase to T_J resulting from internal power dissipation. Then, invert equation 2:

$$
P_D(max) = \Delta T(max) \div R_{\theta JA} = 15^{\circ}\text{C} \div 126^{\circ}\text{C/W} = 119 \,\text{mW}
$$

Finally, invert equation 1 with respect to voltage:

$$
V_{CC(est)} = P_D(max) \div I_{CC} = 119 \text{ mW} \div 12 \text{ mA} = 9.9 \text{ V}
$$

The result indicates that, at T_A , the application and device can dissipate adequate amounts of heat at voltages \leq V_{CC(est)}.

Compare $V_{CC(est)}$ to $V_{CC}(max)$. If $V_{CC(est)} \leq V_{CC}(max)$, then reliable operation between $V_{CC(est)}$ and $V_{CC}(max)$ requires enhanced $R_{\theta JA}$. If $V_{CC(est)} \ge V_{CC(max)}$, then operation between $V_{CC(est)}$ and $V_{CC}(max)$ is reliable under these conditions.

Typical Application

Figure 9. Basic typical application circuit

Package SG, 4-Pin SIP

