

Automotive LED Driver Series

24-channel Constant Current Drivers and 8 line Switch Controllers Embedded Backlight LED Driver

BD94130MUF-M BD94130EFV-M

General Description

BD94130MUF-M/BD94130EFV-M are embedded 24 channel constant current drivers with 12bit PWM dimming and max 8-line switch controllers. This device can set LED constant current value by setting external ISET resistor. Communication with μ-controller via SPI is feasible.

Features

- AEC-Q100 Qualified^(Note 1)
- Functional Safety Supportive Automotive Products
- Integrated 24-channel LED Constant Current Drivers
- Integrated 4/6/8-line Switch Controllers
- SPI Interface (Cascade Connection Feasible)
■ 12bit PWM Dimming
- 12bit PWM Dimming
■ IED Constant Currer
- LED Constant Current Setting by ISET Resistor
- Phase Shift Function
- 6bit Dot Correct (50 % to 100 %)
■ LED Open Detection and LED St
- LED Open Detection and LED Short Detection
■ Adiacent LEDCH Short Detection
- Adjacent LEDCH Short Detection
- PGATE Short Protection
- VINSW Over Voltage Protection
- Slew Rate Control for PMOS Gate Driver
- Abnormality Output FAILB Pin

(Note 1) Grade 1

Application

- Cluster, Center Infotainment Display
- **Other Automotive Backlights**

Key Specification

- Power Supply Voltage Range: 3.0 V to 5.5 V
- LEDCHn Pin Voltage (n = 1 to 24): 20 V
- Maximum LED Output Current: 80 mA
- Operating Temperature Range: -40 °C to +125 °C

(BD94130MUF-M) (BD94130EFV-M)

Package W (Typ) x D (Typ) x H (Max)

VQFN56FCV080 8.0 mm x 8.0 mm x 1.0 mm

 18.5 mm x 9.5 mm x 1.0 mm

VQFN56FCV080 HTSSOP-B54

〇Product structure : Silicon integrated circuit 〇This product has no designed protection against radioactive rays.

Pin Configuration

VQFN56FCV080 (TOP VIEW) HTSSOP-B54 (TOP VIEW)

Pin Description

Pin Description – continued

Block Diagram

Description of Blocks

If there is no description, the mentioned values are typical value.

The suffixes m of the symbol represent the number of gate control ($m = 1$ to 8), and the suffixes n represent the number of current driver (n = 1 to 24), respectively. For example, LEDCHn means LEDCH1, LEDCH2,,,LEDCH24. PGATEm means PGATE1, PGATE2,,,PGATE8. This expression is applicable to the whole of this datasheet.

1. Power Supply (VCC)

The VCC pin has UVLO function (VCCUVLO), and it starts operation at VCC ≥ 2.65 V (Typ) and stops operation at VCC ≤ 2.55 V (Typ). About the condition to release/detect VCC UVLO, refer to Table 53. Connect a ceramic capacitor (Cvcc) to the VCC pin for stable operation. Cvcc range is 1 μ F to 22 μ F and recommended value is 2.2 μ F. If the Cvcc is not connected, unstable operation might occur e.g. oscillation.

2. Power Supply (VIO)

It supplies power to FAILB, SCSB, SCLK, SDI, SDO, VSYNC, and HSYNC input / output from the VIO pin. Connect a ceramic capacitor (C_{VIO}) to the VIO pin for stable operation. C_{VIO} range is 1 μ F to 4.7 μ F and recommended value is 2.2 µF. If the C_{VIO} is not connected, unstable operation might occur e.g. oscillation.

3. Power Supply (VINSW)

It supplies power to the output of PGATE1 to PGATE8 from the VINSW pin. Connect a ceramic capacitor (C_{VINSW}) to the VINSW pin to ensure stability of LED anode voltage. C_{VINSW} range is 1 µF to 100 µF and recommended value is 10 µF. If the CVINSW value is not enough, the LED output might become unstable e.g. flicker.

4. Reference Voltage (VREG15)

VREG15 block generates 1.5 V from VCC, and outputs 1.5 V to the VREG15 pin. It supplies this power (V_{VREG15}) to the internal digital circuit. The VREG15 pin has UVLO function (VREG15UVLO), and it starts operation at V_{VREG15} ≥ 1.35 V (Typ) and stops operation at V_{VREG15} \leq 1.30 V (Typ). It cannot be used to supply power to external components from this IC. About the condition to release/detect VREG15 UVLO, refer to Table 53. Connect a ceramic capacitor (CVREG15) to the VREG15 pin for phase margin. CVREG15 range is 1 µF to 4.7 µF and recommended value is 2.2 µF. If the CVREG15 is not connected, unstable operation might occur e.g. oscillation.

5. Gate Controller

PGATEm pins connect to the gate of external PMOS transistors and this block controls power source timing to LED array. Each PGATEm pin turns on in order from PGATE1 to PGATE8 in one period of VSYNC. The number of PGATEm's ON in one period of VSYNC can be set by PWMFREQ register, refer to PWMFREQ register. PGATEm output HIGH level is V_{VINSW} (Typ) and its LOW level is $V_{VINSW} - 4.5 V$ (Typ).

6. Current Driver

This device has integrated 24-channel constant current driver. Maximum LED current level LEDMAX (<80 mA) of all channels is set by the external resistor R_{ISET} of the ISET pin. The dot corrected current ILEDCHn (50 % to 100 %) is set by the register IREVmn[5:0]. And the PWM dimming is set by the register DTYmn[11:0]. (m = 1 to 8, n = 1 to 24)

Figure 1. LED Current Setting Method for Dimming

6. Current Driver – continued

(1) Output Current Setting

ILEDMAX and ILEDCHn can be calculated by the following equation. Recommended ILEDMAX setting range is from 8.5 mA to 80 mA.

$$
I_{LEDMAX} = \frac{1440k}{R_{ISET}} \qquad [mA]
$$

where RISET is the external resistor value of the ISET pin.

$$
I_{LEDCHn} = I_{LEDMAX} \times \frac{(IREVmn[5:0] + 64)}{127}
$$
 [mA]

(2) Local PWM Dimming Control

PWM dimming frequency and pulse width are set by SPI commands. Constant current driver can be controlled synchronized to the internal signal PWMn (n = 1 to 24) for each channel set by SPI.

However, the constant current driver's minimum pulse width is limited to 0.6 µs. For example, in Matrix application with HSYNC frequency 8 MHz, 12bit resolution is minimum 125 ns in 8-line controller. The controllable range is from 5 to 4095.

The accurate average current of LEDCHn is expressed as follows, considering the deadtime of line controller.

$$
I_{LED_ave} = I_{LEDCHn} \times \sum_{m=1}^{8} (DTYmn + 1) / \{(4,096 + 32 \times 2^{NOOVLAP1} + 32 \times 2^{NOOVLAP2}) \times 8\}
$$
 [mA]

where $NOOVLAP1 = 0$ to 3, $NOOVLAP2 = 0$ to 3, $MULSEL = 0$

(3) Delay Control

This IC integrates 12bit global delay function and 12bit local (each channel) delay function. In case HSYNC frequency is 8 MHz, the shift rate (global) can be set by 1.0 µs steps in 8-line controller.

If the length of the local delay and duty is more than the period, the remaining "ON" width is output at the next VSYNC period as shown on LEDCH24 of Figure 2. Refer to Delay Setting for details of this function.

6. Current Driver - continued

(4) LED Open Detection, LED Short Detection

This IC has LED Open Detection and LED Short Detection.

When LED Open/Short is detected and PWM = HIGH, it outputs FAILB = LOW and the status of LED Open/Short Detection is updated.

LED Open Detection Voltage = VOPDET

LED Short Detection Voltage = VSHDET

(LED Short Detection Voltage can be set with SPI or the external resistor of the LSPSET pin. Refer to LEDSH resister)

This IC can automatically detect or release Error condition (FAILB output, Error register) during PWMn = HIGH. The error condition is retained during PWMn = LOW.

7. FB Control

DC/DC output and LEDCHn voltage is controlled by the FB pin. If the minimum LEDCHn voltage is lower than the Feedback Reference Voltage, the FB sink current increases. If the minimum LEDCHn voltage is higher than the Feedback Reference Voltage, the FB sink current decreases.

FB sink current is controlled by FBDAC[7:0]. The output is updated at the PGATE1 = ON in every VSYNC period indicated by Figure 4. The minimum step current is 0.78 μA (Typ) and maximum sink current is 200 μA (Typ). The resistor network between this device and DC/DC should be set appropriately by considering the current ability.

During the boot interval, the soft start function works. Please refer the register SSTIM[2:0] (Address 0x000: SRSST).

Table 1. Feedback Reference Voltage

Figure 3. FB Control Block Diagram

Figure 4. FB Control Timing

Absolute Maximum Ratings (Ta = 25 °C)

Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit *between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.*

Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with thermal resistance taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

(Note 1) Wide VF variation of LED increases loss at the driver, which results in rise in package temperature. Therefore, the board needs to be designed with attention paid to heat radiation.

(Note 2) This current value is per 1ch. It needs be used within a range not exceeding power dissipation.

Thermal Resistance *(Note 1)*

(Note 1) Based on JESD51-2A (Still-Air).

(Note 2) The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

(Note 3) Using a PCB board based on JESD51-3.

(Note 5) This thermal via connect with the copper pattern of layers 1,2, and 4. The placement and dimensions obey a land pattern.

Recommended Operating Conditions

(Note 1) CLEDCH affects the transient response of LEDCHn pin. Please check the LED current waveform of the narrow PWM duty, the time margin of the LED short detection, the short check of the adjacent LEDCHn pin, the pull up charge. Please refer the register ERRMASK (0x003), PRCEN (0x005), LEDSH (0x006), LSHEXE (0x006)

Electrical Characteristics

(Unless otherwise specified V_{cC} = 3.0 V to 5.5 V, V_{VINSW} = 10 V, V_{VIO} = 1.8 V, Ta = -40 °C to +125 °C)

[LOGIC Input Block (SCSB, SCLK, SDI, VSYNC, HSYNC)]

(Note 1)

ΔILEDC1 = (ILEDn / ILED_AVE - 1) x 100

ILED_n is either pin of LED1 to LED24 current.

ILED_AVE is the average current of LED1 to LED24.

Electrical Characteristics – continued

(Unless otherwise specified V_{cC} = 3.0 V to 5.5 V, V_{VINSW} = 10 V, V_{VIO} = 1.8 V, Ta = -40 °C to +125 °C)

Application Example

Functions of Logic Block

1. Serial Interface and AC Electrical Characteristics

Serial Peripheral Interface (SPI) controls the IC with SCSB, SCLK, SDI, and SDO signals. Start the SPI communication with the initial value of SCSB is HIGH, and that of SCLK and SDI is LOW. When using several devices, connect the SDO pin to the SDI pin of the next device to make cascade connection. SDO signal outputs the SDI input after 16 SCLK pulses. Example of the N address write is shown in the following. The initial value of SDO is LOW, until it is used to output the signal. s
it
ï l

SCSB

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SPI Recommended Operation Condition (Unless otherwise specified Ta = -40 °C to +125 °C, VVIO = 1.8 V, VCC = 3.0 V to 5.5 V)

The maximum frequency of the HSYNC and VSYNC is described in the previous section "Recommended Operating Conditions".

3. SPI Connection

(1) Cascade Connection

Each device can be controlled by connecting the SCLK and the SCSB pins to all devices in parallel, and by connecting each SDO to the SDI of the next device in series. The maximum number of devices that can be cascaded is 20.

Figure 7. Image of Cascade Connection

(2) Individual Connection

Each device can be controlled by connecting the SCLK and the SDI pins to all devices in parallel, and by connecting each SCSB.

4. SPI Data Flow

MCU Write and Read flow is shown as follow. This IC has 4 timing schemes for updating the analog control data.
Type 1 (immediately): Data is updated after SPI access.

- Type 1 (immediately): Data is updated after SPI access.

Type 2 (VSYNC): Data is updated after SPI access
- Type 2 (VSYNC):

Type 3 (VSYNC+GDLY): Data is updated after SPI access and VSYNC rising edge

Type 3 (VSYNC+GDLY): Data is updated after SPI access and VSYNC rising edge Type 3 (VSYNC+GDLY): Data is updated after SPI access and VSYNC rising edge and GDLY.
Type 4 (PWM): Data is updated after SPI access and VSYNC rising edge and PWM in Data is updated after SPI access and VSYNC rising edge and PWM timing.

So there is mismatch between Read data and Control data.

Figure 9. SPI Data Flow

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5. SPI Protocol

(1) Device Address

DevAddr[5:0] of each device is calculated by counting the number of byte of 0x0000 data after the fall-edge of SCSB. When the received DevAddr[5:0] matches with the calculated DevAddr[5:0] of the device, Write/Read function occurs. When the received DevAddr[5:0] does not match the calculated DevAddr[5:0] of the device, the data is not received and is output to SDO. Refer to each protocol for the details.

(2) Number of Transferred Byte when Block Write/Read

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When S = 0 (Block Write / Read) of DevAddr[5:0], set the number of transferred byte (NumOfData) after DevAddr[5:0]. When S = 1, it skip this packet. ("Device Address" -> "Register Address" -> ….) Please access this IC using the settings as shown in Table 2 and Table 3.

SPI Setting				Access to Devices			
B	S	DevAddr[5:0]	NumOfData[8:0]	For Single Device	For All Device		Acceptable
					Same	Different	(Note)
					Data	Data	
	0	0x00	0x002 to 0x147				X
		0x01 to 0x14		Ω			O
$\mathbf 0$		$0x15$ to $0x3F$					X
		0x00	Not sending this data				X
		0x01 to 0x14		∩			O
		$0x15$ to $0x3F$					X
	0	0x00	0x002 to 0x147		Ω		O
		$0x01$ to $0x3E$					X
1		0x3F				O	O
		0x00	Not sending this data		O		O
		$0x01$ to $0x3E$					X
		0x3F				Ω	O

Table 2. Access Table for Write (RW = 0)

(Note) X: This setting is not acceptable. Do not set this condition

Table 3. Access Table for Read (RW = 1)

SPI Setting				Access to Devices			
B	S	DevAddr[5:0]	NumOfData[8:0]	For Single Device	For All Device		Acceptable
					Same	Different	(Note)
					Data	Data	
0	0	0x00	0x002 to 0x16B				
		$0x01$ to $0x14$		∩			
		$0x15$ to $0x3F$					Χ
		0x00	Not sending this data				x
		0x01 to 0x14					
		0x15 to 0x3F					
	0/1	0x00	0x002 to 0x16B				X
		$0x01$ to $0x3E$					
		0x3F					

(Note) X: This setting is not acceptable. Do not set this condition.

(3) Register Address

(4) Data

(5) Single Device, 1 Address Write (Write to Device #1)

SDI: Transfer in the order of DevAddr[5:0], RegAddr[8:0], and Data[15:0].
SDO: Output the transferred data to the next device after SDI input by 2 by

Output the transferred data to the next device after SDI input by 2 bytes.

(6) Single Device, 1 Address Write (Write to Device #3)

SDI: Transfer in the order of DevAddr[5:0], RegAddr[8:0], and Data[15:0]. SDO: Output the transferred data to the next device after SDI input by 2 bytes.

DevAddr[5:0] of each device is calculated by counting the number of byte of 0x00 data after the falling-edge of SCSB. DevAddr[5:0] = (Number of byte of 0x00 data) + 1

Figure 12. SPI Protocol for 1 Address Write to Device #3

SDI: Transfer in the order of DevAddr[5:0], NumOfData[8:0], RegAddr[8:0], and Data[15:0].

SDO: Output the transferred data to the next device after SDI input by 2 bytes.

Figure 13. SPI Protocol for N Address Write to Device #1

Figure 14. SPI Protocol for 1 Address Distinct Data Write to All Devices

0x002 0x001 0x000

Data1

0x002 0x001 0x000 Data₂

Data3

0x002 0x001 0x000

 $RegAddr[8:0] =$ 0x002: Address

5. SPI Protocol – continued

SDI: Transfer in the order of DevAddr[5:0], RegAddr[8:0], and Data[15:0].

SDO: Output the transferred data to the next device after SDI input by 2 bytes.

Figure 15. SPI Protocol for 1 Address Distinct Data Write to All Devices

(10) All Devices, Different N Address Write (Write the different N x 2 bytes data to the same RegAddr[8:0] of all devices)

SDI: Transfer in the order of DevAddr[5:0], NumOfData[8:0], RegAddr[8:0], and Data[15:0]. SDO: Output the transferred data to the next device after SDI input by 2 bytes.

Figure 16. SPI Protocol for N Address Distinct Data Write to All Devices

(11) All Devices, Same N Address Write (Write the same N x 2 bytes data to the same RegAddr[8:0] of all devices) All devices receive data

SDI: Transfer in the order of DevAddr[5:0], NumOfData[8:0], RegAddr[8:0], and Data[15:0].

SDO: Output the transferred data to the next device after SDI input by 2 bytes.

Figure 17. SPI Protocol for N Address Same Data Write to All Devices

Figure 18. SPI Protocol for 1 Address Read from Device #2

0x001 0x000

0x001 0x000 0x001 0x000

$B =$	0:	Target device receives the data
$S =$	U.	Multi
$DevAddr[5:0] =$	$0x02$:	Target device address
$NumOfData[8:0] =$	0x002:	2 address access
$RW =$		Read
$RegAddr[8:0] =$	0x003:	Address

SDI: Transfer in the order of DevAddr[5:0], NumOfData[8:0], and RegAddr[8:0].

SDO: Output the transferred data to the next device after SDI input by 2 bytes.

Figure 19. SPI Protocol for N Address Read from Device #2

(14) Example (Write the data to Device #1 and Device #2)

Example of byte transfer for 2 devices in Cascade Connection.

Figure 20. Transfer Byte Number for Multi Access

As for the register update timing, there are 4 kinds of timing as following.

Type 1. Updated to the newest data immediately when the data is written.

Type 2. Updated to the newest data at the next VSYNC. (Rising edge trigger, after the data is written.)

Type 3. Updated to the newest data at the next VSYNC and GDLY.

Type 4. Updated to the newest data at the next PWM timing. (Rising edge trigger of VSYNC, then rising edge trigger of PWM after the data is written.)

7. Description of Registers

The writing register annotated "-" is not valid.

Address 0x000: SRSST

[Read / Write] Initial value: 0x0030 Update: Immediately

The register data is updated immediately when the new data is written. SWRST is Write-only register.

Bit[6:4] SSTIM

SSTIM[2:0] is the register for setting the soft start time. It sets how the FBDAC[7:0] code changes with HSYNC.

Bit[0] SWRST

SWRST is available when HSYNC is available, because this function uses HSYNC clock.

If SWRST = 1 is written, wait for more than 10 HSYNC pulses before accessing other registers.

Address 0x001: TURNONWAIT

[Read / Write] Initial value: 0x0004 Update: VSYNC

The register data is updated at the next VSYNC signal rising edge after the data is written.

The mask time of PWM output is set by counting the number of VSYNC pulses.

This register value is updated at the 3rd VSYNC pulse after reset is released (UVLO, SWRST). If this register needs to be updated, update this register before the 3rd VSYNC pulse. Write data higher than 0x04.

 $t_{TURNOWAIT} = TURNONWAIT [7: 0]/f_{VSYNC}$ [s] (Except for waiting time until 1st VSYNC pulse)

Table 7. Maximum Turn on Wait Time

7. Description of Registers – continued

[Read / Write] Initial value: 0x003C Update: VSYNC

The register data is updated at the next VSYNC signal rising edge after the data is written. Set the value higher than 0x02. The mask time of ERROR detection is set by counting the number of VSYNC pulses after TURNONWAIT time.

 $t_{SSMASK} = SSMASK[7:0]/f_{VSYNC}$ [s] after TURNONWAIT time

(Except for waiting time until 1st VSYNC pulse)

Table 8. Maximum Soft Start Mask Time

Address 0x003: ERRMASK

[Read / Write] Initial value: 0x0029 Update: VSYNC

The register data is updated at the next VSYNC signal rising edge after the data is written.

Range: over 0x03 (Set for 0x00 to 0x02 also lead to 0x03, register value = writing value) ERROR mask time is set by counting the number of HSYNC pulses.

 $t_{ERRMASK} = ERRMASK[7:0]/f_{HSVNC}$ [s]

If the capacitance of LEDCHn pin CLEDCH is connected, the transient response is affected. Please set the value considering the time margin of LED short detection.

 $(Example)$ ERRMASK = 3: mask 3 or 4 clock (PWMn = HIGH and error signal)

It reset ERRMASK counter when PWMn = LOW. Refer to HSYNC equation about the relationship between HSYNC frequency and VSYNC frequency.

7. Description of Registers – continued

Initial value 0 0 0 0 0 0 0 0 [Read / Write] Initial value: 0x0000 Update: Immediately

The register data is updated immediately when the new data is written.

Bit[9:8] MULSEL[1:0]

Line Switch Controller setting of external PMOS gate. Refer to PWM Delay and ON Duty setting procedure each dimming mode. As for the HSYNC frequency for MULSEL, Refer to the register PWMFREQ. This register prohibits changing the setting during dimming.

Bit[7:6] PGSRCNT[1:0]

Fall Slew Rate Control of external PMOS gate.

Table 11. Fall Slew Rate of External PMOS Gate

Bit[1] PRCEN

According to setting of PRCSEL, pull up charge 'VINSW – 1.2 V' to the LEDCHn pin.

(Note) It is necessary to be careful about reverse pressure resistance.

Bit[0] PRCSEL

Pull up charge period setting of the LEDCHn pin.

Table 13. Pull up charge Period

7. Description of Registers – continued

[Read / Write] Initial value: 0x0100 Update: Immediately

The register data is updated immediately when the new data is written.

This register should be set before PWM dimming. Do not change this register value during dimming.

Bit[13] LSHEXE

Short check sequence of adjacent LEDCHn pin is executed after TURNONWAIT time if LSHEXE = 1 is written before TURNONWAIT time.

Table 14. Short Check of Adjacent LEDCHn Pin

Bit[12:10] FBREF

Feedback reference voltage of FB control block.

Bit[9:8] SMPTIM

The LED channel voltage sampling timing, after PWMn goes from LOW to HIGH. It is necessary to set PWM duty register more than 8 at the dimming.

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Address 0x006: SYSCONFIG2 – continued

Bit[7:6] PWMFREQ (Please update this register until the 4th VSYNC pulse from RESET release.)

The register PWMFREQ defines the number of times PWM turns on during a VSYNC pulse. So the proper HSYNC pulse number is almost proportional to the PWMFREQ. More specifically, considering the NOOVLAP1 and NOOVLAP2 timing, which is the deadtime of PMOSm (m = 1 to 8), the necessary HSYNC pulse number is expressed by the following equation.

 $f_{HSYNC} = f_{VSYNC} \times (4096 + a + b) \times c \times 2^{(PWMFREQ[1:0])}$

NOOVLAP1 register 0: a = 32, 1: a = 64, 2: a = 128, 3: a = 256 NOOVLAP2 register 0: b = 32, 1: b = 64, 2: b = 128, 3: b = 256 MULSEL register 0: c = 8, 1: c = 4, 2: c = 6, 3: c = 6

The ratio f $H_{\text{HSNNC}}/f_{\text{VSNNC}}$ is noted in the Table 17. Here is the example of the register MULSEL = 0 (the case $c = 8$ is substituted in the above formula)

The example of HSYNC pulse number is shown as VSYNC is 60 Hz, 120 Hz, 240 Hz and 480 Hz. The maximum HSYNC frequency is 20 MHz. (Refer to frequency range of electric characteristics)

(Note) Upper: PWM frequency Lower: HSYNC frequency "-" is not acceptable to set this value in PWMFREQ register

Address 0x006: SYSCONFIG2 – continued

Bit[4] LSHEXT

External pin or internal register setting for LED short protection voltage.

As LSHEXT = 0, please connect the LSPSET pin to GND.

As LSHEXT = 1, please set the LED short protection voltage $V_{LSPSETDET}$ by the following equation.

 $V_{LSPSETDET} = 10 \times V_{LSPSET}$

Where VLSPSET is the LSPSET pin voltage (0.5 V to 1.8 V)

Bit[3] LOPEN

This register enables/disables LED Open Error detection.

Table 20. Enable Setting for LED Open Error Detection of LEDCHn

Bit[2] LSHEN

This register enables/disables LED Short Error detection.

Table 21. Enable Setting for LED Short Error Detection of LEDCHn

Bit[1:0] LEDSH[1:0]

This register controls the detection voltage for LED Short Error.

Table 22. LED Short Error Detection Voltage Setting

[Read / Write] Initial value: 0x0000 Update: Immediately or VSYNC

The data in registers (VINSWOVPEN, VINSWOVPREF, AUTOCLR, AUTOOFF and ERRCLR) are updated immediately when the new data is written.

The data in registers (NOOVLAP1, NOOVLAP2, ERRLAT) are updated at the next VSYNC signal rising edge after the data is written. The data in registers (NOOVLAP1, NOOVLAP2) should be set before PWM dimming. AUTOCLR and ERRCLR are Write-only registers.

Bit[10] VINSWOVPEN

This register enables/disables Over Voltage Detection of the VINSW pin.

Table 23. Enable Setting for Over Voltage Detection of the VINSW pin

Bit[9:8] VINSWOVPREF[1:0]

Detection voltage for over voltage of the VINSW pin.

Table 24. Detection Voltage Setting of the VINSW pin

Bit[7:6] NOOVLAP1 (Update this register until 4th VSYNC pulse from RESET release.) None overlap time setting 1. Refer to None overlap function.

As for the HSYNC frequency for NOOVLAP1, please refer to the register PWMFREQ.

Table 25. None Overlap Time Setting1

Bit[5:4] NOOVLAP2 (Update this register until 4th VSYNC pulse from RESET release.) None overlap time setting 2. Refer to None overlap function. As for the HSYNC frequency for NOOVLAP2, please refer to the register PWMFREQ[.](file://///10.10.161.2/6fs/Kisyu/BD94130MUF_EFV/3.データシート/BD94130MUF_EFV_Rev002_未提出_200421.docx)

Table 26. None Overlap Time Setting2

Address 0x007: SYSCONFIG3 – continued

Bit[3] AUTOCLR

AUTOCLR is available in AUTOOFF = 1 setting.

Table 27. AUTOOFF Condition

Bit[2] AUTOOFF

Control ON/OFF condition in LEDCHn output. AUTOOFF condition is latched until released by UVLO or AUTOCLR.

Bit[1] ERRCLR

ERRCLR is available in ERRLAT = 1 setting.

Table 29. Clear Error Register

Bit[0] ERRLAT

Control error register and FAILB output when error is detected.

Table 30. Error Detection Function

[Read / Write] Initial value: 0x0010 Update: Immediately

The register data is updated immediately when the new data is written.

Bit[6:4] DACUP[2:0]

DACUP[2:0] is register for setting the FB DAC's count up step after soft start.

Table 31. FB DAC Code Count Up Step

Bit[3:2] DACDN[1:0]

DACDN[1:0] is register for setting the FB DAC's count down step after soft start.

Table 32. FB DAC Code Count Down Step

Bit[0] MSMODE

MSMODE is register for setting of FB DAC's controller mode or target mode.

$Table 22.$ FB DAC Mode Set

[Read / Write] Initial value: 0x00FF Update: Immediately

Address 0x00A: LEDENM

[Read / Write] Initial value: 0x00FF Update: Immediately

Address 0x00B: LEDENU

[Read / Write] Initial value: 0x00FF Update: Immediately

The register data is updated immediately when the new data is written.

These registers (0x009, 0x00A, 0x00B) enable or disable each LED channel. If '0' is set in LEDEN[n-1] (n = 1 to 24), the channel n is not available. LEDCHn current is turned off, and the status of LED Open/Short Detection and FAILB output are not affected by LEDCHn since disabled channels do not detect LED Open/Short Error.

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TSZ22111 • 15 • 001 **02.Dec.2022 Rev.002**

[Read / Write] Initial value: 0x0000 Update: VSYNC

GDLY is the delay time counted from VSYNC rising edge to PGATE1 fall-edge. The register data is updated at the next VSYNC signal rising edge after the data is written.

MULSEL register 0: c = 8, 1: c = 4, 2: c = 6, 3: c = 6 Table 35. Global Delay Setting

(Note) This count starts from VSYNC rising edge.

[Read / Write] Initial value: 0x0000 Update: PWM

The register data is updated at the next PWM signal rising edge after the data is written.

Address 0x011 to 0x0CF: DTYCNT102 to DTYCNT824

These registers are used to set the PWM pulse width. The setting procedure is the same as that for LEDCH1 with Address set to 0x010.

Address 0x0D0: DLY01

[Read / Write] Initial value: 0x0000 Update: VSYNC+GDLY

The register data is updated at the next VSYNC+GDLY timing after the data is written. DLY01 is the delay time which starts to count after NOOVLAP2.

Address 0x0D0: DLY01 (PWM Delay setting register) – continued

DLY01[11:0]	DLY01 Total Clock Number (clock width @HSYNC)
0x000	
0x001	
0x002	$\overline{2}$
0x003	3
to	to
0xFFC	4092
0xFFD	4093
0xFFE	4094
0xFFF	4095

Table 37. Delay Setting of PWM Output

(Note) This count starts from NOOVLAP2 finish point.

Address 0x0D1 to 0x0E7: DLY02 to DLY24

These registers are used to set the delay width of PWM for LEDCH2 to LEDCH24. The setting procedure is the same as that for LEDCH1 with address set to 0x0D0.

Address 0x0E8: IREV10102

Initial value | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 [Read / Write] Initial value: 0x3F3F Update: immediately

IREV101 is used with current revision of LEDCH1 of PGATE1 from 50 % to 100 %. IREV102 is used with current revision of LEDCH2 of PGATE1 from 50 % to 100 %. IREV register prohibit changing the setting during dimming.

These LED current registers should be updated before LED turns on. The dynamic update during the dimming may affect to the DCDC feedback.

$$
I_{LEDALL} = I_{LEDDC} \times \frac{(IREV_{mn}[5:0] + 64)}{127}
$$
 [mA]

Address 0x0E9 to 0x147: IREVmn

This register is used to make setting of current revision for LEDCH3 to LEDCH24 of PGATE1 and LEDCH1 to LEDCH24 of PGATE2 to PGATE8. The setting procedure is the same as that for LEDCH1 of PGATE1 with address set to 0x0E8.

[Read] Initial value: 0x0000 Update: Immediately

Address 0x149: ERLSH1H

[Read] Initial value: 0x0000 Update: Immediately

The register data is updated immediately when the data is written.

These registers (0x148, 0x149) correspond to the status of LED Short Detection of PGATE1 of LED1 to LED24.

(Note 1) ERRLAT = 0: ERLSHm[n-1] (m = 1 to 8, n = 1 to 24) turns 0, if LED Short Error is released or LEDEN[n-1] = 0 is set or LSHEN = 0 is set. ERRLAT = 1: ERLSHm $[n-1]$ turns 0, if ERRCLR = 1 is set.

Address 0x14A to 0x157: ERLSHm[n-1]

These registers (0x14A to 0x157) correspond to the status of LED Short Detection of PGATE2 to PGATE8 of LED1 to LED24. The setting procedure is the same as that for LED1 to LED24 of PGATE1 with address set to 0x148 and 0x149.

[Read] Initial value: 0x0000 Update: Immediately

Address 0x159: ERLOP1H

[Read] Initial value: 0x0000 Update: Immediately

The register data is updated immediately when the data is written.

These registers (0x158, 0x159) correspond to the status of LED Open Detection of PGATE1 of LED1 to LED24.

(Note 2) ERRLAT = 0: ERLOPm[n-1] (m = 1 to 8, n = 1 to 24) turns 0,if LED Open Error is released or LEDEN[n-1] = 0 is set or LOPEN = 0 is set. ERRLAT = 1: ERLOPm $[n-1]$ turns 0, if ERRCLR = 1 is set.

Address 0x15A to 0x167: ERLOPm[n-1]

These registers (0x15A to 0x167) correspond to the status of LED Short Detection of PGATE2 to PGATE8 of LED1 to LED24. The setting procedure is the same as that for LED1 to LED24 of PGATE1 with Address set to 0x158 and 0x159.

7. Description of Registers – continued Address 0x168: EROTHER

[Read] Initial value: 0x0000 Update: Immediately

The register data is updated immediately when the new data is written.

Bit[5]: EXENG

EXENG register correspond to the status that short check sequence of adjacent LEDCHn is not executed.

Bit[4]: EXEOK

EXEOK register correspond to the status that short check sequence of adjacent LEDCHn is executed.

Bit[3]: ERISETOPEN

ERISETOPEN register correspond to the status of ISET Open Detection.

Table 43. Status of ISET Open Detection

Bit[2]: ERISETOCP

ERISETOCP register correspond to the status of ISET Over Current Detection.

Table 44. Status of ISET Over Current Detection

Bit[0]: ERVINSWOVP

ERVINSWOVP register correspond to the status of VINSW Over Voltage Detection.

Table 45. Status of VINSW Over Voltage Detection

(Note 3) ERRLAT = 0: ERISETOPEN, ERISETOCP and ERVINSWOVP turns 0, if error condition is released. ERRLAT = 1: ERISETOPEN, ERISETOCP and ERVINSWOVP turns 0, if ERRCLR = 1 is set.

[Read] Initial value: 0x0000 Update: Immediately

Address 0x16A: ERLEDH

[Read] Initial value: 0x0000 Update: Immediately

The register data is updated immediately when the data is written.

These registers (0x169, 0x16A) correspond to the status of Short Detection of adjacent LEDCHn.

Table 46. Status of Short Detection of Adjacent LEDCHn

Address 0x16B: ERPGSH

[Read] Initial value: 0x0000 Update: Immediately

The register data is updated immediately when the data is written.

Bit[15:8]: ERPGVINSH[7:0]

This register correspond to the status of Short Detection between the PGATEm pin and the VINSW pin.

Table 47. Status of Short Detection between the PGATEm pin and the VINSW pin

Bit[7:0]: ERPGGSH[7:0]

This register correspond to the status of Short Detection between the PGATEm pin and the GND.

(Note 4) ERLED[n-1], ERPGVINSH[m-1] and ERPGGSH[m-1] turn 0, if ERRCLR = 1 is set.

Application Circuit Diagram

1. The Example of Basic Application

2. The Plural BD94130 Usage (the common SPI and the common DCDC)

Timing Chart

Figure 21. The Boot Sequence for SPI Command Mode

Turn ON Sequence

- [1] Power on VCC and VCCUVLO is released. And power on VINSW.
- [2] Power on VIO and VIOUVLO is released.
- [3] After the EN pin is H, VREG15 turn on. The signal RESET is expressed by the following equation. After RESET is released, the registers can be accessed.
	- RESET = VCCULVO or VIOUVLO or EN or VREG15UVLO or Register
- [4] Set the initial registers until 4th VSYNC period from RESET release. 4th VSYNC period is adjustable by the register TURNONWAIT[7:0]. During the state TURNONWAIT, the IC keeps LEDs off.
- [5] The duty register DTYCNTmn are updated in every VSYNC period for dimming.

Turn OFF Sequence

- [6] Set the register LEDENL, LEDENM and LEDENU registers to 0.
- [7] VREG15 turn off after the EN pin is L. The registers cannot be accessed during RESET = L.
- [8] Power off VIO and VIOUVLO is detected.
- [9] Power off VINSW and VCC.

The first turn on and the last turn off are VCC. And the order of the VIO, EN can be exchanged.

1. Boot Sequence – continued

Turn ON Sequence

- [1] Power on VCC and VCCUVLO is released. And power on VINSW.
- [2] Power on VIO and VIOUVLO is released.
- [3] After the EN pin is H, VREG15 turn on. The signal RESET is expressed by the following equation. After RESET is released, the VSYNC signal becomes valid. And the TEST1 pin must be H.
	- RESET = VCCULVO or VIOUVLO or EN or VREG15UVLO or Register
- [4] Until 4th VSYNC period from RESET release, the IC keeps LEDs off, as TURNONWAIT.
- [5] Input PWM pulse to VSYNC for dimming. VSYNC signal can control LED current directly. PGATEm are all on.

Turn OFF Sequence

- [6] Stop PWM dimming pulse input to VSYNC.
- [7] VREG15 turn off after the EN pin is L.
- [8] Power off VIO and VIOUVLO is detected.
- [9] Power off VINSW and VCC.

The first turn on and the last turn off are VCC. And the order of the VIO, EN can be exchanged.

About PWM Direct Control Mode

If 8 HSYNC clocks counts, PWM Direct Control Mode is shifted to SPI Command Mode.

Even if error is detected, LED keeps ON, and FAILB signal is still HIGH.

Figure 23. SPI Command Mode / PWM Direct Control Mode

2. Matrix Operation and PWM Dimming Setting

(1) Matrix Operation Setting (the line number of gate controller)

The register MULSEL[1:0] controls the number of active PGATE as shown in Figure 24. The unused PGATE5 to PGATE8 asserts always OFF. The selectable gate number is 4, 6, and 8. The necessary HSYNC clock number is depend on the register MULSEL[1:0].

(2) Matrix Operation Setting (PWM frequency)

The register PWMFREQ[1:0] controls the repeated number of active PGATE for VSYNC period. The figure 25 show the example of the one repeat and two repeats. The selectable repeated number is 1, 2, 4, and 8. The necessary HSYNC clock number is depend on the register PWMFREQ[1:0].

Figure 25. PWMFREQ vs LEDCHn output

2. Matrix Operation and PWM Dimming Setting – continued (3) PWM Delay Setting

There are 2 kinds of delay setting.

The register GDLY set the interval from VSYNC to PGATE1, and the GDLY delay affects all PGATEm accordingly. The register DLYn (n = 1 to 24) set the interval from PGATE = ON to the current rising of LEDCHn. (That interval is expressed NOOVLAP2 + DLYn in detail.)

By shifting the starting timing of each LED current, the transient response of the total current is averaged.

Figure 26 shows the delay setting over 1 PGATE in 8 Line Gate Controller. The setting in the figure is 75 % Duty and 50 % Delay for LEDCH1.

If the LED current is finished within the single period of PGATE = ON, the delay setting of DTYmn is expressed as following.

> $0 \leq DLY_n < 4095$ $DTY_{mn} + DLY_n \leq 4095$

Figure 27. Dimming Sequence for Normal Operation

- [1] The register DTYmn access is finished within the present VSYNC period to reflect in the next VSYNC period. If that access is not finished by the VSYNC, the register is not reflected correctly.
- [2] Buffer1 data is updated at VSYNC timing.
- [3] Buffer2 data is updated at VSYNC+GDLY timing.
- [4] Control data (DTYmn) is updated after the delay setting DLYn in the next VSYNC period, DTYmn is reflected to the LEDCHn current.

3. None Overlap Function

None overlap time between PMOSm can be adjustable by the register NOOVLAP1, NOOVLAP2.

NOOVLAP1 is the interval from PMOSm = OFF to PMOS($m+1$) = ON. This is set longer than the PMOS off delay not to cause PMOS = ON simultaneously.

NOOVLAP2 is the interval from PMOS = ON to the beginning of LEDCHn current. This is set longer than the PMOS on delay.

These register adjustable such as 32 clock, 64 clock, 128 clock, 256 clock by HSYNC.

The necessary clock of HSYNC is changed accordingly. Please refer the section of Description of PWMFREQ[1:0] (Address 0x006).

(Note 1) Internal signal for counting PGATE ON timing and LEDCH1 current ON timing.

Figure 28. PGATE1, PGATE2 None Overlap Timing

4. PWM Behavior at Close VSYNC Intervals

In this section, PWM dimming behavior is shown if HSYNC is not equal to ideal frequency. The ideal frequency of HSYNC is 33280 times of VSYNC below example.

(1) HSYNC Frequency less than Ideal Frequency

Example: Delay = 0, Duty = 75 %, PWMFREQ = 0, NOOVLAP1 = NOOVLAP2 = 0, MULSEL = 0

The main counter is reset at the rising edge of VSYNC. The main counter starts counting up by HSYNC and proceed the line control from PGATE1 to PGATE8.

[1] As HSYNC is equal to the ideal frequency, the main counter reaches the full value 33280. The ON interval of PGATE8 and PWMn are proper.

[2] As HSYNC is smaller than the ideal frequency, the main counter does not reach the full value. The ON interval of PGATE8 and PWMn are short.

(2) HSYNC Frequency more than Ideal Frequency

Example: Delay = 0, Duty = 50 %, PWMFREQ = 0, NOOVLAP1 = NOOVLAP2 = 0, MULSEL = 0

[1] As HSYNC is more than the ideal frequency, the main counter continues the full value 33280 without reset. In this blank interval after PGATE8 = OFF, all LEDs turn off. The ON interval of PGATE8 and PWMn are almost proper, but all LEDs brightness is LOW.

5. ERROR Detection and Release

The following are the internal signals on the timing chart:

(1) LED Open Detection

LED Open Error is detected after ERRMASK, and LED Open Error is released as shown in Figure 31. Here $ERRMASK[7:0] = 0x03$

If PWM_OH[n-1] is shorter than ERR_MSKCNTmn[7:0], the LED OPEN is not detected.

Figure 31. LED Open Detection and Release

(1) LED Open Detection - continued

[Case: LED Open Error signal is LOW width]

While PWM_OH[1] = HIGH and LOPDET_IL[1] = LOW with SSEND = HIGH (Soft Start end), if ERR_MSKCNT does not counts up until the ERRMASK, FAILB remains HIGH.

In the same condition, if ERR_MSKCNT counts up until the ERRMASK, FAILB asserts LOW.

Figure 32. LED Open Detection (the error signal is LOW width)

(2) LED Short Detection

(Example) ERRMASK[7:0] = 0x03

LED Short Error is detected after ERRMASK, and LED Short Error is released as shown in Figure 33. If the capacitance of LEDCHn pin CLEDCH is connected, the transient response is affected. Please set the ERRMASK value considering the time margin of LED short detection.

Figure 33. LED Short Detection and Release

(3) VINSW Over Voltage Detection

(Example) VINSWOVPEN = 1, ERRLAT = 0

VINSW Over Voltage is detected after SSEND, and VINSW Over Voltage Error is released as shown in Figure 34. LEDCHn output is not turned off by VINSW Over Voltage Error.

(4) ISET Over Current Detection

(Example) ERRLAT = 0

ISET Over Current is detected after SSEND, and ISET Over Current Error is released as shown in Figure 35. LEDCHn output is turned off by ISET Over Current Error.

Figure 35. ISET Over Current Detection

(5) ISET Open Detection

(Example) ERRLAT = 0 ISET Open is detected after SSEND, and ISET Open Error is released as shown in Figure 36. LEDCHn output is not turned off by ISET Open Error.

Figure 36. ISET Open Detection

(6) Short PGATEm to VINSW Detection (Example) MULSEL = 0

Short PGATEm to VINSW is detected the timing that PGATEm goes from LOW to HIGH during dimming mode. Detected PGATEm becomes Hi-z output. Short Error can release by ERRCLR.

Figure 37. Short PGATEm to VINSW Detection

(7) Short PGATEm to GND Detection (Example) MULSEL = 0

Short PGATEm to GND is detected the timing that PGATEm goes from HIGH to LOW during dimming mode. Detected PGATEm becomes Hi-z output. Short Error can release by ERRCLR.

(8) Short Check Detection of Adjacent LEDCHn

The short check sequence of adjacent the LEDCHn pin is executed at the end of TURNONWAIT interval, if the register LSHEXE = 1 is written by the end of TURNONWAIT interval, where EXEOK status is HIGH. If LSHEXE = 1 is written after TURNONWAIT interval, the short check sequence is not executed and the register EXENG = 1. The short check result can be read from the register ERLED[23:0]. Example both ERLED[5] and ERLED[6] is HIGH, the LEDCH6 pin and the LEDCH7 pin can be judged as short pin.

Figure 39. Short Check Detection of Adjacent LEDCHn

(9) Soft-start Masking Function

LED Open Error cannot be detected during Soft Start (SSEND = LOW). Soft start counter counts up every VSYNC period until the SSMASK setting (SSEND = HIGH) as shown Figure 40 below. LED Open Error can be detected when SSEND = HIGH. It is also the same when LED Short Error is detected.

Time of mask = (TURNONWAIT register + SSMASK register) x VSYNC $(Example)$ SSMASK = 0x3C

Figure 40. Setting for Soft Start Mask

(10) Error Sequence for the Register AUTOOFF AUTOOFF set the abnormal LED = OFF automatically. (Case) the register ERRLAT = 0 and AUTOOFF = 1

Figure 41. Error Sequence for the register AUTOOFF = 1

- [1] If LED Short Error is detected, FAILB asserts LOW.
- [2] LEDCH1 output is turned off automatically. (Only target timing (PGATE1))
- [3] The external condition turns to normal. The LEDCH1 = OFF continues, and FAILB keeps LOW.
- [4] By reading the register ERLSHmn, ERLOPmn, the abnormal LED component can be distinguished.
- [5] Once LEDCH1(PGATE1) is off automatically, IC does not judge the LED Short Error status. The LED keeps off.
- [6] The register AUTOCLR = 1 is written, The automatical off status is cleared.

[7] IC judges LED Short Error. As the external condition is normal, LED turns on and ERLSH1 = 0x000000 and FAILB $=$ HIGH.

(Case) the register ERRLAT = 0 and AUTOOFF = 0

- [2] When the abnormal is detected, LED does not turn off automatically.
- [3] If the abnormal state is released, LED turn on again.

Figure 42. Error Sequence for the register ERRLAT = 1

[1] If LED Short Error is detected, FAILB asserts LOW.

[2] LEDCH1 output is still turned on, because AUTOOFF is 0.

[3] The external condition turns to normal. The register ERLSH1 keeps 0x000001 and FAILB asserts LOW.

[4] By reading the register ERLSHmn, ERLOPmn, the abnormal LED component can be distinguished.

[5] LEDCH1 output is still turned on, because AUTOOFF is 0.

[6] The register ERRCLR = 1 is written, The register ERLSH1 is cleared to 0x000000 and FAILB asserts HIGH.

Condition for Protections

(Note 1) When the IC detects VCCUVLO or VREG15UVLO or TSD or EN, it cannot detect other protection.

(Note 2) Write LEDEN[n-1] = 0 when the error channel is turned off.

(Note) m = 1 to 8, n = 1 to 24

Table 50. Protection Table 2

(Note 1) When the IC detects VCCUVLO or VREG15UVLO or TSD or EN, it cannot detect other protection.

Condition for Protections – continued

(Note 1) When the IC detects VCCUVLO or VREG15UVLO or TSD or EN, it cannot detect other protection. *(Note)* m = 1 to 8

'-': It does not have the function.

'-': It does not have the function.

(Note 1) When the IC detects VCCUVLO or VREG15UVLO or TSD or EN, it cannot detect other protection. *(Note)* n = 1 to 24

Condition for Protections – continued

(Note 1) When the IC detects VCCUVLO or VREG15UVLO or TSD or EN, it cannot detect other protection.

I/O Equivalence Circuit VCC | FB | EN **VCC** Θ \exists FB VCC_D EN GND $\overline{10k}$ 5k ⊣⊢ 100k GND GND GND GND / LGND VREG15 ISET **VCC** VCC VCC GND VREG15 $70k$ ¹ 1k¹ LGND[[] 1k 330k ISET GND 10k GND GND GND HSYNC / VSYNC / SCSB / SDI / SCLK VIO VIO SDO \overline{V} VIO VIO HSYNC SDO VSYNC SCSB SDI 1k ظا $\frac{1}{\sin \theta}$ **SCLK** GND GND GND FAILB **LEDCH1** to LEDCH24 PGATE1 to PGATE8 / VINSW VINSW LEDCH1 $\overline{\overline{\text{GND}}}$ PGATE1 <u>duithe</u> to FAILB to LEDCH24 $\underbrace{\frac{1}{2}}_{\text{LGND}}$ PGATE8 $1M$ GND GND Г GND LGND ⊐ GND

I/O Equivalence Circuit – continued

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

7. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

8. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

9. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

Operational Notes – continued

10. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

Figure 43. Example of Monolithic IC Structure

11. Ceramic Capacitor

When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

12. Thermal Shutdown Circuit (TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's maximum junction temperature rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF power output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

13. Over Current Protection Circuit (OCP)

This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

14. Functional Safety

"ISO 26262 Process Compliant to Support ASIL-*"

A product that has been developed based on an ISO 26262 design process compliant to the ASIL level described in the datasheet.

"Safety Mechanism is Implemented to Support Functional Safety (ASIL-*)"

A product that has implemented safety mechanism to meet ASIL level requirements described in the datasheet. "Functional Safety Supportive Automotive Products"

A product that has been developed for automotive use and is capable of supporting safety analysis with regard to the functional safety.

Note: "ASIL-*" is stands for the ratings of "ASIL-A", "-B", "-C" or "-D" specified by each product's datasheet.

Ordering Information

Marking Diagram

Datasheet

Physical Dimension and Packing Information - continued

Revision History

Notice

Precaution on using ROHM Products

1. If you intend to use our Products in devices requiring extremely high reliability (such as medical equipment $(Note 1)$, aircraft/spacecraft, nuclear power controllers, etc.) and whose malfunction or failure may cause loss of human life, bodily injury or serious damage to property ("Specific Applications"), please consult with the ROHM sales representative in advance. Unless otherwise agreed in writing by ROHM in advance, ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of any ROHM's Products for Specific Applications.

2. ROHM designs and manufactures its Products subject to strict quality control system. However, semiconductor products can fail or malfunction at a certain rate. Please be sure to implement, at your own responsibilities, adequate safety measures including but not limited to fail-safe design against the physical injury, damage to any property, which a failure or malfunction of our Products may cause. The following are examples of safety measures:

[a] Installation of protection circuits or other protective devices to improve system safety

[b] Installation of redundant circuits to reduce the impact of single or multiple circuit failure

- 3. Our Products are not designed under any special or extraordinary environments or conditions, as exemplified below. Accordingly, ROHM shall not be in any way responsible or liable for any damages, expenses or losses arising from the use of any ROHM's Products under any special or extraordinary environments or conditions. If you intend to use our Products under any special or extraordinary environments or conditions (as exemplified below), your independent verification and confirmation of product performance, reliability, etc, prior to use, must be necessary:
	- [a] Use of our Products in any types of liquid, including water, oils, chemicals, and organic solvents
	- [b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
	- [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl2, H₂S, NH₃, SO₂, and NO₂
	- [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
	- [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
	- [f] Sealing or coating our Products with resin or other coating materials
	- [g] Use of our Products without cleaning residue of flux (Exclude cases where no-clean type fluxes is used. However, recommend sufficiently about the residue.); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
	- [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
- 5. Please verify and confirm characteristics of the final or mounted products in using the Products.
- 6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse, is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- 7. De-rate Power Dissipation depending on ambient temperature. When used in sealed area, confirm that it is the use in the range that does not exceed the maximum junction temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
- 9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

- 1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- 2. In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

Precautions Regarding Application Examples and External Circuits

- 1. If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
- 2. You agree that application notes, reference designs, and associated data and information contained in this document are presented only as guidance for Products use. Therefore, in case you use such information, you are solely responsible for it and you must exercise your own independent verification and judgment in the use of such information contained in this document. ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of such information.

Precaution for Electrostatic

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of Ionizer, friction prevention and temperature / humidity control).

Precaution for Storage / Transportation

- 1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
	- [a] the Products are exposed to sea winds or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
	- [b] the temperature or humidity exceeds those recommended by ROHM
	- [c] the Products are exposed to direct sunshine or condensation
	- [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

Precaution for Product Label

A two-dimensional barcode printed on ROHM Products label is for ROHM's internal use only.

Precaution for Disposition

When disposing Products please dispose them properly using an authorized industry waste company.

Precaution for Foreign Exchange and Foreign Trade act

Since concerned goods might be fallen under listed items of export control prescribed by Foreign exchange and Foreign trade act, please consult with ROHM in case of export.

Precaution Regarding Intellectual Property Rights

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