

Application Manual

RV-8803-C7

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DTCXO Temp. Compensated Real-Time Clock Module with I²C-Bus Interface

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Micro Crystal

RV-8803-C7

DTCXO Temp. Compensated Real-Time Clock Module with I²C-Bus Interface

1. OVERVIEW

- RTC module with built-in "Tuning Fork" crystal oscillating at 32.768 kHz
- Counters for hundredths of seconds, seconds, minutes, hours, date, month, year and weekday
- Factory calibrated temperature compensation
- Very high Time Accuracy
	- \circ ±1.5 ppm 0 to +50 \degree C
	- \circ ± 3.0 ppm -40 to +85 \degree C
	- \circ ±7.0 ppm +85 to +105 °C
	- o Aging compensation with OFFSET value
- \cdot I²C-bus interface (up to 400 kHz)
- Periodic Countdown Timer Interrupt function
- Periodic Time Update Interrupt function (seconds, minutes)
- Alarm Interrupts for weekday or date, hour and minute settings
- External Event Input with Interrupt and Time Stamp function
- Programmable Clock Output for peripheral devices (32.768 kHz, 1024 Hz, 1 Hz) with enable/disable function (CLKOE)
- Automatic leap year correction: 2000 to 2099
- Internal Power-On Reset (POR)
- Low voltage detector
- Wide operating voltage range: 1.5 V to 5.5 V
- Very low current consumption: 240 nA (V_{DD} = 3.0 V, T_A = 25°C)
- Operating temperature range: -40 to +85°C (supports extended range from +85°C to +105°C with limitations)
- Ultra small and compact C7 package size, RoHS-compliant and 100% leadfree: 3.2 x 1.5 x 0.8 mm
- Register compatible with Epson RX-8803SA/LC
- Automotive qualification according to AEC-Q200 available

1.1. GENERAL DESCRIPTION

The RV-8803-C7 is a highly accurate real-time clock/calendar module due to its built-in Thermometer and Digital Temperature Compensation circuitry (DTCXO). The Temperature Compensation circuitry is factory calibrated and results in highest time accuracy of ±3.0 ppm across the temperature range from -40 to +85°C and a time accuracy of ±7.0 ppm for the extended range from +85°C to +105°C, and additionally offers an aging offset correction.

The RV-8803-C7 has the smallest package and the lowest current consumption among all temperature compensated RTC modules. Due to its special architecture the RV-8803-C7 provides a very low current consumption of 240 nA.

1.2. APPLICATIONS

The RV-8803-C7 RTC module combines key functions with outstanding performance in an ultra-small ceramic package:

- Factory calibrated Temperature Compensation with temperature measuring every second
- Ultra-Low Power consumption
- Smallest RTC module (embedded XTAL) in an ultra-small 3.2 x 1.5 x 0.8 mm leadfree ceramic package.

These unique features make this product perfectly suitable for many applications:

- Communication: IoT / Wearables / Wireless Sensors and Tags / Handsets
- Automotive: M2M / Navigation & Tracking Systems / Dashboard / Tachometers / Engine Controller Car Audio & Entertainment Systems
- Metering: E-Meter / Heating Counter / Smart Meters / PV Converter/ Utility metering
- Outdoor: ATM & POS systems / Surveillance & Safety systems / Ticketing Systems
- Medical: Glucose Meter / Health Monitoring Systems
Safety: Security & Camera Systems / Door Lock & /
- Security & Camera Systems / Door Lock & Access Control / Tamper Detection
- Consumer: Gambling Machines / TV & Set Top Boxes / White Goods
- Automation: PLC / Data Logger / Home & Factory Automation / Industrial and Consumer Electronics

1.3. ORDERING INFORMATION

Example: RV-8803-C7 TA QC

2. BLOCK DIAGRAM

2.1. PINOUT

2.2. PIN DESCRIPTION

2.3. FUNCTIONAL DESCRIPTION

The RV-8803-C7 is a high accurate, ultra-low power CMOS based Real-Time-Clock Module with embedded 32.768 kHz Crystal. The Xtal 32.768 kHz clock itself is not temperature compensated.

The very high Time Accuracy and stability of ± 3.0 ppm over the temperature range from -40°C to +85°C and of ±7.0 ppm for the extended range from +85°C to +105°C is achieved by the built-in Digital Temperature Compensation circuitry (DTCXO). The factory calibrated correction values are located in the EEPROM and are not accessible for the user. Additionally, there is an Offset Register customer use for aging correction.

The RV-8803-C7 provides standard Clock & Calendar function including seconds, minutes, hours (24), weekdays, date, months, years (with leap year correction) and interrupt functions for an External Event, Periodic Countdown Timer, Periodic Time Update and Alarm. Beside the standard RTC functions, it includes an integrated Temperature Sensor, a Time Stamp function for the External Event Input and 1 Byte of User RAM and offers an I²C-bus (2-wire Interface). Further 2 Bytes can be used as User RAM when the Periodic Countdown Timer is not used (Timer Counter registers 0Bh, 1Bh and 0Ch, 1Ch) and further 3 Bytes when the Alarm function is not used (Alarm registers 08h, 18h; 09h, 19h and 0Ah, 1Ah).

The registers are accessed by selecting a register address and then performing read or write operations. Multiple reads or writes may be executed in a single access, with the address automatically incrementing after each byte. When address is automatically incremented, wrap around occurs from the address FFh to the address 00h (see figure below).

Handling address registers:

2.4. DEVICE PROTECTION DIAGRAM

3. REGISTER ORGANIZATION

Registers are accessed by selecting a register address and then performing read or write operations. Multiple reads or writes may be executed in a single access, with the address automatically incrementing after each byte. The following tables Register Definitions (00h to 0Fh), (10h to 1Fh) and (20h to 2Fh) summarize the function of each register. In the table Register Definitions (00h to 0Fh) and (10h to 1Fh) the GPx bits (where x is between 0 and 5) are 6 register bits which may be used as general purpose storage. These bits are not described in the sections below. All of the GPx bits are cleared when the RV-8803-C7 powers up, and they can therefore be used to allow software to determine if a true Power On Reset has occurred, or to hold other initialization data.

- Address 00h to 0Fh: Basic time and calendar register A dds RAM
Address 10h to 1Fh: Extension register Φ Adds 100th Seconds counter
- Address 10h to 1Fh: Extension register $\mathbb O$
-
- Address 20h to 2Fh: Extension register 2 Capture buffer and Event control
-

Note: When writing or reading a specific function value into/from the Address range 00h to 0Fh the value will be automatically updated in the Address range 10h to 1Fh and vice versa.

In order to not corrupt the accuracy of the temperature compensation and the Time Stamp (Capture) function on the highest 100th Seconds resolution, it is not possible to freeze the clock and calendar register during read-out process, as it is common practice for other RTC's.

Since the time and calendar registers cannot be frozen, there might be a condition that the time registers are incremented while read-out. To avoid reading corrupted (partially incremented) data, special measures and procedures need to be applied (see [TIME DATA READ\)](#page-41-0).

3.1. REGISTER OVERVIEW

After reset, all registers are set according to Table in section [REGISTER RESET VALUES SUMMARY.](#page-24-0)

Register Definitions, Address 00h to 0Fh (Basic time and calendar register):

Register Definitions, Address 10h to 1Fh (Extension register):

Register Definitions, Address 20h to 2Fh (Extension register):

3.2. CLOCK REGISTERS

10h - 100th Seconds (Read Only)

This register holds the count of hundredths of seconds, in two binary coded decimal (BCD) digits. Values will be from 00 to 99.

00h, 11h - Seconds

This register holds the count of seconds, in two binary coded decimal (BCD) digits. Values will be from 00 to 59.

01h, 12h - Minutes

This register holds the count of minutes, in two binary coded decimal (BCD) digits. Values will be from 00 to 59.

02h, 13h - Hours

This register holds the count of hours, in two binary coded decimal (BCD) digits. Values will be from 00 to 23.

3.3. CALENDAR REGISTERS

03h, 14h - Weekday

This register holds the current day of the week. Each bit represents one weekday that is assigned by the user. Values will range from 1 to 7. Do not set 1 to more than one bit.

04h, 15h – Date

This register holds the current day of the month, in two binary coded decimal (BCD) digits. Values will range from 00 to 31. The Reset value XX after POR has to be replaced by a valid initial value (01 to 31). Leap years are correctly handled from 2000 to 2099.

05h, 16h - Month

This register holds the current month, in two binary coded decimal (BCD) digits. Values will range from 01 to 12.

06h, 17h - Year

This register holds the current year, in two binary coded decimal (BCD) digits. Values will range from 00 to 99. Leap years are correctly handled from 2000 to 2099.

07h - RAM

This register holds the bits for general purpose use.

3.4. ALARM REGISTERS

08h, 18h – Minutes Alarm

This register holds the Minutes Alarm Enable bit AE_M and the alarm value for minutes, in two binary coded decimal (BCD) digits. Values will range from 00 to 59.

09h, 19h – Hours Alarm

This register holds the Hours Alarm Enable bit AE_H and the alarm value for hours, in two binary coded decimal (BCD) digits. Values will range from 00 to 23.

0Ah, 1Ah – Weekday/Date Alarm

This register holds the Weekday/Date Alarm Enable bit AE_WD. If the WADA bit is 0 (Bit 6 in Register 0Dh, 1Dh), it holds the alarm value for the day of the week (weekdays assigned by the user). Multiple weekdays can be selected. Values will range from 0000001 to 1111111. If the WADA bit is 1, it holds the alarm value for the date, in two binary coded decimal (BCD) digits. Values will range from 01 to 31. Leap years are correctly handled from 2000 to 2099.

Weekday Alarm when WADA = 0 (Bit 6 in Register 0Dh, 1Dh)

Date Alarm when WADA = 1 (Bit 6 in Register 0Dh, 1Dh)

3.5. PERIODIC COUNTDOWN TIMER CONTROL REGISTERS

0Bh, 1Bh – Timer Counter 0

This register is used to set the lower 8 bits of the Timer Value (preset value) for the Periodic Countdown Timer.

0Ch, 1Ch – Timer Counter 1

This register is used to set the upper 4 bits of the Timer Value (preset value) for the Periodic Countdown Timer.

 (1) This specific function accessed in Address range 00h to 0Fh is automatically updated in Address range 10h to 1Fh and vice versa.

Countdown Period in seconds:

Timer Value
Countdown Period = Timer Clark Fram Timer Clock Frequency

3.6. EXTENSION REGISTER

0Dh, 1Dh – Extension Register

This register is used to specify the target for the Alarm Interrupt function and the Periodic Time Update Interrupt function and to select or set operations for the Periodic Countdown Timer.

3.7. FLAG REGISTER

0Eh, 1Eh – Flag Register

This register holds a variety of status bits. The register may be written at any time to clear any status flag.

3.8. CONTROL REGISTER

0Fh, 1Fh – Control Register

This register is used to control the interrupt event output from the INT pin and the stop/start status of clock and calendar operations.

3.9. OFFSET REGISTER

2Ch – Offset Register

This register holds the OFFSET value for the aging correction.

3.10.CAPTURE BUFFER/EVENT CONTROL REGISTERS

20h – 100th Seconds CP (Read Only)

This register holds a captured (copied) value of the 100th Seconds register (Time Stamp), in two binary coded decimal (BCD) digits. The values are from 00 to 99.

21h - Seconds CP (Read Only)

This register holds a captured (copied) value of the Seconds register (Time Stamp), in two binary coded decimal (BCD) digits. The values are from 00 to 59.

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2Fh – Event Control

This register controls the event detection on the EVI pin. Depending of the EHL bit a high or a low signal can be detected. Moreover a digital glitch filtering can be applied to the EVI signal by selecting a sampling period in the ET field.

3.11.REGISTER RESET VALUES SUMMARY

 (1) This specific function accessed in Address range 00h to 0Fh is automatically updated in Address range 10h to 1Fh and vice versa.

RV-8803-C7 reset values after power on:

4. DETAILED FUNCTIONAL DESCRIPTION

4.1. POWER ON RESET (POR)

The power on reset (POR) is generated at start-up (see [POWER ON AC ELECTRICAL CHARACTERISTICS\)](#page-60-0). All registers including the Counter Registers are initialized to their reset values (see [REGISTER RESET VALUES](#page-24-0) [SUMMARY\)](#page-24-0).

4.2. POWER MANAGEMENT

The circuit is always on and each temperature sensing interval, i.e. every second, is temperature compensated. The digital part is always on, but some functions are clock gated (like I²C). By default, at power up, the circuit will always go to the lower power consumption mode (power-off). Detecting an activity on the I²C will wake-up the digital part of the circuit. To achieve the specified time keeping current consumption, extra features like CLKOUT and I²C interface need to be inactive.

4.3. CLOCK SOURCE

The built-in 32.768 kHz crystal is the clock source for the digital part. After thermal compensation, the RV-8803-C7 provides a very accurate time with temperature compensation for an outstanding low current consumption.

4.4. INTERRUPT OUTPUT

The interrupt pin $\overline{\text{INT}}$ can be triggered by four different functions:

- [PERIODIC COUNTDOWN TIMER INTERRUPT FUNCTION](#page-28-0)
- [PERIODIC TIME UPDATE INTERRUPT FUNCTION](#page-31-0)
- [ALARM INTERRUPT FUNCTION](#page-33-0)
- [EXTERNAL EVENT FUNCTION](#page-35-0)

Interrupt scheme:

(2) When bits TIE, UIE, AIE and EIE are disabled, pin $\overline{\text{INT}}$ will remain high-impedance.

4.4.1.SERVICING INTERRUPTS

The INT pin can indicate four types of interrupts. It outputs the logic OR operation result of these interrupt outputs. When an interrupt is detected (when $\overline{\text{INT}}$ pin produces a negative pulse or is at low level), the TF, UF, AF and EVF flags can be read to determine which interrupt event has occurred.

To keep INT pin from changing to low level, clear the TIE, UIE, AIE and EIE bits. To check whether an event has occurred without outputting any interrupts via the $\overline{\text{INT}}$ pin, software can read the TF, AF and EVF interrupt flags (polling).

Exception: The UF flag is only set to 1 by the periodic time update interrupt function when the UIE bit is set to 1.

4.5. PERIODIC COUNTDOWN TIMER INTERRUPT FUNCTION

The Periodic Countdown Timer Interrupt function generates an interrupt event periodically at any period set from 244.14 μs to 4095 minutes.

When an interrupt event is generated, the $\overline{\text{INT}}$ pin goes to the low level and the TF flag is set to 1 to indicate that an event has occurred. The output on the $\overline{\text{INT}}$ pin is only effective if the TIE bit in the Control Register is set to 1. The low-level output signal on the \overline{INT} pin is automatically cleared after the Auto reset time t_{RTN1}. t_{RTN1} = 122 µs (TD = 00) or t_{RTN1} = 7.813 ms (TD = 01, 10, 11).

4.5.1.PERIODIC COUNTDOWN TIMER DIAGRAM

Diagram of the Periodic Countdown Timer Interrupt function:

- 3 When a Periodic Countdown Timer Interrupt occurs, the TF bit is set to 1.
- $^{\circledR}$ The TF bit retains 1 until it is cleared to 0 by software.
- \circledS If the TIE bit is 1 and a Periodic Countdown Timer Interrupt occurs, the $\overline{\text{INT}}$ pin output goes low.
- 6 The $\overline{\text{INT}}$ pin output remains LOW during the Auto reset time t_{RTN1}, and then it is automatically cleared to 1. The TD field determines the Timer Clock Frequency and the Auto reset time t_{RTN1} . t_{RTN1} = 122 µs (TD = 00) or t_{RTN1} = 7.813 ms (TD = 01, 10, 11).
- $^\mathcal{D}$ When a 0 is written to the TE bit, the Periodic Countdown Timer function is stopped and the $\overline{\text{INT}}$ pin is cleared after the Auto reset time t_{RTN1} .
- 8 If the $\overline{\text{INT}}$ pin is LOW, its status does not change when the TF bit value is cleared to 0.
- $^{\circledR}$ If the INT pin is LOW, its status changes as soon as the TIE bit value is cleared to 0.

4.5.2.USE OF THE PERIODIC COUNTDOWN TIMER

The following registers, fields and bits are related to the Periodic Countdown Timer Interrupt function:

- Timer Counter 0 Register (0Bh, 1Bh) (see [PERIODIC COUNTDOWN TIMER CONTROL REGISTERS\)](#page-17-0)
- Timer Counter 1 Register (0Ch, 1Ch) (see [PERIODIC COUNTDOWN TIMER CONTROL REGISTERS\)](#page-17-0)
- TE bit and TD field (see [EXTENSION REGISTER,](#page-18-0) 0Dh, 1Dh)
- TF bit (see [FLAG REGISTER,](#page-19-0) 0Eh, 1Eh)
- TIE bit (see [CONTROL REGISTER,](#page-20-0) 0Fh, 1Fh)

Prior to entering any timer settings for the Periodic Countdown Timer Interrupt, it is recommended to write a 0 to the TIE and TE bits to prevent inadvertent interrupts on $\overline{\text{INT}}$ pin. When the RESET bit value is 1, the Periodic Countdown Timer Interrupt function event does not occur. When the Periodic Countdown Timer Interrupt function is not used, the 2 Bytes of the Timer Counter registers (0Bh, 1Bh and 0Ch, 1Ch) can be used as RAM bytes. The Timer Clock Frequency selection field TD is used to set the countdown period (source clock) for the Periodic Countdown Timer Interrupt function (four settings are possible).

Procedure to use the Periodic Countdown Timer Interrupt function:

- 1. Initialize bits TE, TIE and TF to 0. In that order, to prevent inadvertent interrupts on $\overline{\text{INT}}$ pin.
- 2. Choose the Timer Clock Frequency and write the corresponding value in the TD field.
- 3. Choose the Countdown Period based on the Timer Clock Frequency, and write the corresponding Timer Value to the registers Timer Counter 0 (0Bh, 1Bh) and Timer Counter 1 (0Ch, 1Ch). See following table.
- 4. Set the TIE bit to 1 if you want to get a hardware interrupt on $\overline{\text{INT}}$ pin.
- 5. Set the TE bit from 0 to 1 to start the Periodic Countdown Timer. The countdown starts at the rising edge of the SCL signal after Bit 0 of the Address D is transferred. See subsequent Figure that shows the start timing.

Countdown Period in seconds:

Timer Value
Countdown Period = Timer Cleak Fract Timer Clock Frequency

Countdown Period:

Start timing of the Periodic Countdown Timer:

4.5.3.FIRST PERIOD DURATION

When the TF flag is set, an interrupt signal on INT is generated if this mode is enabled. See Section INTERRUPT [OUTPUT](#page-26-0) for details on how the interrupt can be controlled.

When starting the timer for the first time, the first period has an uncertainty. The uncertainty is a result of the enable instruction being generated from the interface clock which is asynchronous from the Timer Clock Frequency. Subsequent timer periods do not have such deviation. The amount of deviation for the first timer period depends on the chosen Timer Clock Frequency, see following Table.

First period duration for Timer Value $n^{(1)}$:

At the end of every countdown, the timer sets the Periodic Countdown Timer Flag (bit TF in Flag Register). Bit TF can only be cleared by command. The asserted bit TF can be used to generate an interrupt at pin $\overline{\text{INT}}$.

When reading the Timer Value, the preset value is returned and not the actual value.

4.6. PERIODIC TIME UPDATE INTERRUPT FUNCTION

The Periodic Time Update Interrupt function generates an interrupt event periodically at the One-Second or the One-Minute update time, according to the selected timer source with bit USEL.

When an interrupt event is generated, the $\overline{\text{INT}}$ pin goes to the low level and the UF flag is set to 1 to indicate that an event has occurred. The output on the $\overline{\text{INT}}$ pin is only effective if the UIE bit in the Control Register is set to 1. The low-level output signal on the \overline{INT} pin is automatically cleared after the Auto reset time t_{RTN2}. t_{RTN2} = 500 ms (Second update) or t_{RTN2} = 15.6 ms (Minute update).

4.6.1.PERIODIC TIME UPDATE DIAGRAM

Diagram of the Periodic Time Update Interrupt function:

- the minute update time. The USEL bit determines whether it is the Second or the Minute period with the corresponding Auto reset time t_{RTN2} . t_{RTN2} = 500 ms (Second update) or t_{RTN2} = 15.6 ms (Minute update).
- 2 When a Periodic Time Update Interrupt occurs, the UF flag is set to 1.
- $^{\textrm{\tiny{\textregistered}}}$ The UF flag retains 1 until it is cleared to 0 by software.
- 4 If the UIE bit is 1 and a Periodic Time Update Interrupt occurs, the $\overline{\text{INT}}$ pin output goes low.
- 5 The INT pin output remains low during the Auto reset time t_{RTN2}, and then it is automatically cleared to 1.
- $^{\circledR}$ If the INT pin is low, its status does not change when the UF bit value is cleared to 0.
- $^{\mathcal{D}}$ If the $\overline{\text{INT}}$ pin is low, its status changes as soon as the UIE bit value is cleared to 0.
- $^{3)}$ When UIE bit is 0 and a Periodic Time Update Interrupt event occurs, the UF flag is not set and the $\overline{\text{INT}}$ pin output does not go low.

4.6.2.USE OF THE PERIODIC TIME UPDATE INTERRUPT

The following bits are related to the Periodic Time Update Interrupt function:

- USEL bit (see [EXTENSION REGISTER,](#page-18-0) 0Dh, 1Dh)
- UF bit (see [FLAG REGISTER,](#page-19-0) 0Eh, 1Eh)
- UIE bit (see [CONTROL REGISTER,](#page-20-0) 0Fh, 1Fh)

Prior to entering any other settings, it is recommended to write a 0 to the UIE bit to prevent inadvertent interrupts on $\overline{\text{INT}}$ pin. If the RESET bit is set to 1 (see [CONTROL REGISTER,](#page-20-0) 0Fh, 1Fh) the divider chain is reset and the Periodic Time Update Interrupt does not occur. The reset function only interrupts the Periodic Time Update Interrupt function but does not turn it off.

Procedure to use the Periodic Time Update Interrupt function:

- 1. Initialize bits UIE and UF to 0.
- 2. Choose the timer source clock and write the corresponding value in the USEL bit.
- 3. Set the UIE bit to 1 to enable the Periodic Time Update Interrupt function with hardware interrupt on INT pin.
- 4. The first interrupt will occur after the next event, either second or minute change.

4.7. ALARM INTERRUPT FUNCTION

The Alarm Interrupt function generates an interrupt for alarm settings such as date, weekday, hour or minute settings.

When an interrupt event is generated, the $\overline{\text{INT}}$ pin goes to the low level and the AF flag is set to 1 to indicate that an event has occurred.

4.7.1.ALARM DIAGRAM

Diagram of the Alarm Interrupt function:

- \circ A date, weekday, hour or minute alarm interrupt event occurs when the selected Alarm register match the respective counter. The WADA bit determines whether it is the date or weekday.
- 2 When an Alarm Interrupt event occurs, the AF bit value is set to 1.
- 3 The AF bit retains 1 until it is cleared to 0 by software.
- $^{\circledR}$ If the AIE bit is 1 and an Alarm Interrupt occurs, the $\overline{\text{INT}}$ pin output goes low.
- 5 If the AIE value is changed from 1 to 0 while the $\overline{\text{INT}}$ pin output is low, the $\overline{\text{INT}}$ pin immediately changes its status. While the AF bit value is 1, the $\overline{\text{INT}}$ status can be controlled by the AIE bit.
- 6 If the INT pin is low, its status changes as soon as the AF bit value is cleared from 1 to 0.
- $^\copyright$ If the AIE bit value is 0 when an Alarm Interrupt occurs, the $\overline{\text{INT}}$ pin status does not go low.

4.7.2.USE OF THE ALARM INTERRUPT

The following registers and bits are related to the Alarm Interrupt function:

- Minutes Register (01h, 12h) (see [CLOCK REGISTERS\)](#page-12-0)
- Hours Register (02h, 13h) (see [CLOCK REGISTERS\)](#page-12-0)
- Weekday Register (03h, 14h) (see [CALENDAR REGISTERS\)](#page-13-0)
- Date Register (04h, 15h) (see [CALENDAR REGISTERS\)](#page-13-0)
- Minutes Alarm Register and AE_M bit (08h, 18h) (see [ALARM REGISTERS\)](#page-15-0)
- Hours Alarm Register and AE_H bit (09h, 19h) (see [ALARM REGISTERS\)](#page-15-0)
- Weekday/Date Alarm Register and AE_WD bit (0Ah, 1Ah) (see [ALARM REGISTERS\)](#page-15-0)
- WADA bit (see [EXTENSION REGISTER,](#page-18-0) 0Dh, 1Dh)
- AF bit (see [FLAG REGISTER,](#page-19-0) 0Eh, 1Eh)
- AIE bit (see [CONTROL REGISTER,](#page-20-0) 0Fh, 1Fh)

Prior to entering any timer settings for the Alarm Interrupt, it is recommended to write a 0 to the AIE bit to prevent inadvertent interrupts on $\overline{\text{INT}}$ pin. When the RESET bit value is 1, the Alarm Interrupt function event does not occur. When the Alarm Interrupt function is not used, the 3 Bytes of the Alarm registers (08h, 18h; 09h, 19h and 0Ah, 1Ah) can be used as RAM bytes. In such case, be sure to write a 0 to the AIE bit (if the AIE bit value is 1 and the Alarm registers are used as RAM registers, $\overline{\text{INT}}$ may change to low level unintentionally).

Procedure to use the Alarm Interrupt function:

- 1. Initialize bits AIE and AF to 0.
- 2. Choose weekday alarm or date alarm (weekday/date) by setting the WADA bit. WADA = 0 for weekday alarm or $WADA = 1$ for date alarm.
- 3. Write the desired alarm settings in registers 08h, 18h to 0Ah, 1Ah. The three alarm enable bits, AE_M, AE_H and AE_WD, are used to select the corresponding register that has to be taken into account for match or not. See the following table.
- 4. Set the AIE bit to 1 if you want to get a hardware interrupt on $\overline{\text{INT}}$ pin.

Alarm Interrupt:

 (2) If all AE_x = 1: Alarm event every minute

4.8. EXTERNAL EVENT FUNCTION

The External Event Interrupt and Time Stamp function is enabled by the control bits EIE and ECP. Depending of the EHL bit a high level (positive edge) or low level (negative edge) signal can be regarded as an event and furthermore a digital glitch filtering is applied to the EVI signal when selecting a sampling period in the ET field. If enabled (EIE and ECP set to 1 and EVF flag was cleared to 0 before) and an External Event on EVI pin is detected, the seconds and 100th seconds are captured and copied into the Seconds CP and 100th Seconds CP registers, the $\overline{\text{INT}}$ is issued and the EVF flag is set to 1 to indicate that an external event has occurred.

External Event detection and Time Stamp function:

- (1) When ERST bit is 1 and in case of an External Event detection at the EVI pin the 100th Seconds Register is reset to 0. Moreover, the 100th Seconds CP and Seconds CP registers are also reset to 0, whatever the ECP value is. After the event detection, the ERST bit is reset to 0 automatically. When 1, the reset function may be cancelled when the ERST bit is set back to 0 before an event occurs.
- (2) When RESET bit is 1 the 100th Seconds register is reset to 00 and the Seconds register value remains unchanged (1 Hz clock is stopped).
- (3) When reading the CP registers (Time Stamp), the consistent values at 1^2C START are returned. When a subsequent event occurs during read-out, the new values are not lost and can be read in a later I^2C access.
4.8.1.EXTERNAL EVENT DIAGRAM

Diagram of the External Event function.

Example with positive edge/level detection (EHL = 1) and without event reset function (ERST = 0):

- 5 The EVF flag retains 1 until it is cleared to 0 by software.
- $₆$ </sub> No interrupt occurs on $\overline{\text{INT}}$ pin because the EVF flag was not set back to 0. But, new value (A+2) is captured in the CP registers.
- \circledv If the $\overline{\text{INT}}$ pin is low, its status changes as soon as the EVF flag is cleared to 0, even if EVI input is high level.
- 8 If ECP is set to 0, no capture occurs.
- $\circled{9}$ If the EVI input is 1 (steady state) and the ECP bit is set from 0 to 1, no capture is done.
- 10 While the EVF bit value is 1, the INT status can be controlled by the EIE bit.

4.8.2.USE OF THE EXTERNAL EVENT FUNCTION

The following registers and bits are related to the External Event Interrupt and Time Stamp function:

- 100th Seconds Register (10h) (see [CLOCK REGISTERS\)](#page-12-0)
- Seconds Register (00h, 11h) (see [CLOCK REGISTERS\)](#page-12-0)
- 100th Seconds CP Register (20h) (see [CLOCK REGISTERS\)](#page-12-0)
- Seconds CP Register (21h) (see [CLOCK REGISTERS\)](#page-12-0)
- ECP bit, EHL bit, ET field and ERST bit (see [CAPTURE BUFFER/EVENT CONTROL REGISTERS,](#page-22-0) 2Fh)
- EVF bit (see [FLAG REGISTER,](#page-19-0) 0Eh, 1Eh)
- EIE bit (see [CONTROL REGISTER,](#page-20-0) 0Fh, 1Fh)

Prior to entering any timer settings for the event interrupt, it is recommended to write a 0 to the EIE bit to prevent inadvertent interrupts on $\overline{\text{INT}}$ pin.

Procedure to use the External Event Interrupt and Time Stamp function:

-
- 1. Initialize bits EIE and EVF to 0.
2. Set the ECP bit to 1 if you want Set the ECP bit to 1 if you want to capture the seconds and $100th$ seconds.
- 3. Set the EHL bit to 1 or 0 to choose high or low level detection on pin EVI
- 4. Set the ET field to apply filtering to the EVI pin. See following two diagrams.
- 5. Set the ERST bit to 1 if you want to reset the 100th Seconds, Seconds CP and 100th Seconds CP registers to 0 in case of an event detection. After the event detection, the ERST bit is reset to 0.
- 6. Set the EIE bit to 1 if you want to get a hardware interrupt on $\overline{\text{INT}}$ pin.

 $ET = 00$. Example with positive edge detection:

With digital debounce filter: ET = 01, 10 or 11 (sampling period t_{SP} = 3.9 ms, 15.6 ms or 125 ms). Example with positive edge/level detection:

4.9. CLKOUT FREQUENCY SELECTION

A programmable square wave is available at pin CLKOUT. Operation is controlled by the FD field in the [EXTENSION REGISTER.](#page-18-0) Frequencies of 32.768 kHz (default), 1024 Hz and 1 Hz can be generated for use as a system clock, microcontroller clock, input to a charge pump, or for calibration of the crystal oscillator.

Pin CLKOUT is a push-pull output that is enabled at power-on (when CLKOE is HIGH). CLKOUT can be disabled by setting CLKOE pin LOW. When disabled, the CLKOUT pin is high impedance (tri-state).

The RESET bit function can affect the CLKOUT signal depending on the selected frequency. When the RESET bit is set logic 1 and the CLKOE pin is HIGH, the CLKOUT pin generates a continuous HIGH or LOW for the 1024 Hz and 1 Hz frequency (for more details, see [RESET BIT FUNCTION\)](#page-42-0).

For this table, CLKOE is HIGH.

see [TEMPERATURE COMPENSATION](#page-43-0) and AGING (2) 1024 Hz and 1 Hz are synchronously turned on and off by the RESET bit.

See also [32.768 KHZ ENABLE/DISABLE TIMING.](#page-57-0)

4.10.DIGITAL ARCHITECTURE SUMMARY

The following Figure illustrates the overall architecture of the pin inputs and outputs of the RV-8803-C7.

Digital Architecture Summary:

4.11.SYNCHRONICITY BETWEEN INT SIGNALS AND 1 HZ CLKOUT

The following Figure illustrates the synchronicity between the INT signals from the Periodic Time Update Interrupt function and Periodic Countdown Timer Interrupt function to the 1 Hz CLKOUT signal.

Synchronicity between the $\overline{\text{INT}}$ signals and the 1 Hz CLKOUT:

4.12.TIME DATA READ-OUT

In order to not corrupt the accuracy of the temperature compensation and the Time Capture (Time Stamp) function on the highest 100th Seconds resolution, it is not possible to freeze the clock and calendar register during read-out process, as it is common practice for other RTC's.

Since the time and calendar registers cannot be frozen, there might be a condition that the time registers are incremented by one 1 Hz tick while read-out. Therefore, reading should be completed within one second and to avoid corrupted (partially incremented) data, special measures and procedures need to be applied.

4.12.1. PROCEDURE

If a time read-out sequence starts at the end of a minute there is a special condition that subsequent registers might be incremented by the time update.

Example with faulty reading:

- Expected time read-out = $mm:ss = 01:59$
	- 1. mm:ss = $01:59$ read Seconds = 59
		- mm:ss = 02:00 one 1 Hz tick incremented
	- 2. mm:ss = $02:00$ read Minutes = 02

Effective faulty time read-out = $mm:ss = 02:59$; the failure is 1 minute.

To prevent using corrupted data from partially incremented time and calendar registers, it is recommended to repeat and confirm time and calendar data when reading Seconds = 59 (see following [METHODE TO CONFIRM](#page-41-0) [CORRECT TIME AND CALENDAR READ-OUT\)](#page-41-0).

4.12.2. METHODE TO CONFIRM CORRECT TIME AND CALENDAR READ-OUT

When reading Seconds = 59, it is recommended to repeat and compare the read-out of the Seconds register. If the Seconds register data matches, it confirms that the time and calendar data are valid (no time increment occurred during data read-out). If the Seconds value has changed to 00, the second set of time and calendar data is valid.

- 1. Read required time and calendar information.
- 2. If Seconds data = 59, a repeated reading is required.
- 3. If Seconds data is again 59 seconds, then the first data from the first reading is confirmed to be valid.
- 4. If the Seconds register was incremented (not 59 seconds anymore), then the time and calendar information has been incremented and the second set of data is confirmed to be valid (the first set of data is supposed to be partially incremented during the read-out sequence and therefore is invalid).

4.13.RESET BIT FUNCTION

The RESET bit is used for a software-based accurate and safe starting of the time circuits.

The RESET bit function causes the upper part of the prescaler (F_2 to F_{14}) to be held in reset and thus no 1 Hz ticks are generated. The RESET bit function will not affect the CLKOUT of 32.768 kHz (see also [CLKOUT FREQUENCY](#page-38-0) [SELECTION\)](#page-38-0).

RESET bit functional diagram:

The time circuits can then be set and do not increment until the RESET bit is released.

Setting the clock and calendar values using the RESET bit function:

- 1. Set RESET bit to 1 to prevent a timer update while setting the time.
- 2. Write the desired clock and calendar values to the registers (year, month, date, weekday, hours, minutes and seconds). The 100th seconds register was cleared to 00 when setting the RESET bit to 1.
- 3. Release RESET bit to 0 to start the time circuits.

4.14.ERST BIT FUNCTION

The Event Reset bit ERST is used for an external event triggered highly accurate time adjustment (synchronizing).

If the ERST bit is 1 and in case of an External Event detection on the EVI pin, the prescaler and counters at below the second are reset to 0 (2 Hz to 8 kHz). This means that the 100th Seconds Register (100 Hz) is reset to 0. Moreover, the time stamp in the 100th Seconds CP and Seconds CP registers are also reset to 0, whatever the ECP value is. After the event detection, the ERST bit is reset to 0 automatically.

Setting the clock and calendar values synchronous to an External Event detection:

- 1. Initialize the External Event Function according to [USE OF THE EXTERNAL EVENT FUNCTION](#page-37-0) with bits EIE and ERST set to 1.
- 2. When interrupt pin $\overline{\text{INT}}$ is triggered by the External Event Function, write the desired clock and calendar values to the registers (year, month, date, weekday, hours, minutes and seconds). The 100th Seconds register is cleared to 00 automatically.
- 3. After the event detection, the ERST bit is reset to 0 automatically.

See also [EXTERNAL EVENT FUNCTION.](#page-35-0)

5. TEMPERATURE COMPENSATION

5.1. FREQUENCIES

Xtal 32.768 kHz

The Xtal 32.768 kHz clock is not temperature compensated. Due to its negative temperature coefficient with a parabolic frequency deviation, a change of up to -150 ppm across the standard operating temperature range of -40°C to +85°C can result (for the extended operating range of -40°C to 105°C a frequency change of -225 ppm can result). The 32.768 kHz oscillator frequency on all devices is tested not to exceed a time deviation of ± 20 ppm (parts per million) at 25°C.

Frequencies from 4096 Hz to 64 Hz

These frequencies are digitally temperature compensated with a Time Accuracy of ± 3 ppm over the standard temperature range of -40°C to +85°C and of ±7 ppm for the extended temperature range from +85°C to 105°C. The clock at the 16.384 kHz level of the divider chain is modified by adding or subtracting 32.768 kHz level pulses. The pulses are added or subtracted according to the expected frequency deviation computed by the temperature compensation algorithm. The digital compensation method (adding and subtracting clock pulses) is affecting the cycle-to-cycle jitter of the digitally compensated frequencies shown below.

- 4096 Hz (Periodic Countdown Timer Interrupt)
- 1024 Hz (CLKOUT)
- 100 Hz (External Event Interrupt)
- 64 Hz (Periodic Countdown Timer Interrupt)

Aging compensation can be done with the OFFSET value (see [AGING CORRECTION\)](#page-44-0).

1 Hz and Clock / Calendar

The 1 Hz clock is temperature compensated and using both, digital coarse compensation and digital fine adjustment. The Time Accuracy and the Frequency Accuracy is ±3 ppm for every 1 Hz period over the standard temperature range of -40°C to +85°C and ±7 ppm for the extended temperature range from +85°C to +105°C. The temperature compensation algorithm adjusts every 1 Hz period with a resolution of about 0.1 ppm. This precise and accurate 1 Hz clock is used to increment all subsequent clock and calendar registers. Aging compensation can be done with the OFFSET value (see [AGING CORRECTION\)](#page-44-0).

5.2. TIME ACCURACY VS. TEMPERATURE CHARACTERISTICS

5.3. COMPENSATION VALUES

Each device is factory calibrated over the full temperature range, and the individual compensation values are stored in the EEPROM of the Digital Temperature Compensation Unit (DTCU). The EEPROM is not accessible for the user.

5.4. AGING CORRECTION

An aging adjustment or accuracy tuning can be done with the OFFSET value. The correction is purely digitally and has only the effect of shifting the time vs. temperature curve vertically up or down. It has no effect on the time vs. temperature characteristics of the final frequency. The OFFSET value contains a two's complement number with a range of -2⁶ to +2⁶-1 adjustment steps. The minimal correction step (one LSB) is ±1/(32768*128) = ±0.2384 ppm. The maximum correction range is roughly ±7.4 ppm. Note that the signed offset value OFFSET corresponds to the actual offset value of the measured frequency. The user has access to this field (see [OFFSET REGISTER\)](#page-21-0).

The OFFSET value is determined by the following process:

- 1. Set the OFFSET field to 0 to ensure correction is not occurring.
- 2. Select the 1 Hz frequency on the CLKOUT pin.
- 3. Measure the frequency Fmeas at the output pin in Hz.
- 4. Compute the offset value required in ppm: $POffset = ((Fmeas 1)*1'000'000)$
- 5. Compute the offset value in steps: Offset = $POffset(1/(32768*128))$ = $POffset(0.2384)$
- 6. If Offset > 31, the frequency is too high to be corrected.
- 7. Else if 0 ≤ Offset ≤ 31, set OFFSET = Offset
- 8. Else if $-32 \leq$ Offset ≤ -1 , set OFFSET = Offset + 64
- 9. Else the frequency is too low to be corrected.

Examples:

- If 1.0000012 Hz is measured when the 1 Hz clock is selected, the offset is +0.0000012 Hz, which is +0.0000012 Hz / 10⁻⁶ Hz = +1.2 ppm. The positive offset value is then calculated as follows: +1.2 ppm / 0.2384 ppm = $+5.03$, the rounded integral part is $+5$. In binary, OFFSET = 000101.
- If 0.9999949 Hz is measured when the 1 Hz clock is selected, the offset is -0.0000051 Hz, which is -0.0000051 Hz / 10⁻⁶ Hz = -5.1 ppm. The negative offset value is then calculated as follows: -5.1 ppm / 0.2384 ppm = -21.39 , the rounded integral part is -21 . The unsigned value is then $-21 +64 = +43$. In binary, OFFSET = 101011.

5.5. CLOCKING SCHEME

Clocking Scheme with CLKOUT and Interrupts:

5.6. MEASURING TIME ACCURACY AT CLKOUT PIN

The simplest method to verify the time accuracy of the Digital Temperature Compensation Unit (DTCU) is to measure the compensated 1 Hz frequency at the CLKOUT pin. The 1 Hz clock frequency contains digitally temperature compensation clocks with digital fine adjustment and represents the fully time accuracy of the device.

5.6.1.MEASURING 1 HZ AT CLKOUT PIN

- 1. Select the 1 Hz frequency at CLKOUT:
	- a. Set the FD field to 10 = 1 Hz (see [EXTENSION REGISTER,](#page-18-0) 0Dh, 1Dh).
	- b. Set the CLKOUT pin into output mode by setting the CLKOE pin to high level.
- 2. Measuring equipment and setup:
	- a. Use a high-precision universal counter to observe the 1 Hz frequency accuracy on CLKOUT pin.
	- b. Trigger on the rising edge of the hybrid signal (gate time ≥ 1 second). Each 1 Hz clock measured at the rising edge fully representing the accuracy of the DTCU.

1 Hz time accuracy at CLKOUT pin (hybrid signal):

5.7. MEASURING TIME ACCURACY AT INT PIN

The Periodic Time Update Interrupt function can also be used to verify the time accuracy of the Digital Temperature Compensation Unit (DTCU) by measuring the compensated 1 Hz frequency at the $\overline{\text{INT}}$ output pin. However this procedure is a little more sophisticated than using the method with the CLKOUT pin.

5.7.1.MEASURING 1 HZ WITH THE PERIODIC TIME UPDATE INTERRUPT FUNCTION

- 1. Select the Periodic Time Update Interrupt function with the frequency 1 Hz at the $\overline{\text{INT}}$ output pin:
	- a. Write 0 to UIE and UF bits
	- b. Choose USEL = $0 = 1$ Hz, $t_{RTN2} = 500$ ms (Default value) (see [EXTENSION REGISTER,](#page-18-0) 0Dh, 1Dh)
	- c. Set UIE bit to 1 to enable the Periodic Time Update Interrupt function with hardware interrupt on $\overline{\text{INT}}$ pin.
	- d. The first interrupt will occur after the next event.
- 2. Measuring equipment and setup:
	- a. Use a high-precision universal counter to observe the frequency stability on $\overline{\text{INT}}$ output pin.
	- b. If measuring the 1 Hz clock it suffices to measure only one period to verify the time accuracy. Trigger on the falling edge of the hybrid signal (gate time ≥ 1 second).

1 Hz time accuracy at $\overline{\text{INT}}$ pin with the Periodic Time Update Interrupt function (hybrid signal):

6. I ²C INTERFACE

The I²C interface is for bidirectional, two-line communication between different ICs or modules. The RV-8803-C7 is accessed at addresses 64h/65h, and supports Fast Mode (up to 400 kHz). The I^2C interface consists of two lines: one bi-directional data line (SDA) and one clock line (SCL). Both lines are connected to a positive supply via pullup resistors. Data transfer is initiated only when the interface is not busy.

6.1. BIT TRANSFER

One data bit is transferred during each clock pulse. The data on the SDA line remains stable during the HIGH period of the clock pulse, as changes in the data line at this time are interpreted as a control signals. Data changes should be executed during the LOW period of the clock pulse (see Figure below).

Bit transfer:

6.2. START AND STOP CONDITIONS

Both data and clock lines remain HIGH when the bus is not busy. A HIGH-to-LOW transition of the data line, while the clock is HIGH, is defined as the START condition (S). A LOW-to-HIGH transition of the data line, while the clock is HIGH, is defined as the STOP condition (P) (see Figure below).

Definition of START and STOP conditions:

A START condition which occurs after a previous START but before a STOP is called a Repeated START condition, and functions exactly like a normal STOP followed by a normal START.

6.3. DATA VALID

After a START condition, SDA is stable for the duration of the high period of SCL. The data on SDA may be changed during the low period of SCL. There is one clock pulse per bit of data. Each data transfer is initiated with a START condition and terminated with a STOP condition. The number of data bytes transferred between the START and STOP conditions is not limited. The information is transmitted byte-wise and each receiver acknowledges with a ninth bit.

In order to not corrupt the accuracy of the temperature compensation and the Time Capture (Time Stamp) function on the highest 100th Seconds resolution, it is not possible to freeze the clock and calendar register during read-out process, as it is common practice for other RTC's.

Since the time and calendar registers cannot be frozen, there might be a condition that the time registers are incremented while read-out. To avoid reading corrupted (partially incremented) data, special measures and procedures need to be applied (see [TIME DATA READ-OUT\)](#page-41-1).

6.4. SYSTEM CONFIGURATION

Since multiple devices can be connected with the I^2C -bus, all I^2C -bus devices have a fixed and unique device number built-in to allow individual addressing of each device.

The device that controls the I^2C -bus is the Master; the devices which are controlled by the Master are the Slaves. A device generating a message is a Transmitter; a device receiving a message is the Receiver. The RV-8803-C7 acts as a Slave-Receiver or Slave-Transmitter.

Before any data is transmitted on the I^2C -bus, the device which should respond is addressed first. The addressing is always carried out with the first byte transmitted after the START procedure. The clock signal SCL is only an input signal, but the data signal SDA is a bidirectional line.

6.5. ACKNOWLEDGE

The number of data bytes transferred between the START and STOP conditions from transmitter to receiver is unlimited. Each byte of eight bits is followed by an acknowledge cycle.

- A slave receiver, which is addressed, must generate an acknowledge cycle after the reception of each byte.
- Also a master receiver must generate an acknowledge cycle after the reception of each byte that has been clocked out of the slave transmitter.
- The device that acknowledges must pull-down the SDA line during the acknowledge clock pulse, so that the SDA line is stable LOW during the HIGH period of the acknowledge-related clock pulse (set-up and hold times must be taken into consideration).
- A master receiver must signal an end of data to the transmitter by not generating an acknowledge cycle on the last byte that has been clocked out of the slave. In this event the transmitter must leave the data line HIGH to enable the master to generate a STOP condition.

Data transfer and acknowledge on the I^2C -bus:

6.6. SLAVE ADDRESS

On the I²C-bus the 7-bit slave address 0110010b is reserved for the RV-8803-C7. The entire I²C-bus slave address byte is shown in the following table.

After a START condition, the I²C slave address has to be sent to the RV-8803-C7 device. The R/W bit defines the direction of the following single or multiple byte data transfer. The 7-bit address is transmitted MSB first. If this address is 0110010b, the RV-8803-C7 is selected, the eighth bit indicates a read (R/W = 1) or a write (R/W = 0) operation (results in 65h or 64h) and the RV-8803-C7 supplies the ACK. The RV-8803-C7 ignores all other address values and does not respond with an ACK.

In the write operation, a data transfer is terminated by sending either the STOP condition or the START condition of the next data transfer.

6.7. WRITE OPERATION

Master transmits to Slave-Receiver at specified address. The Register Address is an 8-bit value that defines which register is to be accessed next. After writing one byte, the Register Address is automatically incremented by 1.

Master writes to slave RV-8803-C7 at specific address:

- 1) Master sends out the START condition.
- 2) Master sends out Slave Address, 64h for the RV-8803-C7; the R/W bit is a 0 indicating a write operation.
- 3) Acknowledgement from RV-8803-C7.
- 4) Master sends out the Register Address to RV-8803-C7.
- 5) Acknowledgement from RV-8803-C7.
- 6) Master sends out the Data to write to the specified address in step 4).
- 7) Acknowledgement from RV-8803-C7.
- 8) Steps 6) and 7) can be repeated if necessary. The address is automatically incremented in the RV-8803-C7.
-
- 9) Master sends out the STOP Condition.

6.8. READ OPERATION AT SPECIFIC ADDRESS

Master reads data from slave RV-8803-C7 at specific address:

- 1) Master sends out the START condition.
- 2) Master sends out Slave Address, 64h for the RV-8803-C7; the R/W bit is a 0 indicating a write operation.
- 3) Acknowledgement from RV-8803-C7.
- 4) Master sends out the Register Address to RV-8803-C7.
- 5) Acknowledgement from RV-8803-C7.
- 6) Master sends out the Repeated START condition (or STOP condition followed by START condition)
- 7) Master sends out Slave Address, 65h for the RV-8803-C7; the R/W bit is a 1 indicating a read operation.
- 8) Acknowledgement from RV-8803-C7.
- At this point, the Master becomes a Receiver and the Slave becomes the Transmitter.
- 9) The Slave sends out the Data from the Register Address specified in step 4).
- 10) Acknowledgement from Master.
- 11) Steps 9) and 10) can be repeated if necessary.
- The address is automatically incremented in the RV-8803-C7.
- 12) The Master, addressed as Receiver, can stop data transmission by not generating an acknowledge on the last byte that has been sent from the Slave-Transmitter. In this event, the Slave-Transmitter must leave the data line HIGH to enable the Master to generate a STOP condition.
- 13) Master sends out the STOP condition.

6.9. READ OPERATION

Master reads data from slave RV-8803-C7 immediately after first byte:

- 1) Master sends out the START condition.
- 2) Master sends out Slave Address, 65h for the RV-8803-C7; the R/W bit is a 1 indicating a read operation.
- 3) Acknowledgement from RV-8803-C7.
- At this point, the Master becomes a Receiver and the Slave becomes the Transmitter.
- 4) The RV-8803-C7 sends out the Data from the last accessed Register Address incremented by 1.
- 5) Acknowledgement from Master.
- 6) Steps 4) and 5) can be repeated if necessary.
- The address is automatically incremented in the RV-8803-C7.
- 7) The Master, addressed as Receiver, can stop data transmission by not generating an acknowledge on the last byte that has been sent from the Slave-Transmitter. In this event, the Slave-Transmitter must leave the data line HIGH to enable the Master to generate a STOP condition.
- 8) Master sends out the STOP condition.

FREE-CLOCKING I² 6.10. C-BUS

When the I^2C Master goes through a reset (e.g. power down) right in the middle of transmitting or receiving a byte from the RV-8803-C7, the RV-8803-C7 is not aware of the reset and since the RV-8803-C7 does not have a timeout function it may well wait for the next clock event to send or receive a bit, not listening to any start condition which is likely to occur after a reset as startup sequence. When RV-8803-C7 holds SDA low the I²C-bus is blocked. Now it is the master's job to recover the bus and restore control to the main program. If the data line (SDA) is stuck LOW, the master has to clear it.

The following procedure is recommended:

- 1. Master tries to assert a Logic 1 on the SDA line
- 2. Master still sees a Logic 0 and then generates a clock pulse on SCL 0-1-0 (LOW-HIGH-LOW)
- 3. Master examines SDA: If $SDA = 0$, go to Step 2; this loop may be required up to 9 times. If $SDA = 1$, go to Step 4
- 4. Generate a STOP condition

7. ELECTRICAL SPECIFICATIONS

7.1. ABSOLUTE MAXIMUM RATINGS

The following Table lists the absolute maximum ratings.

Absolute Maximum Ratings according to IEC 60134:

MM: Machine Model, according to JESD22-A115.

(3) Latch-up testing, according to JESD78., Class I (room temperature), level A (100 mA).

(4) Supports extended operating temperature range from +85°C to +105°C with limitations.

7.2. OPERATING PARAMETERS

For this Table, TA = -40°C to +85°C unless otherwise indicated. V_{DD} = 1.5 to 5.5 V, fosc = 32.768 kHz, TYP values at 25°C and 3.0 V.

Operating Parameters:

 (1) See also V_{DD} Backup and recovery AC Electrical Characteristics and Parameters in sectio[n BACKUP AND RECOVERY.](#page-61-0)

⁽²⁾ Clocks operating and RAM and registers retained. Including temperature sensing and compensation.

(3) CLKOUT is held LOW during power on delay t_{POR1} and is HIGH during the power on reset duration t_{POR2}.

Note that between V_{LOW1} (1.2 V) and $V_{DD MIN}$ (1.5 V), time accuracy specification limits are not guaranteed.

 $^{(4)}$ All inputs and outputs are at 0 V or V_{DD} .

(5) 2.2k pull-up resistors on SCL/SDA, excluding external peripherals and pull-up resistor current. All other inputs (besides SDA and SCL) are at 0 V or VDD. Test conditions: Continuous burst read/write, 55h data pattern, 25 μs between each data byte, 20 pF load on each bus pin. (6) When CLKOUT is enabled (CLKOE is HIGH) the additional V_{DD} supply current $ΔI_{VDD}$ can be calculated as follows:

 $ΔI_{VDD} = C_L x V_{DD} x f_{OUT}, e.g. ΔI_{VDD} = 10 pF x 3.0 V x 32'768 Hz = 980 nA ≈ 1 μA$

For this Table, TA = -40°C to +85°C unless otherwise indicated. V_{DD} = 1.5 to 5.5 V, fosc= 32.768 kHz, TYP values at 25°C and 3.0 V.

Operating Parameters (continued):

7.2.1.TEMPERATURE COMPENSATION AND CURRENT CONSUMPTION

7.2.2.32.768 KHZ ENABLE/DISABLE TIMING

32.768 kHz CLKOUT enable and disable times. The 32.768 kHz CLKOUT is enabled synchronously and disabled asynchronously:

Hint: The other CLKOUT frequencies, 1024 Hz and 1 Hz, are asynchronously enabled and disabled.

7.3. OSCILLATOR PARAMETERS

For this Table, TA = -40°C to +85°C unless otherwise indicated. V_{DD} = 1.5 to 5.5 V, fosc = 32.768 kHz, TYP values at 25°C and 3.0 V.

Oscillator Parameters:

See also TIME ACCURACY [VS. TEMPERATURE CHARACTERISTICS.](#page-43-1)

7.3.1.TIME ACCURACY 1 HZ EXAMPLES

Time accuracy of the temperature compensated 1 Hz signal, $TA = -40^{\circ}C$ to $+85^{\circ}C$:

Time accuracy of the temperature compensated 1 Hz signal, including extended range from +85°C to +105°C:

7.4. POWER ON AC ELECTRICAL CHARACTERISTICS

The following Figure and table describe the power on AC electrical characteristics for the CLKOUT pin.

Power On AC Electrical Characteristics:

For this Table, TA = -40°C to +85°C unless otherwise indicated. V_{DD} = 1.5 to 5.5 V, TYP values at 25°C and 3.0 V.

SYMBOL PARAMETER CONDITIONS MIN TYP MAX UNIT V_{DDR1} V_{DD} rising slew rate at initial power on reset (POR) $CLKOE = V_{DD}$ 0.1 | | 1 | V/ms t_{port} t_{hor c} research power on delay the second the second of the second the second the second the second that the second the second the second that the second the second the second that the second the second the secon t_{POR2} Power on reset duration 80 500 ms V_{DDR1:EXT} V_{DD} rising slew rate at initial power on reset (POR) for extended temperature range $V_{DD} = 1.6$ to 5.5 V, $CLKOE = V_{DD}$ $T_A = +85 °C$ to $+105 °C$ 0.1 | | 1 | V/ms tPOR1:EXT Power on delay Power on delay
for extended temperature range $T_A = +85 \degree C$ to +105 $\degree C$ tPOR2:EXT Power on reset duration Fower on reset duration
for extended temperature range $\begin{vmatrix} 1 & 1 & 1 \\ 0 & 1 & 1 \\ 0 & 0 & 1 \end{vmatrix}$ 500 ms

Power On AC Electrical Parameters:

7.5. BACKUP AND RECOVERY

During a backup event with a backup voltage V_{DD} higher than V_{LOW1} (V_{LOW2}) the CLKOUT function is operating including the Temperature compensation and the RAM and registers are retained. Pay attention to the CLKOUT function if the power supply voltage V_{DD} of the RV-8803-C7 sharply goes up and down, meaning V_{DD} is changing between Main power voltage and Backup capacitor voltage. The CLKOUT signal can then disappear for several milliseconds when the voltage change is to sharp.

- 1. Choose a valid V_{DD} range for the CLKOUT function. E.g. 1.6 V to 3.6 V (see [OPERATING PARAMETERS\)](#page-55-0).
- 2. Ensure that the slew rates V_{DDF} and V_{DDR2} fulfill their specifications.
- 3. Check if these required specifications are fulfilled on your system.

The following Figure and Table describe the backup and recovery AC electrical characteristics (valid example with a backup voltage > V_{LOW1} (V_{LOW2})).

V_{DD} Backup and recovery AC Electrical Characteristics:

For this Table, $Ta = -40^{\circ}C$ to $+85^{\circ}C$ unless otherwise indicated.

I 2 7.6. C-BUS CHARACTERISTICS

The following Figure and Table describe the I²C AC electrical parameters.

²C AC Parameter Definitions:

For the following Table, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, TYP values at 25 $^{\circ}C$.

I²C AC Electrical Parameters:

I ²C-bus access:

For devices with production date code 853 and earlier it is recommended before going to idle mode to complete the I²C-bus access always with a Read Operation followed by the STOP condition (see [READ OPERATION\)](#page-53-0).

8. TYPICAL APPLICATION CIRCUIT

8.1. OPERATING RV-8803-C7 WITH BACKUP CAPACITOR

9. PACKAGE

9.1. DIMENSIONS AND SOLDER PAD LAYOUT

All dimensions in mm typical.

9.1.1.RECOMMENDED THERMAL RELIEF

When connecting a pad to a copper plane, thermal relief is recommended.

9.2. MARKING AND PIN #1 INDEX

10.MATERIAL COMPOSITION DECLARATION & ENVIRONMENTAL INFORMATION

10.1.HOMOGENOUS MATERIAL COMPOSITION DECLARATION

Homogenous material information according to IPC-1752 standard

10.2.MATERIAL ANALYSIS & TEST RESULTS

Homogenous material information according to IPC-1752 standard

| No. | Item Component | Sub Item Material | RoHS | | | | | | Halogen | | | | Phthalates | | | |
|----------------|---------------------------------|---------------------------------------|-------------|----|----|---------|-----|--------|---------|----------------|----|--------|-------------------|-----|-------|--------------|
| | Name | Name | 운 | රි | 운 | $C + 6$ | BBB | PBDE | Щ | \overline{O} | 运 | | BBP | DBP | DEHP | PINIC |
| 1 | Resonator | Quartz Crystal | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd |
| $\mathbf{2}$ | Electrodes | Cr+Au | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd |
| 3 | Housing | Ceramic | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd |
| 4 | Metal Lid | Kovar Lid & Plating | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd |
| 5 | Seal | Solder Preform | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd |
| 6 | Terminations | Int. & ext. terminals | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd |
| $\overline{7}$ | Conductive adhesive | Silver filled Silicone glue | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd |
| 8 | CMOS IC | Silicon & Gold bumps | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd | nd |
| | MDL | Measurement Detection Limit | 2 ppm | | | 5 ppm | | 50 ppm | | | | 0.003% | | | 0.01% | |

Test methods:

RoHS Test method with reference to IEC 62321-5: 2013 MDL: 2 ppm (PBB / PBDE: 5 ppm)
 Halogen Test method with reference to BS EN 14582:2007 MDL: 50 ppm **Halogen** Test method with reference to BS EN 14582:2007 MDL: 50 ppm
Phthalates Test method with reference to EN 14372 MDL: 0.003 % (DINP 0.01%) Test method with reference to EN 14372

 $nd = not detectable$

10.3.RECYCLING MATERIAL INFORMATION

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10.4.ENVIRONMENTAL PROPERTIES & ABSOLUTE MAXIMUM RATINGS

Terminal finish:

11.SOLDERING INFORMATION

12.HANDLING PRECAUTIONS FOR MODULES WITH EMBEDDED CRYSTALS

The built-in tuning-fork crystal consists of pure Silicon Dioxide in crystalline form. The cavity inside the package is evacuated and hermetically sealed in order for the crystal blank to function undisturbed from air molecules, humidity and other influences.

Shock and vibration:

Keep the crystal / module from being exposed to **excessive mechanical shock and vibration**. Micro Crystal guarantees that the crystal / module will bear a mechanical shock of 5000 g / 0.3 ms.

The following special situations may generate either shock or vibration:

Multiple PCB panels - Usually at the end of the pick & place process the single PCBs are cut out with a router. These machines sometimes generate vibrations on the PCB that have a fundamental or harmonic frequency close to 32.768 kHz. This might cause breakage of crystal blanks due to resonance. Router speed should be adiusted to avoid resonant vibration.

Ultrasonic cleaning - Avoid cleaning processes using ultrasonic energy. These processes can damage crystals due to mechanical resonance of the crystal blank.

Overheating, rework high temperature exposure:

Avoid overheating the package. The package is sealed with a seal ring consisting of 80% Gold and 20% Tin. The eutectic melting temperature of this alloy is at 280°C. Heating the seal ring up to >280°C will cause melting of the metal seal which then, due to the vacuum, is sucked into the cavity forming an air duct. This happens when using hot-air-gun set at temperatures >300°C.

Use the following methods for rework:

- Use a hot-air- gun set at 270°C.
- Use 2 temperature controlled soldering irons, set at 270°C, with special-tips to contact all solder-joints from both sides of the package at the same time, remove part with tweezers when pad solder is liquid.
13.PACKING & SHIPPING INFORMATION

Peel Method:

Medial section removal, both lateral stripes remain on carrier