

EFR32FG12 Gecko Proprietary Protocol SoC Family Data Sheet

The Gecko proprietary protocol family of SoCs is part of the Wireless Gecko portfolio. Gecko SoCs are ideal for enabling energyfriendly proprietary protocol networking for IoT devices.

The single-die solution provides industry-leading energy efficiency, ultra-fast wakeup times, a scalable power amplifier, an integrated balun and no-compromise MCU features.

Gecko applications include:

- Home and Building Automation and Security
- Metering
- Electronic Shelf Labels
- Industrial Automation
- Commercial and Retail Lighting and Sensing

KEY FEATURES

- 32-bit ARM® Cortex®-M4 core with 40 MHz maximum operating frequency
- Up to 1 MB of flash and 256 kB of RAM
- Pin-compatible across EFR32FG families (exceptions apply for 5V-tolerant pins)
- 12-channel Peripheral Reflex System, Low-Energy Sensor Interface & Multichannel Capacitive Sense Interface
- Autonomous Hardware Crypto Accelerator and True Random Number Generator
- Integrated PA with up to 19 dBm (2.4 GHz) or 20 dBm (Sub-GHz) TX power
- Integrated balun for 2.4 GHz
- Robust peripheral set and up to 65 GPIO

1. Feature List

The EFR32FG12 highlighted features are listed below.

- **Low Power Wireless System-on-Chip**
	- High Performance 32-bit 40 MHz ARM Cortex[®]-M4 with DSP instruction and floating-point unit for efficient signal processing
	- Embedded Trace Macrocell (ETM) for advanced debugging
	- Up to 1024 kB flash program memory
	- Up to 256 kB RAM data memory
	- 2.4 GHz and Sub-GHz radio operation
	- Transmit power:
	- 2.4 GHz radio: Up to 19 dBm
		- Sub-GHz radio: Up to 20 dBm

• **Low Energy Consumption**

- 8.4 mA RX current at 38.4 kbps, GFSK, 169 MHz
- 10.0 mA RX current at 1 Mbps, GFSK, 2.4 GHz
- 11 mA RX current at 250 kbps, DSSS-OQPSK, 2.4 GHz
- 8.5 mA TX current at 0 dBm output power at 2.4 GHz
- 35.3 mA TX current at 14 dBm output power at 868 MHz
- 70 μA/MHz in Active Mode (EM0)
- 1.5 μA EM2 DeepSleep current (16 kB RAM retention and RTCC running from LFRCO)

• **High Receiver Performance**

- -94.8 dBm sensitivity at 1 Mbit/s GFSK, 2.4 GHz
- -102.7 dBm sensitivity at 250 kbps DSSS-OQPSK, 2.4 GHz
- -126.2 dBm sensitivity at 600 bps, GFSK, 915 MHz
- -120.6 dBm sensitivity at 2.4 kbps, GFSK, 868 MHz
- -107.4 dBm sensitivity at 4.8 kbps, OOK, 433 MHz
- -112.2 dBm sensitivity at 38.4 kbps, GFSK, 169 MHz

• **Supported Modulation Formats**

- 2/4 (G)FSK with fully configurable shaping
- BPSK / DBPSK TX
- OOK / ASK
- Shaped OQPSK / (G)MSK
- Configurable DSSS and FEC
- **Supported Protocols**
	- Proprietary Protocols
	- Wi-SUN
	- Wireless M-Bus
	- Selected IEEE 802.15.4g SUN-FSK PHYs
	- Low Power Wide Area Networks
- **Suitable for Systems Targeting Compliance With:**
	- FCC Part 90.210 Mask D, FCC part 15.247, 15.231, 15.249
	- ETSI Category I Operation, EN 300 220, EN 300 328
	- ARIB T-108, T-96
	- China regulatory
- **Wide selection of MCU peripherals**
	- 12-bit 1 Msps SAR Analog to Digital Converter (ADC)
	- 2 × Analog Comparator (ACMP)
	- 2 × Digital to Analog Converter (VDAC)
	- 3 × Operational Amplifier (Opamp)
	- Digital to Analog Current Converter (IDAC)
	- Low-Energy Sensor Interface (LESENSE)
	- Multi-channel Capacitive Sense Interface (CSEN)
	- Up to 54 pins connected to analog channels (APORT) shared between analog peripherals
	- Up to 65 General Purpose I/O pins with output state retention and asynchronous interrupts
	- 8 Channel DMA Controller
	- 12 Channel Peripheral Reflex System (PRS)
	- 2 × 16-bit Timer/Counter
		- 3 or 4 Compare/Capture/PWM channels
	- 2 × 32-bit Timer/Counter
		- 3 or 4 Compare/Capture/PWM channels
	- 32-bit Real Time Counter and Calendar
	- 16-bit Low Energy Timer for waveform generation
	- 32-bit Ultra Low Energy Timer/Counter for periodic wake-up from any Energy Mode
	- 3 × 16-bit Pulse Counter with asynchronous operation
	- 2 × Watchdog Timer with dedicated RC oscillator
	- 4 × Universal Synchronous/Asynchronous Receiver/Transmitter (UART/SPI/SmartCard (ISO 7816)/IrDA/I2S)
	- Low Energy UART (LEUART™)
	- 2×1^2 C interface with SMBus support and address recognition in EM3 Stop
- **Wide Operating Range**
	- 1.8 V to 3.8 V single power supply
	- Integrated DC-DC, down to 1.8 V output with up to 200 mA load current for system
	- Standard (-40 °C to 85 °C) and Extended (-40 °C to 125 °C) temperature grades available

• **Support for Internet Security**

- General Purpose CRC
- True Random Number Generator
- 2 × Hardware Cryptographic Acceleration for AES 128/256, SHA-1, SHA-2 (SHA-224 and SHA-256) and ECC
- **QFN48 7x7 mm Package**
- **QFN68 8x8 mm Package**
- **BGA125 7x7 mm Package**

2. Ordering Information

Figure 2.1. Ordering Code Key

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3. System Overview

3.1 Introduction

The EFR32 product family combines an energy-friendly MCU with a highly integrated radio transceiver. The devices are well suited for any battery operated application as well as other systems requiring high performance and low energy consumption. This section gives a short introduction to the full radio and MCU system. The detailed functional description can be found in the EFR32xG12 Wireless Gecko Reference Manual.

A block diagram of the EFR32FG12 family is shown in Figure 3.1 Detailed EFR32FG12 Block Diagram on page 8. The diagram shows a superset of features available on the family, which vary by OPN. For more information about specific device features, consult [Ordering Information.](#page-2-0)

Figure 3.1. Detailed EFR32FG12 Block Diagram

3.2 Radio

The Gecko family features a radio transceiver supporting proprietary wireless protocols.

3.2.1 Antenna Interface

The 2.4 GHz antenna interface consists of two pins (2G4RF_IOP and 2G4RF_ION) that interface directly to the on-chip BALUN. The 2G4RF ION pin should be grounded externally.

The external components and power supply connections for the antenna interface typical applications are shown in the RF Matching Networks section.

3.2.2 Fractional-N Frequency Synthesizer

The EFR32FG12 contains a high performance, low phase noise, fully integrated fractional-N frequency synthesizer. The synthesizer is used in receive mode to generate the LO frequency used by the down-conversion mixer. It is also used in transmit mode to directly generate the modulated RF carrier.

The fractional-N architecture provides excellent phase noise performance combined with frequency resolution better than 100 Hz, with low energy consumption. The synthesizer has fast frequency settling which allows very short receiver and transmitter wake up times to optimize system energy consumption.

3.2.3 Receiver Architecture

The EFR32FG12 uses a low-IF receiver architecture, consisting of a Low-Noise Amplifier (LNA) followed by an I/Q down-conversion mixer, employing a crystal reference. The I/Q signals are further filtered and amplified before being sampled by the IF analog-to-digital converter (IFADC).

The IF frequency is configurable from 150 kHz to 1371 kHz. The IF can further be configured for high-side or low-side injection, providing flexibility with respect to known interferers at the image frequency.

The Automatic Gain Control (AGC) block adjusts the receiver gain to optimize performance and avoid saturation for excellent selectivity and blocking performance. The 2.4 GHz radio is calibrated at production to improve image rejection performance. The sub-GHz radio can be calibrated on-demand by the user for the desired frequency band.

Demodulation is performed in the digital domain. The demodulator performs configurable decimation and channel filtering to allow receive bandwidths ranging from 0.1 to 2530 kHz. High carrier frequency and baud rate offsets are tolerated by active estimation and compensation. Advanced features supporting high quality communication under adverse conditions include forward error correction by block and convolutional coding as well as Direct Sequence Spread Spectrum (DSSS) for 2.4 GHz and sub-GHz bands.

A Received Signal Strength Indicator (RSSI) is available for signal quality metrics, for level-based proximity detection, and for RF channel access by Collision Avoidance (CA) or Listen Before Talk (LBT) algorithms. An RSSI capture value is associated with each received frame and the dynamic RSSI measurement can be monitored throughout reception.

The EFR32FG12 features integrated support for antenna diversity to mitigate the problem of frequency-selective fading due to multipath propagation and improve link budget. Support for antenna diversity is available for specific PHY configurations in 2.4 GHz and sub-GHz bands. Internal configurable hardware controls an external switch for automatic switching between antennae during RF receive detection operations.

Note: Due to the shorter preamble of 802.15.4 and BLE packets, RX diversity is not supported.

3.2.4 Transmitter Architecture

The EFR32FG12 uses a direct-conversion transmitter architecture. For constant envelope modulation formats, the modulator controls phase and frequency modulation in the frequency synthesizer. Transmit symbols or chips are optionally shaped by a digital shaping filter. The shaping filter is fully configurable, including the BT product, and can be used to implement Gaussian or Raised Cosine shaping.

Carrier Sense Multiple Access - Collision Avoidance (CSMA-CA) or Listen Before Talk (LBT) algorithms can be automatically timed by the EFR32FG12. These algorithms are typically defined by regulatory standards to improve inter-operability in a given bandwidth between devices that otherwise lack synchronized RF channel access.

3.2.5 Wake on Radio

The Wake on Radio feature allows flexible, autonomous RF sensing, qualification, and demodulation without required MCU activity, using a subsystem of the EFR32FG12 including the Radio Controller (RAC), Peripheral Reflex System (PRS), and Low Energy peripherals.

3.2.6 RFSENSE

The RFSENSE peripheral generates a system wakeup interrupt upon detection of wideband RF energy at the antenna interface, providing true RF wakeup capabilities from low energy modes including EM2, EM3 and EM4.

RFSENSE triggers on a relatively strong RF signal and is available in the lowest energy modes, allowing exceptionally low energy consumption. RFSENSE does not demodulate or otherwise qualify the received signal, but software may respond to the wakeup event by enabling normal RF reception.

Various strategies for optimizing power consumption and system response time in presence of false alarms may be employed using available timer peripherals.

3.2.7 Flexible Frame Handling

EFR32FG12 has an extensive and flexible frame handling support for easy implementation of even complex communication protocols. The Frame Controller (FRC) supports all low level and timing critical tasks together with the Radio Controller and Modulator/Demodulator:

- Highly adjustable preamble length
- Up to 2 simultaneous synchronization words, each up to 32 bits and providing separate interrupts
- Frame disassembly and address matching (filtering) to accept or reject frames
- Automatic ACK frame assembly and transmission
- Fully flexible CRC generation and verification:
	- Multiple CRC values can be embedded in a single frame
	- 8, 16, 24 or 32-bit CRC value
	- Configurable CRC bit and byte ordering
- Selectable bit-ordering (least significant or most significant bit first)
- Optional data whitening
- Optional Forward Error Correction (FEC), including convolutional encoding / decoding and block encoding / decoding
- Half rate convolutional encoder and decoder with constraint lengths from 2 to 7 and optional puncturing
- Optional symbol interleaving, typically used in combination with FEC
- Symbol coding, such as Manchester or DSSS, or biphase space encoding using FEC hardware
- UART encoding over air, with start and stop bit insertion / removal
- Test mode support, such as modulated or unmodulated carrier output
- Received frame timestamping

3.2.8 Packet and State Trace

The EFR32FG12 Frame Controller has a packet and state trace unit that provides valuable information during the development phase. It features:

- Non-intrusive trace of transmit data, receive data and state information
- Data observability on a single-pin UART data output, or on a two-pin SPI data output
- Configurable data output bitrate / baudrate
- Multiplexed transmitted data, received data and state / meta information in a single serial data stream

3.2.9 Data Buffering

The EFR32FG12 features an advanced Radio Buffer Controller (BUFC) capable of handling up to 4 buffers of adjustable size from 64 bytes to 4096 bytes. Each buffer can be used for RX, TX or both. The buffer data is located in RAM, enabling zero-copy operations.

3.2.10 Radio Controller (RAC)

The Radio Controller controls the top level state of the radio subsystem in the EFR32FG12. It performs the following tasks:

- Precisely-timed control of enabling and disabling of the receiver and transmitter circuitry
- Run-time calibration of receiver, transmitter and frequency synthesizer
- Detailed frame transmission timing, including optional LBT or CSMA-CA

3.2.11 Random Number Generator

The Frame Controller (FRC) implements a random number generator that uses entropy gathered from noise in the RF receive chain. The data is suitable for use in cryptographic applications.

Output from the random number generator can be used either directly or as a seed or entropy source for software-based random number generator algorithms such as Fortuna.

3.3 Power

The EFR32FG12 has an Energy Management Unit (EMU) and efficient integrated regulators to generate internal supply voltages. Only a single external supply voltage is required, from which all internal voltages are created. An optional integrated DC-DC buck regulator can be utilized to further reduce the current consumption. The DC-DC regulator requires one external inductor and one external capacitor.

The EFR32FG12 device family includes support for internal supply voltage scaling, as well as two different power domains groups for peripherals. These enhancements allow for further supply current reductions and lower overall power consumption.

AVDD and VREGVDD need to be 1.8 V or higher for the MCU to operate across all conditions; however the rest of the system will operate down to 1.62 V, including the digital supply and I/O. This means that the device is fully compatible with 1.8 V components. Running from a sufficiently high supply, the device can use the DC-DC to regulate voltage not only for itself, but also for other PCB components, supplying up to a total of 200 mA.

3.3.1 Energy Management Unit (EMU)

The Energy Management Unit manages transitions of energy modes in the device. Each energy mode defines which peripherals and features are available and the amount of current the device consumes. The EMU can also be used to turn off the power to unused RAM blocks, and it contains control registers for the DC-DC regulator and the Voltage Monitor (VMON). The VMON is used to monitor multiple supply voltages. It has multiple channels which can be programmed individually by the user to determine if a sensed supply has fallen below a chosen threshold.

3.3.2 DC-DC Converter

The DC-DC buck converter covers a wide range of load currents and provides up to 90% efficiency in energy modes EM0, EM1, EM2 and EM3, and can supply up to 200 mA to the device and surrounding PCB components. Patented RF noise mitigation allows operation of the DC-DC converter without degrading sensitivity of radio components. Protection features include programmable current limiting, short-circuit protection, and dead-time protection. The DC-DC converter may also enter bypass mode when the input voltage is too low for efficient operation. In bypass mode, the DC-DC input supply is internally connected directly to its output through a low resistance switch. Bypass mode also supports in-rush current limiting to prevent input supply voltage droops due to excessive output current transients.

3.3.3 Power Domains

The EFR32FG12 has two peripheral power domains for operation in EM2 and lower. If all of the peripherals in a peripheral power domain are configured as unused, the power domain for that group will be powered off in the low-power mode, reducing the overall current consumption of the device.

Table 3.1. Peripheral Power Subdomains

3.4 General Purpose Input/Output (GPIO)

EFR32FG12 has up to 65 General Purpose Input/Output pins. Each GPIO pin can be individually configured as either an output or input. More advanced configurations including open-drain, open-source, and glitch-filtering can be configured for each individual GPIO pin. The GPIO pins can be overridden by peripheral connections, like SPI communication. Each peripheral connection can be routed to several GPIO pins on the device. The input value of a GPIO pin can be routed through the Peripheral Reflex System to other peripherals. The GPIO subsystem supports asynchronous external pin interrupts.

3.5 Clocking

3.5.1 Clock Management Unit (CMU)

The Clock Management Unit controls oscillators and clocks in the EFR32FG12. Individual enabling and disabling of clocks to all peripherals is performed by the CMU. The CMU also controls enabling and configuration of the oscillators. A high degree of flexibility allows software to optimize energy consumption in any specific application by minimizing power dissipation in unused peripherals and oscillators.

3.5.2 Internal and External Oscillators

The EFR32FG12 supports two crystal oscillators and fully integrates four RC oscillators, listed below.

- A high frequency crystal oscillator (HFXO) with integrated load capacitors, tunable in small steps, provides a precise timing reference for the MCU. Crystal frequencies in the range from 38 to 40 MHz are supported. An external clock source such as a TCXO can also be applied to the HFXO input for improved accuracy over temperature.
- A 32.768 kHz crystal oscillator (LFXO) provides an accurate timing reference for low energy modes.
- An integrated high frequency RC oscillator (HFRCO) is available for the MCU system, when crystal accuracy is not required. The HFRCO employs fast startup at minimal energy consumption combined with a wide frequency range.
- An integrated auxilliary high frequency RC oscillator (AUXHFRCO) is available for timing the general-purpose ADC and the Serial Wire Viewer port with a wide frequency range.
- An integrated low frequency 32.768 kHz RC oscillator (LFRCO) can be used as a timing reference in low energy modes, when crystal accuracy is not required.
- An integrated ultra-low frequency 1 kHz RC oscillator (ULFRCO) is available to provide a timing reference at the lowest energy consumption in low energy modes.

3.6 Counters/Timers and PWM

3.6.1 Timer/Counter (TIMER)

TIMER peripherals keep track of timing, count events, generate PWM outputs and trigger timed actions in other peripherals through the PRS system. The core of each TIMER is a 16-bit counter with up to 4 compare/capture channels. Each channel is configurable in one of three modes. In capture mode, the counter state is stored in a buffer at a selected input event. In compare mode, the channel output reflects the comparison of the counter to a programmed threshold value. In PWM mode, the TIMER supports generation of pulse-width modulation (PWM) outputs of arbitrary waveforms defined by the sequence of values written to the compare registers, with optional dead-time insertion available in timer unit TIMER_0 only.

3.6.2 Wide Timer/Counter (WTIMER)

WTIMER peripherals function just as TIMER peripherals, but are 32 bits wide. They keep track of timing, count events, generate PWM outputs and trigger timed actions in other peripherals through the PRS system. The core of each WTIMER is a 32-bit counter with up to 4 compare/capture channels. Each channel is configurable in one of three modes. In capture mode, the counter state is stored in a buffer at a selected input event. In compare mode, the channel output reflects the comparison of the counter to a programmed threshold value. In PWM mode, the WTIMER supports generation of pulse-width modulation (PWM) outputs of arbitrary waveforms defined by the sequence of values written to the compare registers, with optional dead-time insertion available in timer unit WTIMER_0 only.

3.6.3 Real Time Counter and Calendar (RTCC)

The Real Time Counter and Calendar (RTCC) is a 32-bit counter providing timekeeping in all energy modes. The RTCC includes a Binary Coded Decimal (BCD) calendar mode for easy time and date keeping. The RTCC can be clocked by any of the on-board oscillators with the exception of the AUXHFRCO, and it is capable of providing system wake-up at user defined instances. When receiving frames, the RTCC value can be used for timestamping. The RTCC includes 128 bytes of general purpose data retention, allowing easy and convenient data storage in all energy modes down to EM4H.

3.6.4 Low Energy Timer (LETIMER)

The unique LETIMER is a 16-bit timer that is available in energy mode EM0 Active, EM1 Sleep, EM2 Deep Sleep, and EM3 Stop. This allows it to be used for timing and output generation when most of the device is powered down, allowing simple tasks to be performed while the power consumption of the system is kept at an absolute minimum. The LETIMER can be used to output a variety of waveforms with minimal software intervention. The LETIMER is connected to the Real Time Counter and Calendar (RTCC), and can be configured to start counting on compare matches from the RTCC.

3.6.5 Ultra Low Power Wake-up Timer (CRYOTIMER)

The CRYOTIMER is a 32-bit counter that is capable of running in all energy modes. It can be clocked by either the 32.768 kHz crystal oscillator (LFXO), the 32.768 kHz RC oscillator (LFRCO), or the 1 kHz RC oscillator (ULFRCO). It can provide periodic Wakeup events and PRS signals which can be used to wake up peripherals from any energy mode. The CRYOTIMER provides a wide range of interrupt periods, facilitating flexible ultra-low energy operation.

3.6.6 Pulse Counter (PCNT)

The Pulse Counter (PCNT) peripheral can be used for counting pulses on a single input or to decode quadrature encoded inputs. The clock for PCNT is selectable from either an external source on pin PCTNn_S0IN or from an internal timing reference, selectable from among any of the internal oscillators, except the AUXHFRCO. The peripheral may operate in energy mode EM0 Active, EM1 Sleep, EM2 Deep Sleep, and EM3 Stop.

3.6.7 Watchdog Timer (WDOG)

The watchdog timer can act both as an independent watchdog or as a watchdog synchronous with the CPU clock. It has windowed monitoring capabilities, and can generate a reset or different interrupts depending on the failure mode of the system. The watchdog can also monitor autonomous systems driven by PRS.

3.7 Communications and Other Digital Peripherals

3.7.1 Universal Synchronous/Asynchronous Receiver/Transmitter (USART)

The Universal Synchronous/Asynchronous Receiver/Transmitter is a flexible serial I/O interface. It supports full duplex asynchronous UART communication with hardware flow control as well as RS-485, SPI, MicroWire and 3-wire. It can also interface with devices supporting:

- ISO7816 SmartCards
- IrDA
- I 2S

3.7.2 Low Energy Universal Asynchronous Receiver/Transmitter (LEUART)

The unique LEUARTTM provides two-way UART communication on a strict power budget. Only a 32.768 kHz clock is needed to allow UART communication up to 9600 baud. The LEUART includes all necessary hardware to make asynchronous serial communication possible with a minimum of software intervention and energy consumption.

3.7.3 Inter-Integrated Circuit Interface (I2C)

The $12C$ interface enables communication between the MCU and a serial $12C$ bus. It is capable of acting as both a master and a slave and supports multi-master buses. Standard-mode, fast-mode and fast-mode plus speeds are supported, allowing transmission rates from 10 kbit/s up to 1 Mbit/s. Slave arbitration and timeouts are also available, allowing implementation of an SMBus-compliant system. The interface provided to software by the I^2C peripheral allows precise timing control of the transmission process and highly automated transfers. Automatic recognition of slave addresses is provided in active and low energy modes.

3.7.4 Peripheral Reflex System (PRS)

The Peripheral Reflex System provides a communication network between different peripherals without software involvement. Peripherals producing Reflex signals are called producers. The PRS routes Reflex signals from producers to consumer peripherals, which in turn perform actions in response. Edge triggers and other functionality such as simple logic operations (AND, OR, NOT) can be applied by the PRS to the signals. The PRS allows peripheral to act autonomously without waking the MCU core, saving power.

3.7.5 Low Energy Sensor Interface (LESENSE)

The Low Energy Sensor Interface LESENSETM is a highly configurable sensor interface with support for up to 16 individually configurable sensors. By controlling the analog comparators, ADC, and DAC, LESENSE is capable of supporting a wide range of sensors and measurement schemes, and can for instance measure LC sensors, resistive sensors and capacitive sensors. LESENSE also includes a programmable finite state machine which enables simple processing of measurement results without CPU intervention. LESENSE is available in energy mode EM2, in addition to EM0 and EM1, making it ideal for sensor monitoring in applications with a strict energy budget.

3.8 Security Features

3.8.1 General Purpose Cyclic Redundancy Check (GPCRC)

The GPCRC block implements a Cyclic Redundancy Check (CRC) function. It supports both 32-bit and 16-bit polynomials. The supported 32-bit polynomial is 0x04C11DB7 (IEEE 802.3), while the 16-bit polynomial can be programmed to any value, depending on the needs of the application.

3.8.2 Crypto Accelerator (CRYPTO)

The Crypto Accelerator is a fast and energy-efficient autonomous hardware encryption and decryption accelerator. EFR32 devices support AES encryption and decryption with 128- or 256-bit keys, ECC over both GF(P) and GF(2^m), SHA-1 and SHA-2 (SHA-224 and SHA-256).

Supported block cipher modes of operation for AES include: ECB, CTR, CBC, PCBC, CFB, OFB, GCM, CBC-MAC, GMAC and CCM.

Supported ECC NIST recommended curves include P-192, P-224, P-256, K-163, K-233, B-163 and B-233.

The CRYPTO1 block is tightly linked to the Radio Buffer Controller (BUFC) enabling fast and efficient autonomous cipher operations on data buffer content. It allows fast processing of GCM (AES), ECC and SHA with little CPU intervention.

CRYPTO also provides trigger signals for DMA read and write operations.

3.8.3 True Random Number Generator (TRNG)

The TRNG is a non-deterministic random number generator based on a full hardware solution. The TRNG is validated with NIST800-22 and AIS-31 test suites as well as being suitable for FIPS 140-2 certification (for the purposes of cryptographic key generation).

Note: TRNG operation is only supported at VSCALE2. TRNG cannot be used at VSCALE0.

3.8.4 Security Management Unit (SMU)

The Security Management Unit (SMU) allows software to set up fine-grained security for peripheral access, which is not possible in the Memory Protection Unit (MPU). Peripherals may be secured by hardware on an individual basis, such that only priveleged accesses to the peripheral's register interface will be allowed. When an access fault occurs, the SMU reports the specific peripheral involved and can optionally generate an interrupt.

3.9 Analog

3.9.1 Analog Port (APORT)

The Analog Port (APORT) is an analog interconnect matrix allowing access to many analog peripherals on a flexible selection of pins. Each APORT bus consists of analog switches connected to a common wire. Since many clients can operate differentially, buses are grouped by X/Y pairs.

3.9.2 Analog Comparator (ACMP)

The Analog Comparator is used to compare the voltage of two analog inputs, with a digital output indicating which input voltage is higher. Inputs are selected from among internal references and external pins. The tradeoff between response time and current consumption is configurable by software. Two 6-bit reference dividers allow for a wide range of internally-programmable reference sources. The ACMP can also be used to monitor the supply voltage. An interrupt can be generated when the supply falls below or rises above the programmable threshold.

3.9.3 Analog to Digital Converter (ADC)

The ADC is a Successive Approximation Register (SAR) architecture, with a resolution of up to 12 bits at up to 1 Msps. The output sample resolution is configurable and additional resolution is possible using integrated hardware for averaging over multiple samples. The ADC includes integrated voltage references and an integrated temperature sensor. Inputs are selectable from a wide range of sources, including pins configurable as either single-ended or differential.

3.9.4 Capacitive Sense (CSEN)

The CSEN peripheral is a dedicated Capacitive Sensing block for implementing touch-sensitive user interface elements such a switches and sliders. The CSEN peripheral uses a charge ramping measurement technique, which provides robust sensing even in adverse conditions including radiated noise and moisture. The peripheral can be configured to take measurements on a single port pin or scan through multiple pins and store results to memory through DMA. Several channels can also be shorted together to measure the combined capacitance or implement wake-on-touch from very low energy modes. Hardware includes a digital accumulator and an averaging filter, as well as digital threshold comparators to reduce software overhead.

3.9.5 Digital to Analog Current Converter (IDAC)

The IDAC can source or sink a configurable constant current. This current can be driven on an output pin or routed to the selected ADC input pin for capacitive sensing. The full-scale current is programmable between 0.05 µA and 64 µA with several ranges consisting of various step sizes.

3.9.6 Digital to Analog Converter (VDAC)

The Digital to Analog Converter (VDAC) can convert a digital value to an analog output voltage. The VDAC is a fully differential, 500 ksps, 12-bit converter. The opamps are used in conjunction with the VDAC, to provide output buffering. One opamp is used per singleended channel, or two opamps are used to provide differential outputs. The VDAC may be used for a number of different applications such as sensor interfaces or sound output. The VDAC can generate high-resolution analog signals while the MCU is operating at low frequencies and with low total power consumption. Using DMA and a timer, the VDAC can be used to generate waveforms without any CPU intervention. The VDAC is available in all energy modes down to and including EM3.

3.9.7 Operational Amplifiers

The opamps are low power amplifiers with a high degree of flexibility targeting a wide variety of standard opamp application areas, and are available down to EM3. With flexible built-in programming for gain and interconnection they can be configured to support multiple common opamp functions. All pins are also available externally for filter configurations. Each opamp has a rail to rail input and a rail to rail output. They can be used in conjunction with the VDAC peripheral or in stand-alone configurations. The opamps save energy, PCB space, and cost as compared with standalone opamps because they are integrated on-chip.

3.10 Reset Management Unit (RMU)

The RMU is responsible for handling reset of the EFR32FG12. A wide range of reset sources are available, including several power supply monitors, pin reset, software controlled reset, core lockup reset, and watchdog reset.

3.11 Core and Memory

3.11.1 Processor Core

The ARM Cortex-M processor includes a 32-bit RISC processor integrating the following features and tasks in the system:

- ARM Cortex-M4 RISC processor achieving 1.25 Dhrystone MIPS/MHz
- Memory Protection Unit (MPU) supporting up to 8 memory segments
- Up to 1024 kB flash program memory
- Up to 256 kB RAM data memory
- Configuration and event handling of all peripherals
- 2-pin Serial-Wire debug interface

3.11.2 Memory System Controller (MSC)

The Memory System Controller (MSC) is the program memory unit of the microcontroller. The flash memory is readable and writable from both the Cortex-M and DMA. The flash memory is divided into two blocks; the main block and the information block. Program code is normally written to the main block, whereas the information block is available for special user data and flash lock bits. There is also a read-only page in the information block containing system and device calibration data. Read and write operations are supported in energy modes EM0 Active and EM1 Sleep.

3.11.3 Linked Direct Memory Access Controller (LDMA)

The Linked Direct Memory Access (LDMA) controller allows the system to perform memory operations independently of software. This reduces both energy consumption and software workload. The LDMA allows operations to be linked together and staged, enabling sophisticated operations to be implemented.

3.12 Memory Map

The EFR32FG12 memory map is shown in the figures below. RAM and flash sizes are for the largest memory configuration.

Figure 3.2. EFR32FG12 Memory Map — Core Peripherals and Code Space

Figure 3.3. EFR32FG12 Memory Map — Peripherals

3.13 Configuration Summary

The features of the EFR32FG12 are a subset of the feature set described in the device reference manual. The table below describes device specific implementation of the features. Remaining modules support full configuration.

Table 3.2. Configuration Summary

4. Electrical Specifications

4.1 Electrical Characteristics

All electrical parameters in all tables are specified under the following conditions, unless stated otherwise:

- Typical values are based on T_{AMB}=25 °C and V_{DD}= 3.3 V, by production test and/or technology characterization.
- Radio performance numbers are measured in conducted mode, based on Silicon Laboratories reference designs using output power-specific external RF impedance-matching networks for interfacing to a 50 Ω antenna.
- Minimum and maximum values represent the worst conditions across supply voltage, process variation, and operating temperature, unless stated otherwise.

Refer to [4.1.2.1 General Operating Conditions](#page-24-0) for more details about operational supply and temperature limits.

4.1.1 Absolute Maximum Ratings

Stresses above those listed below may cause permanent damage to the device. This is a stress rating only and functional operation of the devices at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. For more information on the available quality and reliability data, see the Quality and Reliability Monitor Report at<http://www.silabs.com/support/quality/pages/default.aspx>.

- 1.When a GPIO pin is routed to the analog block through the APORT, the maximum voltage = IOVDD.
- 2.Valid for IOVDD in valid operating range or when IOVDD is undriven (high-Z). If IOVDD is connected to a low-impedance source below the valid operating range (e.g. IOVDD shorted to VSS), the pin voltage maximum is IOVDD + 0.3 V, to avoid exceeding the maximum IO current specifications.
- 3. To operate above the IOVDD supply rail, over-voltage tolerance must be enabled according to the GPIO_Px_OVTDIS register. Pins with over-voltage tolerance disabled have the same limits as Standard GPIO.

4.1.2 Operating Conditions

When assigning supply sources, the following requirements must be observed:

- VREGVDD must be greater than or equal to AVDD, DVDD, RFVDD, PAVDD and all IOVDD supplies.
- VREGVDD = AVDD
- DVDD ≤ AVDD
- IOVDD ≤ AVDD
- RFVDD ≤ AVDD
- PAVDD ≤ AVDD

4.1.2.1 General Operating Conditions

Table 4.2. General Operating Conditions

- 4. When the CSEN peripheral is used with chopping enabled (CSEN_CTRL_CHOPEN = ENABLE), IOVDD must be equal to AVDD.
- 5. The system designer should consult the characteristic specs of the capacitor used on DECOUPLE to ensure its capacitance value stays within the specified bounds across temperature and DC bias.
- 6. VSCALE0 to VSCALE2 voltage change transitions occur at a rate of 10 mV / usec for approximately 20 usec. During this transition, peak currents will be dependent on the value of the DECOUPLE output capacitor, from 35 mA (with a 1 µF capacitor) to 70 mA (with a 2.7 µF capacitor).

4.1.3 Thermal Characteristics

Table 4.3. Thermal Characteristics

4.1.4 DC-DC Converter

Test conditions: L_DCDC=4.7 µH (Murata LQH3NPN4R7MM0L), C_DCDC=4.7 µF (Samsung CL10B475KQ8NQNC), V_DCDC_I=3.3 V, V_DCDC_O=1.8 V, I_DCDC_LOAD=50 mA, Heavy Drive configuration, F_DCDC_LN=7 MHz, unless otherwise indicated.

Table 4.4. DC-DC Converter

Note:

1. Due to internal dropout, the DC-DC output will never be able to reach its input voltage, V_{VREGVDD}.

- 2. LP mode controller is a hysteretic controller that maintains the output voltage within the specified limits.
- 3. LPCMPBIASEMxx refers to either LPCMPBIASEM234H in the EMU_DCDCMISCCTRL register or LPCMPBIASEM01 in the EMU_DCDCLOEM01CFG register, depending on the energy mode.
- 4. Drive levels are defined by configuration of the PFETCNT and NFETCNT registers. Light Drive: PFETCNT=NFETCNT=3; Medium Drive: PFETCNT=NFETCNT=7; Heavy Drive: PFETCNT=NFETCNT=15.
- 5. Output voltage under/over-shoot and regulation are specified with C_{DCDC} 4.7 µF. Different settings for DCDCLNCOMPCTRL must be used if C_{DCDC} is lower than 4.7 μ F. See Application Note AN0948 for details.

4.1.5 Current Consumption

4.1.5.1 Current Consumption 3.3 V without DC-DC Converter

Unless otherwise indicated, typical conditions are: VREGVDD = AVDD = DVDD = RFVDD = PAVDD = 3.3 V. T = 25 °C. DCDC is off. Minimum and maximum values in this table represent the worst conditions across process variation at T = 25 °C.

Table 4.5. Current Consumption 3.3 V without DC-DC Converter

4.1.5.2 Current Consumption 3.3 V using DC-DC Converter

Unless otherwise indicated, typical conditions are: VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD = 1.8 V DC-DC output. T = 25 °C. Minimum and maximum values in this table represent the worst conditions across process variation at T = 25 °C.

Note:

1. DCDC Low Noise DCM Mode = Light Drive (PFETCNT=NFETCNT=3), F=3.0 MHz (RCOBAND=0), ANASW=DVDD.

2. CMU_HFXOCTRL_LOWPOWER=0.

3. DCDC Low Noise CCM Mode = Light Drive (PFETCNT=NFETCNT=3), F=6.4 MHz (RCOBAND=4), ANASW=DVDD.

4. DCDC Low Power Mode = Medium Drive, LPOSCDIV=1, LPCMPBIASEM234H=0, LPCLIMILIMSEL=1, ANASW=DVDD.

5. CMU_LFRCOCTRL_ENVREF = 1, CMU_LFRCOCTRL_VREFUPDATE = 1

4.1.5.3 Current Consumption 1.8 V without DC-DC Converter

Unless otherwise indicated, typical conditions are: VREGVDD = AVDD = DVDD = RFVDD = PAVDD = 1.8 V. T = 25 °C. DCDC is off. Minimum and maximum values in this table represent the worst conditions across process variation at T = 25 °C.

Table 4.7. Current Consumption 1.8 V without DC-DC Converter

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4.1.5.4 Current Consumption Using Radio 3.3 V with DC-DC

Unless otherwise indicated, typical conditions are: VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD = 1.8 V. T = 25 °C. Minimum and maximum values in this table represent the worst conditions across process variation at T = 25 °C.

4.1.6 Wake Up Times

Table 4.9. Wake Up Times

Note:

1. Time from wake up request until first instruction is executed. Wakeup results in device reset.

2.Scaling up from VSCALE0 to VSCALE2 requires approximately 30.3 µs + 28 HFCLKs.

3. VSCALE0 to VSCALE2 voltage change transitions occur at a rate of 10 mV/µs for approximately 20 µs. During this transition, peak currents will be dependent on the value of the DECOUPLE output capacitor, from 35 mA (with a 1 µF capacitor) to 70 mA (with a 2.7 µF capacitor).

4. Scaling down from VSCALE2 to VSCALE0 requires approximately 2.8 µs + 29 HFCLKs.

4.1.7 Brown Out Detector (BOD)

Table 4.10. Brown Out Detector (BOD)

4.1.8 Frequency Synthesizer

Table 4.11. Frequency Synthesizer

4.1.9 2.4 GHz RF Transceiver Characteristics

4.1.9.1 RF Transmitter General Characteristics for 2.4 GHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 2.45 GHz.

Table 4.12. RF Transmitter General Characteristics for 2.4 GHz Band

Note:

1.Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.9.2 RF Receiver General Characteristics for 2.4 GHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 2.45 GHz.

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4.1.9.3 RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate

Unless otherwise indicated, typical conditions are: $T = 25$ °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz. Maximum duty cycle of 85%.

Table 4.14. RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate

Note:

1. For 2476 MHz, 1.5 dB of power backoff is used to achieve this value.

2. For 2478 MHz, 4.2 dB of power backoff is used to achieve this value.

4.1.9.4 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz.

Table 4.15. RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate

4.1.9.5 RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate

Unless otherwise indicated, typical conditions are: $T = 25$ °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz. Maximum duty cycle of 85%.

Table 4.16. RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate

Note:

1. For 2472 MHz, 1.3 dB of power backoff is used to achieve this value.

2. For 2474 MHz, 3.8 dB of power backoff is used to achieve this value.

3. For 2476 MHz, 7 dB of power backoff is used to achieve this value.

4. For 2478 MHz, 11.2 dB of power backoff is used to achieve this value.

4.1.9.6 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate

Unless otherwise indicated, typical conditions are: $T = 25$ °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz¹.

Table 4.17. RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate

Note:

1. For the BLE 2Mbps in-band blocking performance, there may be up to 5 spurious response channels where the requirement of 30.8% PER is not met and therefore an exception will need to be taken for each of these frequencies to meet the requirements of the BLE standard.

2. Reference signal is defined 2GFSK at -67 dBm, Modulation index = 0.5, BT = 0.5, Bit rate = 2 Mbps, desired data = PRBS9; interferer data = PRBS15; frequency accuracy better than 1 ppm.

4.1.9.7 RF Transmitter Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 2.45 GHz. Maximum duty cycle of 66%.

Table 4.18. RF Transmitter Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band

1. Reference packet is defined as 20 octet PSDU, modulated according to 802.15.4-2011 DSSS-OQPSK in the 2.4GHz band, with pseudo-random packet data content.

2. For 2415 MHz, 2 dB of power backoff is used to achieve this value.

3. For 2475 MHz, 2 dB of power backoff is used to achieve this value.

4. For 2480 MHz, 13 dB of power backoff is used to achieve this value.

5. Specified at maximum power output level of 10 dBm.

4.1.9.8 RF Receiver Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 2.45 GHz.

Table 4.19. RF Receiver Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band

1. Reference signal is defined as O-QPSK DSSS per 802.15.4, Frequency range = 2400-2483.5 MHz, Symbol rate = 62.5 ksymbols/s.

2. Reference sensitivity level is -85 dBm.

3. Filter is characterized as a symmetric bandpass centered on the adjacent channel having a 3dB bandwidth of 4.6 MHz and stopband rejection better than 26 dB beyond 3.15 MHz from the adjacent carrier.

4. Due to low-IF frequency, there is some overlap of adjacent channel and image channel bands. Adjacent channel CW blocker tests place the Interferer center frequency at the Desired frequency ± 5 MHz on the channel raster, whereas the image rejection test places the CW interferer near the image frequency of the Desired signal carrier, regardless of the channel raster.

5. This is an IEEE 802.11b/g ERP-PBCC 22 MBit/s signal as defined by the IEEE 802.11 specification and IEEE 802.11g addendum.

4.1.10 Sub-GHz RF Transceiver Characteristics

4.1.10.1 Sub-GHz RF Transmitter characteristics for 915 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 915 MHz.

1.Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

2. Definition of reference signal is O-QPSK DSSS per 802.15.4, Frequency Range = 902-928 MHz, Data rate = 250 kbps, 16-chip PN sequence mapping.

4.1.10.2 Sub-GHz RF Receiver Characteristics for 915 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 915 MHz.

4.1.10.3 Sub-GHz RF Transmitter characteristics for 868 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 868 MHz.

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Note:

1.Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.10.4 Sub-GHz RF Receiver Characteristics for 868 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 868 MHz.

Table 4.23. Sub-GHz RF Receiver Characteristics for 868 MHz Band

1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.797 kHz, channel spacing = 12.5 kHz.

2. Definition of reference signal is 38.4 kbps 2GFSK, BT=0.5, Δf = 20 kHz, RX channel BW = 74.809 kHz, channel spacing = 100 kHz.

3. Definition of reference signal is 500 kbps 2GFSK, BT=0.5, Δf = 125 kHz, RX channel BW = 753.320 kHz.

4. Definition of reference signal is 20 kbps BPSK

5. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.10.5 Sub-GHz RF Transmitter characteristics for 490 MHz Band

Unless otherwise indicated, typical conditions are: $T = 25 °C$, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 490 MHz.

Note:

1.Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.10.6 Sub-GHz RF Receiver Characteristics for 490 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 490 MHz.

Table 4.25. Sub-GHz RF Receiver Characteristics for 490 MHz Band

1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.798 kHz, channel spacing = 12.5 kHz.

2. Definition of reference signal is 38.4 kbps 2GFSK, BT=0.5, Δf = 20 kHz, RX channel BW = 74.809 kHz, channel spacing = 100 kHz.

3. Definition of reference signal is 10 kbps 2GFSK, BT=0.5, Δf = 5 kHz, RX channel BW = 20.038 kHz.

4. Definition of reference signal is 100 kbps 2GFSK, BT=0.5, Δf = 50 kHz, RX channel BW = 198.024 kHz.

5. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.10.7 Sub-GHz RF Transmitter characteristics for 433 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 433 MHz.

1.Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.10.8 Sub-GHz RF Receiver Characteristics for 433 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 433 MHz.

Table 4.27. Sub-GHz RF Receiver Characteristics for 433 MHz Band

1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.798 kHz, channel spacing = 12.5 kHz.

- 2. Definition of reference signal is 50 kbps 2GFSK, BT=0.5, Δf = 25 kHz, RX channel BW = 99.012 kHz, channel spacing = 200 kHz.
- 3. Definition of reference signal is 4.8 kbps OOK, RX channel BW = 306.036 kHz, channel spacing = 500 kHz.

4. Definition of reference signal is 100 kbps 2GFSK, BT=0.5, Δf = 50 kHz, RX channel BW = 198.024 kHz, channel spacing = 200 kHz.

5. Definition of reference signal is 9.6 kbps 4GFSK, BT=0.5, inner deviation = 0.8 kHz, RX channel BW = 8.5 kHz, channel spacing = 12.5 kHz.

6. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.10.9 Sub-GHz RF Transmitter characteristics for 315 MHz Band

Unless otherwise indicated, typical conditions are: $T = 25 °C$, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 315 MHz.

Note:

1.Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.
4.1.10.10 Sub-GHz RF Receiver Characteristics for 315 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 315 MHz.

Table 4.29. Sub-GHz RF Receiver Characteristics for 315 MHz Band

1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.798 kHz, channel spacing = 12.5 kHz.

2. Definition of reference signal is 38.4 kbps 2GFSK, BT=0.5, Δf = 20 kHz, RX channel BW = 74.809 kHz, channel spacing = 100 kHz.

3. Definition of reference signal is 500 kbps 2GFSK, BT=0.5, Δf = 125 kHz, RX channel BW = 753.320 kHz.

4. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.10.11 Sub-GHz RF Transmitter Characteristics for 169 MHz Band

Unless otherwise indicated, typical conditions are: $T = 25 °C$, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 169 MHz.

Note:

1.Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

2. Typical value marginally passes specification. Additional margin can be obtained by increasing the order of the harmonic filter.

4.1.10.12 Sub-GHz RF Receiver Characteristics for 169 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 169 MHz.

Table 4.31. Sub-GHz RF Receiver Characteristics for 169 MHz Band

1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.798 kHz, channel spacing = 12.5 kHz.

2. Definition of reference signal is 38.4 kbps 2GFSK, BT=0.5, Δf = 20 kHz, RX channel BW = 74.809 kHz, channel spacing = 100 kHz.

3. Definition of reference signal is 500 kbps 2GFSK, BT=0.5, Δf = 125 kHz, RX channel BW = 753.320 kHz.

4. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.11 Modem

Parameter Minimum Symbol Test Condition Minimum Typ Max Unit Receive bandwidth $|BW_{RX}|$ Configurable range with 38.4 MHz crystal 0.1 – 2530 kHz IF frequency $|f_{IF}$ Configurable range with 38.4 MHz crystal. Selected steps available. 150 **-** 1371 kHz DSSS symbol length \vert SL_{DSSS} \vert Configurable in steps of 1 chip \vert 2 \vert \vert \vert \vert 32 \vert chips DSSS bits per symbol $|BPS_{DSSS}|$ Configurable $|1 \t - |4 \t |$ bits/ symbol

Table 4.32. Modem

4.1.12 Oscillators

4.1.12.1 Low-Frequency Crystal Oscillator (LFXO)

Table 4.33. Low-Frequency Crystal Oscillator (LFXO)

Note:

1. Total load capacitance as seen by the crystal.

2. The effective load capacitance seen by the crystal will be $C_{LFXO_T}/2$. This is because each XTAL pin has a tuning cap and the two caps will be seen in series by the crystal.

3. Block is supplied by AVDD if ANASW = 0, or DVDD if ANASW=1 in EMU_PWRCTRL register.

4. In CMU_LFXOCTRL register.

4.1.12.2 High-Frequency Crystal Oscillator (HFXO)

Table 4.34. High-Frequency Crystal Oscillator (HFXO)

Note:

1. Total load capacitance as seen by the crystal.

2. The effective load capacitance seen by the crystal will be C_{HFXO-T} /2. This is because each XTAL pin has a tuning cap and the two caps will be seen in series by the crystal.

4.1.12.3 Low-Frequency RC Oscillator (LFRCO)

Table 4.35. Low-Frequency RC Oscillator (LFRCO)

Note:

1. In CMU_LFRCOCTRL register.

2.Block is supplied by AVDD if ANASW = 0, or DVDD if ANASW=1 in EMU_PWRCTRL register.

4.1.12.4 High-Frequency RC Oscillator (HFRCO)

Table 4.36. High-Frequency RC Oscillator (HFRCO)

4.1.12.5 Auxiliary High-Frequency RC Oscillator (AUXHFRCO)

Table 4.37. Auxiliary High-Frequency RC Oscillator (AUXHFRCO)

4.1.12.6 Ultra-low Frequency RC Oscillator (ULFRCO)

Table 4.38. Ultra-low Frequency RC Oscillator (ULFRCO)

4.1.13 Flash Memory Characteristics¹

Table 4.39. Flash Memory Characteristics¹

Note:

1. Flash data retention information is published in the Quarterly Quality and Reliability Report.

2. From setting the ERASEPAGE bit in MSC_WRITECMD to 1 until the BUSY bit in MSC_STATUS is cleared to 0. Internal setup and hold times for flash control signals are included.

3. Mass erase is issued by the CPU and erases all flash.

4. Device erase is issued over the AAP interface and erases all flash, SRAM, the Lock Bit (LB) page, and the User data page Lock Word (ULW).

5. From setting the DEVICEERASE bit in AAP_CMD to 1 until the ERASEBUSY bit in AAP_STATUS is cleared to 0. Internal setup and hold times for flash control signals are included.

6. Measured at 25 °C.

4.1.14 General-Purpose I/O (GPIO)

Table 4.40. General-Purpose I/O (GPIO)

1. GPIO input threshold are proportional to the IOVDD supply, except for RESETn which is proportional to AVDD.

2. In GPIO_Pn_CTRL register.

3. GPIO pull-ups are referenced to the IOVDD supply, except for RESETn, which connects to AVDD.

4.1.15 Voltage Monitor (VMON)

Table 4.41. Voltage Monitor (VMON)

4.1.16 Analog to Digital Converter (ADC)

Specified at 1 Msps, ADCCLK = 16 MHz, BIASPROG = 0, GPBIASACC = 0, unless otherwise indicated.

Table 4.42. Analog to Digital Converter (ADC)

1. The absolute voltage allowed at any ADC input is dictated by the power rail supplied to on-chip circuitry, and may be lower than the effective full scale voltage. All ADC inputs are limited to the ADC supply (AVDD or DVDD depending on

EMU_PWRCTRL_ANASW). Any ADC input routed through the APORT will further be limited by the IOVDD supply to the pin.

2.PSRR is referenced to AVDD when ANASW=0 and to DVDD when ANASW=1 in EMU_PWRCTRL.

- 3. In ADCn_CTRL register.
- 4. In ADCn_BIASPROG register.
- 5. Derived from ADCCLK.

6. Internal reference option used corresponds to selection 2V5 in the SINGLECTRL_REF or SCANCTRL_REF register field. The differential input range with this configuration is ± 1.25 V. Typical value is characterized using full-scale sine wave input. Minimum value is production-tested using sine wave input at 1.5 dB lower than full scale.

7. External reference is 1.25 V applied externally to ADCnEXTREFP, with the selection CONF in the SINGLECTRL_REF or SCANCTRL_REF register field and VREFP in the SINGLECTRLX_VREFSEL or SCANCTRLX_VREFSEL field. The differential input range with this configuration is \pm 1.25 V.

4.1.17 Analog Comparator (ACMP)

Table 4.43. Analog Comparator (ACMP)

4.1.18 Digital to Analog Converter (VDAC)

DRIVESTRENGTH = 2 unless otherwise specified. Primary VDAC output.

1. In differential mode, the output is defined as the difference between two single-ended outputs. Absolute voltage on each output is limited to the single-ended range.

2.Supply current specifications are for VDAC circuitry operating with static output only and do not include current required to drive the load.

- 3. Current from HFPERCLK is dependent on HFPERCLK frequency. This current contributes to the total supply current used when the clock to the DAC peripheral is enabled in the CMU.
- 4. PSRR calculated as 20 $*$ log₁₀(Δ VDD / Δ V_{OUT}), VDAC output at 90% of full scale
- 5. Entire range is monotonic and has no missing codes.
- 6. Gain is calculated by measuring the slope from 10% to 90% of full scale. Offset is calculated by comparing actual VDAC output at 10% of full scale to ideal VDAC output at 10% of full scale with the measured gain.

4.1.19 Current Digital to Analog Converter (IDAC)

Table 4.45. Current Digital to Analog Converter (IDAC)

1. In IDAC_CURPROG register.

2. The IDAC is supplied by either AVDD, DVDD, or IOVDD based on the setting of ANASW in the EMU_PWRCTRL register and PWRSEL in the IDAC_CTRL register. Setting PWRSEL to 1 selects IOVDD. With PWRSEL cleared to 0, ANASW selects between AVDD (0) and DVDD (1).

4.1.20 Capacitive Sense (CSEN)

Table 4.46. Capacitive Sense (CSEN)

1. Current is specified with a total external capacitance of 33 pF per channel. Average current is dependent on how long the peripheral is actively sampling channels within the scan period, and scales with the number of samples acquired. Supply current for a specific application can be estimated by multiplying the current per sample by the total number of samples per period (total_current = single_sample_current * (number_of_channels * accumulation)).

4.1.21 Operational Amplifier (OPAMP)

Unless otherwise indicated, specified conditions are: Non-inverting input configuration, VDD = 3.3 V, DRIVESTRENGTH = 2, MAIN-OUTEN = 1, C_{LOAD} = 75 pF with OUTSCALE = 0, or C_{LOAD} = 37.5 pF with OUTSCALE = 1. Unit gain buffer and 3X-gain connection as specified in table footnotes^{[1](#page-100-0)[2](#page-100-0)}.

Table 4.47. Operational Amplifier (OPAMP)

4.1.22 Pulse Counter (PCNT)

Table 4.48. Pulse Counter (PCNT)

4.1.23 Analog Port (APORT)

Table 4.49. Analog Port (APORT)

Note:

1. Supply current increase that occurs when an analog peripheral requests access to APORT. This current is not included in reported peripheral currents. Additional peripherals requesting access to APORT do not incur further current.

2.Specified current is for continuous APORT operation. In applications where the APORT is not requested continuously (e.g. periodic ACMP requests from LESENSE in EM2), the average current requirements can be estimated by mutiplying the duty cycle of the requests by the specified continuous current number.

4.1.24 I2C

4.1.24.1 I2C Standard-mode (Sm)¹

Table 4.50. I2C Standard-mode (Sm)¹

Note:

1. For CLHR set to 0 in the I2Cn_CTRL register.

2. For the minimum HFPERCLK frequency required in Standard-mode, refer to the I2C chapter in the reference manual.

3. The maximum SDA hold time ($t_{HD\ DA}$) needs to be met only when the device does not stretch the low time of SCL (t_{LOW}).

4.1.24.2 I2C Fast-mode (Fm)¹

Table 4.51. I2C Fast-mode (Fm)¹

Note:

1. For CLHR set to 1 in the I2Cn_CTRL register.

2. For the minimum HFPERCLK frequency required in Fast-mode, refer to the I2C chapter in the reference manual.

3. The maximum SDA hold time $(t_{HD, DAT})$ needs to be met only when the device does not stretch the low time of SCL (t_{LOW}).

4.1.24.3 I2C Fast-mode Plus (Fm+)¹

Table 4.52. I2C Fast-mode Plus (Fm+)¹

Note:

1. For CLHR set to 0 or 1 in the I2Cn_CTRL register.

2. For the minimum HFPERCLK frequency required in Fast-mode Plus, refer to the I2C chapter in the reference manual.

4.1.25 USART SPI

SPI Master Timing

Table 4.53. SPI Master Timing

Note:

1.Applies for both CLKPHA = 0 and CLKPHA = 1 (figure only shows CLKPHA = 0).

2. Measurement done with 8 pF output loading at 10% and 90% of V_{DD} (figure shows 50% of V_{DD}).

3. t_{HFPERCLK} is one period of the selected HFPERCLK.

Figure 4.1. SPI Master Timing Diagram

SPI Slave Timing

Table 4.54. SPI Slave Timing

Note:

1.Applies for both CLKPHA = 0 and CLKPHA = 1 (figure only shows CLKPHA = 0).

2. Measurement done with 8 pF output loading at 10% and 90% of V_{DD} (figure shows 50% of V_{DD}).

3. t_{HFPERCLK} is one period of the selected HFPERCLK.

Figure 4.2. SPI Slave Timing Diagram

4.2 Typical Performance Curves

Typical performance curves indicate typical characterized performance under the stated conditions.

Figure 4.3. EM0 Active Mode Typical Supply Current vs. Temperature

Figure 4.4. EM1 Sleep Mode Typical Supply Current vs. Temperature

Typical supply current for EM2, EM3 and EM4H using standard software libraries from Silicon Laboratories.

Figure 4.5. EM2, EM3, EM4H and EM4S Typical Supply Current vs. Temperature

Figure 4.6. EM0 and EM1 Mode Typical Supply Current vs. Supply

Typical supply current for EM2, EM3 and EM4H using standard software libraries from Silicon Laboratories.

Figure 4.7. EM2, EM3, EM4H and EM4S Typical Supply Current vs. Supply

4.2.2 DC-DC Converter

Default test conditions: CCM mode, LDCDC = 4.7 μH, CDCDC = 4.7 μF, VDCDC_I = 3.3 V, VDCDC_O = 1.8 V, FDCDC_LN = 7 MHz

Figure 4.8. DC-DC Converter Typical Performance Characteristics

Figure 4.9. DC-DC Converter Transition Waveforms

4.2.3 2.4 GHz Radio

Figure 4.10. 2.4 GHz RF Transmitter Output Power

Figure 4.11. 2.4 GHz RF Receiver Sensitivity

5. Typical Connection Diagrams

5.1 Power

Typical power supply connections for direct supply, without using the internal DC-DC converter, are shown in the following figure.

Figure 5.1. EFR32FG12 Typical Application Circuit: Direct Supply Configuration without DC-DC converter

Typical power supply circuits using the internal DC-DC converter are shown below. The MCU operates from the DC-DC converter supply. For low RF transmit power applications less than 13dBm, the RF PA may be supplied by the DC-DC converter. For OPNs supporting high power RF transmission, the RF PA must be directly supplied by VDD for RF transmit power greater than 13 dBm.

Figure 5.2. EFR32FG12 Typical Application Circuit: Configuration with DC-DC converter (PAVDD from VDCDC)

Figure 5.3. EFR32FG12 Typical Application Circuit: Configuration with DC-DC converter (PAVDD from VDD)

5.2 RF Matching Networks

Typical RF matching network circuit diagrams are shown in Figure 5.4 Typical 2.4 GHz RF impedance-matching network circuits on page 118 for applications in the 2.4GHz band, and in Figure 5.5 Typical Sub-GHz RF impedance-matching network circuits on page 118 for applications in the sub-GHz band. Application-specific component values can be found in application notes *AN923: EFR32 sub-GHz Matching Guide* and *AN930: EFR32 2.4 GHz Matching Guide*. For low RF transmit power applications less than 13dBm, the twoelement match is recommended. For OPNs supporting high power RF transmission, the four-element match is recommended for high RF transmit power (> 13dBm).

Figure 5.4. Typical 2.4 GHz RF impedance-matching network circuits

Sub-GHz Match Topology 2 (500-915 MHz)

Figure 5.5. Typical Sub-GHz RF impedance-matching network circuits

5.3 Other Connections

Other components or connections may be required to meet the system-level requirements. Application Note AN0002: "Hardware Design Considerations" contains detailed information on these connections. Application Notes can be accessed on the Silicon Labs website (www.silabs.com/32bit-appnotes).

6. Pin Definitions

6.1 BGA125 2.4 GHz and Sub-GHz Device Pinout

Figure 6.1. BGA125 2.4 GHz and Sub-GHz Device Pinout

Note:

6.2 BGA125 2.4 GHz Device Pinout

Figure 6.2. BGA125 2.4 GHz Device Pinout

6.3 BGA125 Sub-GHz Device Pinout

Figure 6.3. BGA125 Sub-GHz Device Pinout

Figure 6.4. QFN68 2.4 GHz and Sub-GHz Device Pinout

Figure 6.5. QFN68 Sub-GHz Device Pinout

Note:

Figure 6.6. QFN48 2.4 GHz and Sub-GHz Device Pinout

Figure 6.7. QFN48 2.4 GHz Device Pinout

Figure 6.8. QFN48 Sub-GHz Device Pinout

6.9 GPIO Functionality Table

A wide selection of alternate functionality is available for multiplexing to various pins. The following table shows the name of each GPIO pin, followed by the functionality available on that pin. Refer to [6.10 Alternate Functionality Overview](#page-183-0) for a list of GPIO locations available for each function.

Table 6.9. GPIO Functionality Table

6.10 Alternate Functionality Overview

A wide selection of alternate functionality is available for multiplexing to various pins. The following table shows the name of the alternate functionality in the first column, followed by columns showing the possible LOCATION bitfield settings and the associated GPIO pin. Refer to [6.9 GPIO Functionality Table](#page-140-0) for a list of functions available on each GPIO pin.

Note: Some functionality, such as analog interfaces, do not have alternate settings or a LOCATION bitfield. In these cases, the pinout is shown in the column corresponding to LOCATION 0.

Table 6.10. Alternate Functionality Overview

6.11 Analog Port (APORT) Client Maps

The Analog Port (APORT) is an infrastructure used to connect chip pins with on-chip analog clients such as analog comparators, ADCs, DACs, etc. The APORT consists of a set of shared buses, switches, and control logic needed to configurably implement the signal routing. Figure 6.9 APORT Connection Diagram on page 199 shows the APORT routing for this device family (note that available features may vary by part number). A complete description of APORT functionality can be found in the Reference Manual.

Client maps for each analog circuit using the APORT are shown in the following tables. The maps are organized by bus, and show the peripheral's port connection, the shared bus, and the connection from specific bus channel numbers to GPIO pins.

In general, enumerations for the pin selection field in an analog peripheral's register can be determined by finding the desired pin connection in the table and then combining the value in the Port column (APORT__), and the channel identifier (CH__). For example, if pin

PF7 is available on port APORT2X as CH23, the register field enumeration to connect to PF7 would be APORT2XCH23. The shared bus used by this connection is indicated in the Bus column.

Table 6.11. ACMP0 Bus and Pin Mapping

Table 6.12. ACMP1 Bus and Pin Mapping

Table 6.13. ADC0 Bus and Pin Mapping

Table 6.15. IDAC0 Bus and Pin Mapping

Table 6.16. VDAC0 / OPA Bus and Pin Mapping

7. BGA125 Package Specifications

7.1 BGA125 Package Dimensions

Figure 7.1. BGA125 Package Drawing

Table 7.1. BGA125 Package Dimensions

Note:

1.All dimensions shown are in millimeters (mm) unless otherwise noted.

2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.

7.2 BGA125 PCB Land Pattern

Figure 7.2. BGA125 PCB Land Pattern Drawing

Table 7.2. BGA125 PCB Land Pattern Dimensions

Note:

1.All dimensions shown are in millimeters (mm) unless otherwise noted.

2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.

3. This Land Pattern Design is based on the IPC-7351 guidelines.

4. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60 µm minimum, all the way around the pad.

5. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.

6. The stencil thickness should be 0.125 mm (5 mils).

7. The ratio of stencil aperture to land pad size should be 1:1.

8. A No-Clean, Type-3 solder paste is recommended.

9. The recommended card reflow profile is per the JEDEC/IPC J-STD-020C specification for Small Body Components.

7.3 BGA125 Package Marking

Figure 7.3. BGA125 Package Marking

The package marking consists of:

- PPPPPPPPP The part number designation.
	- 1. Family Code (B | M | F)
	- 2. G (Gecko)
	- 3. Series (1, 2,...)
	- 4. Device Configuration (1, 2,...)
	- 5. Performance Grade (P | B | V)
	- 6. Feature Code (1, 2,...)
	- 7. TRX Code (3 = TXRX | 2= RX | 1 = TX)
	- 8. Band (1 = Sub-GHz | 2 = 2.4 GHz | 3 = Dual-band)
	- 9. Flash (J = 1024K | H = 512K | G = 256K | F = 128K | E = 64K | D = 32K)
	- 10. Temperature Grade (G = -40 to 85 | I = -40 to 125)
- YY The last 2 digits of the assembly year.
- WW The 2-digit workweek when the device was assembled.
- TTTTTT A trace or manufacturing code. The first letter is the device revision.

8. QFN48 Package Specifications

8.1 QFN48 Package Dimensions

Figure 8.1. QFN48 Package Drawing

Table 8.1. QFN48 Package Dimensions

Note:

1.All dimensions shown are in millimeters (mm) unless otherwise noted.

2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.

3. This drawing conforms to the JEDEC Solid State Outline MO-220, Variation VKKD-4.

4. Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

8.2 QFN48 PCB Land Pattern

Figure 8.2. QFN48 PCB Land Pattern Drawing

Table 8.2. QFN48 PCB Land Pattern Dimensions

Note:

1.All dimensions shown are in millimeters (mm) unless otherwise noted.

2. This Land Pattern Design is based on the IPC-7351 guidelines.

3. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60 µm minimum, all the way around the pad.

4. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.

5. The stencil thickness should be 0.125 mm (5 mils).

6. The ratio of stencil aperture to land pad size can be 1:1 for all perimeter pads.

7. A 4x4 array of 0.75 mm square openings on a 1.00 mm pitch can be used for the center ground pad.

8. A No-Clean, Type-3 solder paste is recommended.

9. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.
8.3 QFN48 Package Marking

Figure 8.3. QFN48 Package Marking

The package marking consists of:

- PPPPPPPPP The part number designation.
	- 1. Family Code (B | M | F)
	- 2. G (Gecko)
	- 3. Series (1, 2,...)
	- 4. Device Configuration (1, 2,...)
	- 5. Performance Grade (P | B | V)
	- 6. Feature Code (1, 2,...)
	- 7. TRX Code (3 = TXRX | 2 = RX | 1 = TX)
	- 8. Band (1 = Sub-GHz | 2 = 2.4 GHz | 3 = Dual-band)
	- 9. Flash (J = 1024K | H = 512K | G = 256K | F = 128K | E = 64K | D = 32K)
	- 10. Temperature Grade (G = -40 to 85 | I = -40 to 125)
- YY The last 2 digits of the assembly year.
- WW The 2-digit workweek when the device was assembled.
- TTTTTT A trace or manufacturing code. The first letter is the device revision.

9. QFN68 Package Specifications

9.1 QFN68 Package Dimensions

Table 9.1. QFN68 Package Dimensions

Note:

1.All dimensions shown are in millimeters (mm) unless otherwise noted.

2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.

3. This drawing conforms to the JEDEC Solid State Outline MO-220, Variation VKKD-4.

4. Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

9.2 QFN68 PCB Land Pattern

Table 9.2. QFN68 PCB Land Pattern Dimensions

Note:

1.All dimensions shown are in millimeters (mm) unless otherwise noted.

2. This Land Pattern Design is based on the IPC-7351 guidelines.

3. All dimensions shown are at Maximum Material Condition (MMC). Least Material Condition (LMC) is calculated based on a Fabrication Allowance of 0.05mm.

4. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60 µm minimum, all the way around the pad.

5. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.

6. The stencil thickness should be 0.100 mm (4 mils).

7. The ratio of stencil aperture to land pad size can be 1:1 for all pads.

8. A 3x3 array of 1.50 mm square openings on a 1.80 mm pitch can be used for the center ground pad.

9. A No-Clean, Type-3 solder paste is recommended.

10. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

9.3 QFN68 Package Marking

Figure 9.3. QFN68 Package Marking

The package marking consists of:

- PPPPPPPPPP The part number designation.
- TTTTTT A trace or manufacturing code. The first letter is the device revision.
- YY The last 2 digits of the assembly year.
- WW The 2-digit workweek when the device was assembled.

10. Revision History

Revision 1.7

June, 2021

• Added Wi-SUN protocol in [1. Feature List](#page-1-0) and [Table 2.1 Ordering Information on page 3](#page-2-0).

Revision 1.6

September, 2020

• Updated External PA supply connection conditions for SPUR_{HARM}_ETSI and SPUR_{OOB}_ETSI in section [4.1.10.3 Sub-GHz RF Trans](#page-59-0)[mitter characteristics for 868 MHz Band.](#page-59-0)

Revision 1.5

December, 2019

- In the front page block diagram, updated the lowest energy mode for LETIMER.
- Updated [3.6.4 Low Energy Timer \(LETIMER\)](#page-13-0) lowest energy mode.
- In [5.2 RF Matching Networks](#page-117-0), corrected document references for component values and added document reference to IPD solutions.
- Corrected the max value of the Absolute voltage on 2G4RF_IOP, 2G4RF_ION, SUBGRF_OP and SUBGRF_ON pins in Absolute Maximum Ratings table in [4.1.1 Absolute Maximum Ratings](#page-21-0).
- Added a Note about the operating voltage in [3.8.3 True Random Number Generator \(TRNG\).](#page-14-0)

Revision 1.4

July, 2019

- In Feature List, removed wake on radio.
- In System Overview:
	- Replaced reference to "modules" with "peripherals" or "blocks"
	- Renamed GPCRC section
- In Electrical Specifications, reordered footnotes according to order of appearance in the table.
- In General Operating Conditions for f_{CORF} :
	- Added conditions for all usable wait state settings
	- Corrected maximum specification from 20 MHz to 7 MHz for test condition VSCALE0, MODE = WS0
- In sub-GHz specifications, replaced references to "PAVDD" with "External PA Supply" for clarity.
- In RF Transmitter Characteristics for 802.15.4 DSSS-OQPSK in the 2.4GHz Band, updated parameter for EVM.
- In Sub-GHz RF Transmitter characteristics for 915 MHz Band:
	- Corrected test conditions for:
		- SPUR_{HARM FCC 14}, in non-restricted bands
		- SPUR_{OOB FCC} $_{14}$, in non-restricted bands
		- SPUR_{HARM FCC 20}, in non-restricted bands
		- SPUROOB FCC 20, in non-restricted bands
	- Updated typical specification from -52 dBm to -62 dBm and maximum specification from -46 dBm to -56 dBm for:
		- SPUR_{OOB FCC} 20, in restricted bands (30-88 MHz)
		- SPUR_{OOB FCC} $_{14}$, in restricted bands (30-88 MHz)
	- Added footnote to PSD.
- In [4.1.10.2 Sub-GHz RF Receiver Characteristics for 915 MHz Band](#page-55-0), updated typical specification from -60 dBm to -61 dBm and maximum specification from -54 dBm to -55 dBm for SPUR_{RX} ARIB, 930-1000 MHz, RBW=100 kHz.
- Corrected units for F_{RANGE} in:
	- Sub-GHz RF Receiver Characteristics for 490 MHz Band
	- Sub-GHz RF Receiver Characteristics for 315 MHz Band
	- Sub-GHz RF Receiver Characteristics for 169 MHz Band
- In LFRCO, updated test conditions for fLFRCO.
- In GPIO, added footnotes to V_{II} and V_{IH} .
- \cdot In VMON, updated test conditions for I_{VMON} .
- In VDAC, updated test conditions for I_{DAC} , 200 Hz refresh rate.
- In CSEN, updated test conditions for:
	- \cdot C_{EXTMAX}
	- I_{CSEN} BOND
	- I_{CSEN} EM2
	- ICSEN ACTIVE
- In Pin Definitions, updated pin descriptions for DECOUPLE and RESETn.
- Updated feature code in:
	- [7.3 BGA125 Package Marking](#page-211-0)
	- [8.3 QFN48 Package Marking](#page-216-0)

Revision 1.3

June, 2018

- [4.1.5.4 Current Consumption Using Radio 3.3 V with DC-DC:](#page-34-0) Updated typical 802.15.4 receive current specifications.
- [Table 6.9 GPIO Functionality Table on page 141](#page-140-0): Changed presentation to order table by pin name instead of pin location.

Revision 1.2

February, 2018

- Added new orderable part numbers for QFN68 variants and associated packaging and pinout information.
- **Absolute Maximum Ratings Table**: Added footnote to clarify IOVDD over-voltage operation conditions.
- **APORT Connection Diagram**: Corrected OPA output connections to route through "Y" buses.

Revision 1.1

October, 2017

- Updated **Ordering Table** to revision-C OPNs.
- Added high-temperature part numbers to **Ordering Table** and added associated specifications / content throughout document.
- Updated product highlights on **Front Page** and **Feature List** for consistency across EFR32xG1x family documentation.
- **System Overview Updates**
	- Expanded Receiver Architecture section.
	- Clarified / corrected energy mode mentions in RTCC and Opamp sections.
	- Memory maps updated with LE peripherals and new formatting.
- **Absolute Maximum Ratings** Table:
	- Removed redundant I_{VSSMAX} line.
	- Added footnote to clarify V_{DIGPIN} specification for 5V tolerant GPIO.
- **General Operating Conditions** Table:
	- Removed redundant footnote about shorting VREGVDD and AVDD together.
	- Added footnote about IOVDD voltage restriction when CSEN peripheral is used with chopping enabled.
	- Added footnote for additional information on peak current during voltage scaling operations.

• **RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate** Table:

- Sensitivity, Co-channel interferer and Selectivity typical numbers updated to latest phy characterization data.
- BLOCK_{OOB} specifications changed to show Min values instead of Typ.
- **RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate** Table:
	- SAT Typical value corrected from 5 to 10 dBm.
	- BLOCK_{OOB} specifications removed.
- RF Receiver Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band Table: Footnote added to BLOCK_{80211G} specification to clarify blocker signal definition.
- **Sub-GHz RF Receiver Characteristics for 915 MHz Band** Table: Added O-QPSK DSSS phy specifications.
- Sub-GHz RF Transmitter Characteristics for 868 MHz Band Table: SPUR_{OOB ETSI} below 1 GHz Typ corrected from -60 to -42 dBm.
- **Sub-GHz RF Receiver Characteristics for 490 MHz Band** Table: Corrected 10 kbps 2GFSK reference signal bandwidth to 20.038 kHz.
- **Flash Memory Characteristics** Table:
	- Added timing measurement clarification for Device Erase and Mass Erase.
	- Device Erase Time typical values corrected from 69 to 82 ms.
- **Analog to Digital Converter (ADC)** Table:
	- Added header text for general specification conditions.
	- Added footnote for clarification of input voltage limits.
- **Digital to Analog Converter (VDAC)** Table: Gain Error min/max specifications relaxed for REFSEL on 1V25LN, VDD, and EXT settings.
- **Current Digital to Analog Converter (IDAC)** Table: Total accuracy STEPSEL value setting corrected from 0x80 to 0x10.
- **Analog Port (APORT)** Table: Operation in EM2/EM3 supply current changed from 915 to 67 nA (silicon fix from rev B to C).
- **2.4 GHz RF Transmitter Output Power** Figure: Extended temperature range to 125 C.
- **2.4 GHz RF Receiver Sensitivity** Figure: Updated with latest characterization data and added 125 C operational plots.
- **Typical Sub-GHz Impedance-matching network circuits** Figure: Corrected split between two examples from 450 MHz to 500 MHz.
- Minor typographical corrections, including capitalization, mis-spellings and punctuation marks, throughout document.
- Minor formatting and styling updates, including table formats, TOC location, and boilerplate information throughout document.

Revision 1.0

2017-04-14

- Added Thermal Characteristics table.
- Finalized specification tables. All tables were updated with latest characterization data and production test limits.
- Updated typical performance graphs for DC-DC.
- Minor typographical, clarity, and consistency improvements.
- Condensed pin function tables with new formatting.

Revision 0.6

2017-02-23

- Updated 2 Mbps 2GFSK receiver specifications with latest characteriztion data.
- Added table-wide conditions to 2GFSK 1 Mbps and 2 Mbps receiver tables.
- Clarified opamp noise measurement conditions in electrical spec table.

Revision 0.5

2017-02-03

- New corporate stylesheet applied.
- Updated device block diagrams on front page and in System Overview.
- Updated Feature List with latest characterization numbers.
- "Bluetooth Smart" changed to "Bluetooth Low Energy" throughout document.
- All OPNs changed to revision B.
- Minor typographical corrections and clarifications in System Overview.
- Electrical Characteristics Table Changes
	- All specification tables updated with latest characterization data and production test limits.
	- Split 2.4 GHz 2GFSK tables into separate tables for 1 Mbps and 2 Mbps data rates.
	- Split HFRCO/AUXHFRCO table into separate tables for HFRCO and AUXHFRCO.
	- OPAMP, CSEN, and VDAC specification line items updated to match test conditions.
	- Added tables for Analog Port (APORT) and Pulse Counter (PCNT).
- Added Typical Performance Curves for supply current, DCDC, and RF parameters.
- Added missing alternate functions and descriptions to Pinout and Alternate Function tables.
- Added APORT Connection Diagram.
- Corrected Package Marking description for QFN48 and BGA125.
- Corrected Package Marking diagram for QFN48.

Revision 0.2

2016-09-21

Initial release.