

[ADIN1110](http:/www.analog.com/ADIN1110.html)

Robust, Industrial, Low Power 10BASE-T1L Ethernet MAC-PHY

FEATURES

- ► 10BASE-T1L IEEE Standard 802.3cg-2019 compliant
- ► Cable reach up to 1700 m with 1.0 V/2.4 V
- ► Integrated MAC with SPI
	- ► Supports OPEN Alliance 10BASE-T1x MAC-PHY serial interface
	- ► 16 MAC address filters
	- ► High and low priority queues with 28 kB buffer
	- ► Cut through or store forward operation
	- ► IEEE 1588 timestamp support
	- ► Statistics counters
- ► Low power consumption: 42 mW (dual-supply, 1.0 V p-p)
- ► Diagnostics
	- ► Cable fault detection with TDR
	- \blacktriangleright Link quality indicator with MSE
	- ► Frame generator and checker
	- ► Multiple loopback modes
	- ► IEEE test mode support
- ► Supports 1.0 V p-p and 2.4 V p-p transmit levels
- ► 4-pin MDI (RXN, RXP, TXN, and TXP)
- ► Suitable for intrinsic safety applications
- ► External termination resistors
- ► Autonegotiation
- ► 25 MHz crystal or external clock input
- ► Electromagnetic compatibility (EMC) test standards
	- \blacktriangleright IEC 61000-4-4 electrical fast transient (\pm 4 kV)
	- ► IEC 61000-4-2 ESD (±4 kV contact discharge)
	- ► IEC 61000-4-2 ESD (±8 kV air discharge)
	- \blacktriangleright IEC 61000-4-6 conducted immunity (10 V/m)
	- ► IEC 61000-4-5 surge (±4 kV)
	- ► IEC 61000-4-3 radiated immunity (Class A)
	- ► EN55032 radiated emissions (Class B)
- ► Small package: [40-lead \(6 mm × 6 mm\) LFCSP](#page--1-0)
- ► Temperature range
	- ► Industrial: −40°C to +85°C
	- ► Extended: −40°C to +105°C

APPLICATIONS

- ► Field instruments
- ► Building automation and fire safety
- ► Factory automation
- ► Edge sensors and actuators
- ► Condition monitoring and machine connectivity

Rev. A

[DOCUMENT FEEDBACK](https://form.analog.com/Form_Pages/feedback/documentfeedback.aspx?doc=ADIN1110.pdf&product=ADIN1110&rev=A) [TECHNICAL SUPPORT](http://www.analog.com/en/content/technical_support_page/fca.html)

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FUNCTIONAL BLOCK DIAGRAM

Figure 1.

GENERAL DESCRIPTION

The ADIN1110 is an ultra low power, single port, 10BASE-T1L transceiver design for industrial Ethernet applications and is compliant with the IEEE® 802.3cg-2019™ Ethernet standard for long reach, 10 Mbps single pair Ethernet (SPE). Featuring an integrated media access control (MAC) interface, the ADIN1110 enables direct connectivity with a variety of host controllers via a 4-wire serial peripheral interface (SPI). This SPI enables the use of lower power processors without an integrated MAC, which provides for the lowest overall system level power consumption. The SPI can be configured to use the Open Alliance SPI protocol or a generic SPI protocol.

The programmable transmit levels, external termination resistors, and independent receive and transmit pins make the ADIN1110 suited to intrinsic safety applications.

The ADIN1110 has an integrated voltage supply monitoring and power-on reset (POR) circuitry to improve system level robustness.

The ADIN1110 is available in a 40 -lead, 6 mm \times 6 mm lead frame [chip scale package \(LFCSP\).](#page--1-0)

REVISION HISTORY

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10/2021—Revision 0: Initial Version

SPECIFICATIONS

AVDD_H = AVDD_L = VDDIO = 3.3 V, DVDD_1P1 supplied from internal low dropout (LDO) regulator (DVDD_1P1 = DLDO_1P1), and all specifications are at −40°C to +105°C, unless otherwise noted.

SPECIFICATIONS

Table 1. General Specifications (Continued)

 $1 \text{ } C_L = ((C1 \times C2)/(C1 + C2) + C_{STRAY})$, where C_{STRAY} is the stray capacitance including routing and package parasitics.

 2 R_p and C_p are the values of the equivalent parallel RC circuit to ac ground (R_p||R_c), modeling the driving point impedance of the XTAL_I/CLK_IN pin.

Table 2. 10BASE-T1L Specifications

TIMING CHARACTERISTICS

POWER-UP TIMING

Table 3. Power-Up Timing

¹ The minimum time interval is referenced to the last supply to reach its rising threshold. There is no specific power supply sequencing required.

SPI

¹ Guaranteed by design and characterization. Not production tested.

² All input signals are specified with rise time (t_R) = fall time (t_F) = 5 ns (10% to 90% of VDDIO) and timed from a voltage level of 1.2 V.

 \approx

TIMING CHARACTERISTICS

³ Capacitive load on the SDO pin is 10 pF.

Figure 3. Serial Interface Timing Diagram

ABSOLUTE MAXIMUM RATINGS

 T_A = 25°C, unless otherwise noted.

Table 5.

See the [Pin Configuration and Function Descriptions](#page-8-0) section for a full list of SPI pins.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

THERMAL RESISTANCE

 θ_{JA} is the natural convection junction to ambient thermal resistance measured in a one cubic foot sealed enclosure.

Table 6. Thermal Resistance

 $1 - \theta_{JA}$ is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

² Test Condition 1: thermal impedance simulated values are based on a JE-DEC 2S2P thermal test board with thermal vias. See JEDEC JESD51.

ELECTROSTATIC DISCHARGE (ESD) RATINGS

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

Human body model (HBM) per ANSI/ESDA/JEDEC JS-001.

Field induced charged device model (FICDM) per ANSI/ESDA/JE-DEC JS-002.

ESD Ratings for ADIN1110

Table 7. ADIN1110, 40-Lead LFCSP

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

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PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

2. DNC = DO NOT CONNECT. THESE PINS MUST BE LEFT OPEN CIRCUIT.

Figure 4. Pin Configuration

Table 8. Pin Function Descriptions (Hardware Pin Configuration Groupings Are Subject to Change)

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

Table 8. Pin Function Descriptions (Hardware Pin Configuration Groupings Are Subject to Change) (Continued)

¹ Where a pin is shared between a functional signal and a hardware pin configuration, the hardware pin configuration signal is listed last and the pin is referred to using the functional signals name throughout the data sheet.

² All of the hardware configuration pins have internal pull-down resistors. The default mode of operation without any external resistors connected to these pins is shown in [Table 11](#page-16-0). If an alternative mode of operation is required, use 4.7 kΩ pull-up resistors.

TYPICAL PERFORMANCE CHARACTERISTICS

Figure 6. Power vs. Temperature, AVDD_H = 3.3 V, VDDIO = 3.3 V, AVDD_L = 1.8 V, Internal LDO Circuit

Figure 7. Power vs. Temperature, 3.3 V Single Supply, Internal LDO Circuit

Figure 8. Power vs. Temperature, AVDD_H = 3.3 V, AVDD_L = VDDIO = 1.8 V, Internal LDO Circuit

Figure 9. Power vs. Temperature, AVDD_H = AVDD_L = 3.3 V, VDDIO = 1.8 V, Internal LDO Circuit

The ADIN1110 is a low power, single port 10 Mb/s Ethernet MAC-PHY device, compliant with the IEEE 802.3cg Ethernet standard for long reach, 10 Mbps single pair Ethernet.

The ADIN1110 integrates the following features:

- ► 10BASE-T1L Ethernet PHY core and MAC with all the associated common analog circuitry
- \blacktriangleright Input and output clock buffering
- ► SPI and subsystem registers
- ► Control logic to manage the reset and clock
- ► Hardware configuration pins
- ► Two configurable LED pins

POWER SUPPLY DOMAINS

The ADIN1110 has four power supply domains and requires a minimum of one supply rail, as follows:

- \triangleright AVDD H is the analog power supply input for the analog front end (AFE) circuitry in the ADIN1110.
- ► AVDD L is the analog supply voltage for the internal LDO circuits. AVDD_L can be connected to the AVDD_H rail when in single supply mode, or to an alternative lower voltage rail when the device is configured with dual supplies for lower power consumption.
- ► DVDD 1P1 is the 1.1 V digital core power supply input. DVDD 1P1 can operate from an internal 1.1 V LDO output coming from the DLDO_1P1 pin to the DVDD_1P1 pin. Alternatively, DVDD 1P1 can be driven from an external 1.1 V supply for greater power efficiency.
- ► VDDIO enables the SPI voltage supply to be configured independently of the other circuitry on the ADIN1110. VDDIO can be connected directly to the AVDD_L rail.

In a single supply application, connect AVDD_H = AVDD_L = VDDIO and use the internal 1.1 V LDO circuit for DVDD_1P1. The appropriate supply voltage used depends on the end application and cable length. For long reach/trunk applications, the higher transmit amplitude of 2.4 V p-p requires AVDD_H = 3.3 V, whereas spur applications can use a lower transmit amplitude of 1.0 V p-p with an $AVDD_H = 1.8$ V.

ANALOG FRONT END

The AFE stage consists of a hybrid stage, 9-level DAC, line driver, analog receive filter, input buffer, and ADC.

The line driver transmits the signal onto the line via the MDI interface pins, TXP and TXN. The hybrid stage subtracts the transmitted signal from the received signal on the MDI pins, allowing full duplex operation on the single-pair cable.

The received signal then goes through the analog receive filter and reaches the input buffer before sending it to the ADC.

MAC

The MAC included in the ADIN1110 supports 16 different MAC addresses. The MAC also has one low priority receive first in, first out (FIFO), one high priority receive FIFO and one transmit FIFO. These FIFOs can ship data in store and forward mode when using the generic SPI protocol, and in either store and forward or cut through mode when using the OPEN Alliance protocol.

A generic version and OPEN Alliance version of the SPI protocol are available. The data is transferred over the SPI half duplex using the generic SPI protocol and full duplex using the OPEN Alliance SPI protocol. See the [MAC SPI](#page-29-0) section.

Interrupt (INT)

The ADIN1110 is capable of generating an interrupt to a host processor using the $\overline{\text{INT}}$ pin in response to a variety of user selectable conditions. The following conditions can be selected to generate an interrupt:

- ► Link status change
- ► Receive FIFO data available
- ► Frame transmit complete or transmit space available
- ► Time stamp captured
- ► Operation error detected
- ► PHY related interrupts

When an interrupt occurs, the system can poll the MAC status registers (STATUS0 and STATUS1) to determine the origin of the interrupt.

TRANSMIT AMPLITUDE CONFIGURATION

The ADIN1110 supports two transmit amplitude modes of operation as follows:

- \blacktriangleright 1.0 V p-p and 2.4 V p-p mode (high level)
- ► 1.0 V p-p only mode

The high level transmission mode allows the ADIN1110 to support both voltage levels. The operating level can then be automatically configured during autonegotiation (if enabled) based on the link partner capabilities. Note that in high level transmit operating mode, AVDD H must be supplied with 2.4 V or higher for the device to work properly.

The mode of operation is configured through the TX2P4_EN hardware configuration pin (see the [Transmit Amplitude](#page-17-0) section). The ADIN1110 also configures the default value of the transmit level register bits used for the autonegotiation process based on the level configured on that pin.

The ADIN1110 is configured in high level transmit operating mode by default if the $\overline{T}X2P4$ EN pin is left floating (internal pull-down resistor).

MASTER/SLAVE CONFIGURATION

The 10BASE-T1L standard uses a master/slave clock scheme. This scheme is commonly used in full duplex transceiver standards with echo cancellation.

On a 10BASE-T1L link, one PHY is designated as the master, and the other PHY as the slave. Autonegotiation is used to determine which PHY is the master and which is the slave. Master and slave assignment does not generally matter.

Software Configuration

The master and slave configuration bit (CFG_MST) is used to configure the PHY role. This bit is only used when autonegotiation is disabled. Otherwise, this bit is set or reset during the autonegotiation process (see the Autonegotiation section).

Table 9. CFG_MST Settings

AUTONEGOTIATION

The ADIN1110 uses the autonegotiation capability in accordance with IEEE 802.3 Clause 98, providing a mechanism for exchanging information between PHYs to allow link partners to agree to a common mode of operation. During the autonegotiation process, the PHY advertises its own capabilities and compares to those received from the link partner. The concluded operating mode is the transmit amplitude mode and master/slave preference common across the two devices.

If a link is dropped, the autonegotiation process restarts automatically. An autonegotiation restart can also be requested by writing to the autonegotiation restart bit (AN_RESTART) in the autonegotiation control register.

The autonegotiation process takes some time to complete, depending on the number of pages exchanged, but is always the fastest way to bring up a link. Clause 98 of the IEEE 802.3 standard details the timers related to autonegotiation.

Note that autonegotiation is enabled by default for the ADIN1110, and it is strongly recommended that autonegotiation is always enabled.

Transmit Amplitude Resolution

Autonegotiation is used to resolve the transmit amplitude resolution. The PHY can be configured to support both 1.0 V p-p and 2.4 V p-p transmit levels or to operate with 1.0 V p-p transmit level only through the hardware configuration (see [Table 15\)](#page-17-0). This configuration can also be done in software using the 10BASE-T1L high level transmit operating mode ability (AN_ADV_B10L_TX_LVL_HI_ABL)

and 10BASE-T1L high level transmit operating mode request (AN_ADV_B10L_TX_LVL_HI_REQ) register bits.

To operate at 2.4 V p-p transmit level, both the local and remote PHYs must advertise that they are capable of operating at 2.4 V, and at least one PHY must request 2.4 V p-p transmit level operation.

If it is required to only operate the PHY at 1.0 V p-p transmit level operation, AN_ADV_B10L_TX_LVL_HI_ABL must be 0 so that 2.4 V p-p transmit level operation is not advertised. In this case, autonegotiation can only resolve to 1.0 V p-p transmit level operation, irrespective of what setting the remote PHY advertises.

Master/Slave Resolution

Autonegotiation is also used to resolve master or slave status. The PHY can be configured to prefer slave or prefer master through the hardware configuration (see [Table 14\)](#page-17-0). If autonegotiation is disabled, the MS_SEL hardware configuration pin sets the default master/slave selection. Note that the recommended use of the ADIN1110 is with autonegotiation enabled.

During autonegotiation, when prefer slave is selected, and the remote end is prefer or forced master, the local PHY is set to slave (and remote to master). When the remote end is prefer or forced slave, the local PHY is set to master (and remote to slave).

MDI

The media dependent interface (MDI) connects the ADIN1110 to the Ethernet network via a twisted wire pair.

The ADIN1110 requires an external hybrid between the TXN/TXP and RXN/RXP pins and the twisted wire pair. This external hybrid allows the system to have full duplex communication by removing the local transmit signal from the combined signal on the cable, leaving the desired receive signal.

The ADIN1110 hybrid requires a specific topology and values for proper operation. Figure 10 shows the topology and values for the components. Consider the size, power, and voltage rating of these components in the context of other system requirements, for example, requirements of intrinsic safety.

Figure 10. Recommended Hybrid for the ADIN1110

RESET OPERATION

The ADIN1110 supports the following chip resets:

- ► Power-on reset
- ► Hardware reset
- ► Software reset
- ► MAC subsystem reset
- ► PHY subsystem reset

All of these resets put the ADIN1110, including the PHY core and MAC, into a known state. Whenever the MAC is reset, the SDO pin is pulled down and the TS_TIMER pin is driven to a low state.

Power-On Reset

The ADIN1110 includes power monitoring circuitry to monitor all of the supplies. During power-up, the ADIN1110 is held in hardware reset until each of the supplies has crossed its minimum rising threshold value and the power is considered good.

The POR module includes brownout protection by monitoring the supplies to detect if one or more of the supplies falls below a minimum falling threshold value. If brownout detection occurs, the device is held in hardware reset until the power is good.

Hardware Reset

A hardware reset is initiated by the power-on reset circuitry or by asserting the RESET pin low. Bring the RESET pin low for a minimum of 10 µs. Deglitch circuitry is included on this pin to reject pulses shorter than 0.3 μs.

When the RESET pin is deasserted, all the input/output (I/O) pins are held in tristate mode, the hardware configuration pins are latched, and the I/O pins are configured for their functional mode. When all the external and internal supplies are valid and stable, the crystal oscillator circuit is enabled. After the crystal starts up and stabilizes, the phase-locked loop PLL is enabled. After approximately 50 ms (maximum) from the deassertion of the RESET pin, all the internal clocks are valid, internal logic is released from reset, and all the management interface registers are accessible so that the device can be programmed.

Software Reset

A full chip software reset can be initiated by writing 1 to the SWRESET field of the RESET register.

If a transmission is taking place when the SPI software reset is initiated, the frame transmission stops abruptly and a runt or a frame

with a bad cyclic redundancy check (CRC) may be transmitted. Once the MAC-PHY is reset, the ADIN1110 is ready to bring up links.

When this software reset is initiated, a full initialization of the chip, almost equivalent to a hardware reset, is done. The I/O pins are held in tristate mode, the hardware configuration pins are latched, and the I/O pins are configured for their functional mode. The crystal oscillator circuit is enabled, and after the crystal starts up and stabilizes, the PLL is enabled. Approximately 10 ms (maximum) after writing the SOFT_RST keys, the internal logic is released from reset and all the management interface registers are accessible.

MAC Subsystem Reset

A MAC only software reset is initiated by writing the required pair of keys to the SOFT_RST register.

The reset is applied for about 1.2 us. The MAC subsystem reset interrupts any transmit/receive packet exchange between the MAC and the PHY, but does not drop any existing link nor prevents a link from establishing. The PHY management registers are not initialized.

The PHY needs to be out of software power-down to trigger a MAC only software reset.

PHY Subsystem Reset

The PHY subsystem is the part of the ADIN1110 that incorporates the 10BASE-T1L PHY transceiver analog and digital circuits. A PHY subsystem reset is initiated by setting the PHY subsystem reset register bit (CRSM_PHY_SUBSYS_RST). When this bit is set, the PHY subsystem is reset. The reset is applied for about 1.2 µs and then this bit self clears. All of the PHY digital circuitry is reset and any existing link drops. The management registers are not initialized by this reset, and access to all the management registers is available during the PHY subsystem reset. This is a short reset and can be used to put the device into a known state while retaining any software initialization of the device.

LED FUNCTIONS

LED 0 and LED 1 can be configured to display various activities of the ADIN1110 using the LED function feature. The LED function is configurable using the LED0_FUNCTION and LED1_FUNCTION bits (see the [LED Control Register](#page-94-0) section).

The 7, 8, 9, and 10 (decimal) bit settings for LEDx_FUNCTION are not available in LED Mode 2.

¹ The 7, 8, 9, and 10 (decimal) settings in the LEDx_FUNCTION bits are not available in LED Mode 2.

² See [Table 2](#page-4-0).

Typical Use

The LED 0 and LED 1 pins can be used to connect external LEDs to indicate the ADIN1110 link status and transmit or receive activity. The activity assigned to each LED is configurable through LED CNTRL (see the [LED Control Register](#page-94-0) section).

The LED pins are suitable for ultra low power LEDs. The maximum output current for the LED_0 and LED_1 pins is 8 mA with a VDDIO = 3.3 V. For higher LED power requirements, the use of an external transistor is recommended.

The LED x pins can also be connected to a host microcontroller general-purpose input/output (GPIO) (configured as a pulse-width modulated input or hardware interrupt). This configuration can be useful in applications where the user interface needs to be fully handled by an external host controller (for example, an external LED module or display). Note that in this context, it is recommended to place a low value resistance in series between the ADIN1110 LED x pins and the controller to avoid any potential transient surge current.

LED Pin Multiplexing

An internal multiplexer needs to be configured to enable the LED_1 signal on the LED 1 pin. LED 1 is disabled by default and can be enabled using the DIGIO_LED1_PINMUX bits (see the [Pin Mux](#page-92-0) [Configuration 1 Register](#page-92-0) section).

The LED 0 pin does not need multiplexing.

LED Polarity

The LED 0 and LED 1 pins can be configured to support various LED circuit polarities through the LED polarity mode feature (see the [LED Polarity Register](#page-96-0) section). Three polarity modes are available for each LED, as follows:

- ► Autosense (default)
- ► Active high
- ► Active low

In autosense mode, the ADIN1110 automatically senses the pin at power-up or reset to select the appropriate polarity configuration. In active high mode, the ADIN1110 is configured to drive the LED from the anode side. In active low mode, the ADIN1110 is configured to drive the LED from the cathode side.

Example circuits are described in the [LED Circuit Examples](#page-25-0) section.

LED Mode

LED 0 and LED 1 activity behavior can be configured using the two LED modes, as follows:

 \blacktriangleright LED Mode 1: blink duty cycle defined using the LED0_BLINK_TIME_CNTRL register (see the [LED_0 On/Off](#page-93-0)

[Blink Time Register](#page-93-0) section) and LED1_BLINK_TIME_CNTRL register (see the [LED 1 On/Off Blink Time Register](#page-93-0) section), respectively

► LED Mode 2: blink duty cycle automatically defined by the ADIN1110 based on activity level (%)

LINK STATUS PIN

The LINK_ST pin is asserted high when the link status bit (AN_LINK_STATUS) is asserted and indicates that the link between the ADIN1110 and its link partner is active.

By default, the LINK_ST signal is active high and can be configured to be either active high or active low using the DIG-IO_LINK_ST_POLARITY bit (see the [Pin Mux Configuration 1](#page-92-0) [Register](#page-92-0) section).

POWER-DOWN MODES

The ADIN1110 supports two power-down modes, as follows:

- ► Hardware power-down
- ► Software power-down

The lowest power mode is hardware power-down mode in which the device is turned off and the registers are not accessible.

Hardware Power-Down Mode

Hardware power-down mode can be used when no operation is required on the ADIN1110 and the power consumption needs to be minimized. The ADIN1110 enters hardware power-down mode when the RESET pin is asserted and held low. In this mode, all analog and digital circuits are disabled, the clocks are gated off, all the I/O pins are held in tristate mode, and the only power is the leakage power of the circuits. The management registers are not accessible in this mode.

Software Power-Down Mode

Software power-down mode can be used to configure the ADIN1110 registers before bringing a link up. The ADIN1110 can be configured to enter software power-down mode after reset using the SWPD EN pin. The ADIN1110 can also be instructed to enter software power-down mode by setting the software power-down bit (CRSM_SFT_PD).

The software power-down status bit (CRSM_SFT_PD_RDY) indicates that the device is in the software power-down state. In software powerdown mode, the analog and digital circuits are in a low power state, and the PLL is active and can provide output clocks if configured to do so. Any signal or energy on the MDI pins is ignored and no link is brought up. The management interface registers are accessible, and the device can be configured using software. The ADIN1110 exits software power-down mode when the CRSM_SFT_PD bit is cleared. At this point, the MAC-PHY starts autonegotiation and attempts to bring up a link after autonegotiation completes.

HARDWARE CONFIGURATION PINS

The ADIN1110 can operate in unmanaged or managed configurations with the use of the hardware configuration pins.

The hardware configuration pins are standard pins with an alternate bootstrap function. The ADIN1110 reads the configuration pin level immediately after power-up, hardware reset or software reset, and configures the PHY settings accordingly. When activated, the ADIN1110 immediately attempts to bring up a link on the PHY and the hardware configuration pins can be used with their main pin function. These pins can be used in unmanaged or managed configuration.

The unmanaged configuration refers to the ADIN1110 PHY parameters being configured by the hardware configuration pins. This mode can be used when the system requires a static configuration of the ADIN1110 port settings without the need for software control.

The managed configuration refers to the full control of the ADIN1110 using software via the SPI. The PHY and the MAC layer can be configured in software. The configuration pins can be connected to the external host or hardware configured using pull-up/pull-down resistors. When active the host controller can override any of the ADIN1110 configuration with the hardware pins after power-up, hardware reset or software reset.

UNMANAGED APPLICATIONS

In unmanaged applications, it is possible to configure the ADIN1110 using the hardware configuration pins without any software intervention.

The software power-down after reset must be disabled for unmanaged applications, or the ADIN1110 remains in power-down indefinitely because the device can only exit power-down from register operation using the SPI (see the [Software Power-Down Mode](#page-15-0) section).

MANAGED APPLICATIONS

In managed applications, the ADIN1110 can be configured by a host controller via the SPI. The host controller can dynamically configure the device as required by the application.

In managed applications, the software power-down after reset functionality can be enabled because the host controller brings the ADIN1110 to active mode using the management interface.

HARDWARE CONFIGURATION PIN FUNCTIONS

The following functions are configurable from the ADIN1110 hardware configuration pins:

- ► Software power-down mode after reset
- ► Transmit amplitude configuration
- ► Master/slave selection
- ► SPI protocol configuration

All of the hardware configuration pins have internal pull-down resistors. The default mode of operation without any external resistors connected to these pins is shown in Table 11. If an alternative mode of operation is required, use 4.7 kΩ pull-up resistors.

Table 11. Default Hardware Configuration Modes

¹ A low value series resistor is recommended.

² An external pull-down resistor is recommended.

Software Power-Down After Reset

If the ADIN1110 is configured so that it does not enter software power-down mode after reset, the ADIN1110 starts autonegotiation when it exits reset and attempts to bring up a link after autonegotiation completes. If the ADIN1110 is configured so that it enters software power-down mode after reset, the ADIN1110 waits in software power-down mode until it is configured over the SPI. At this point, the PHY configuration can be set to exit software power-down by software.

Table 13. Software Power-Down (Hardware Configuration)

Master/Slave Preference

The MS SEL hardware configuration pin is shared with the TS_TIMER pin and configures the default master/slave selection. If MS_SEL is pulled low during power-up or reset, the device is configured by default to prefer slave (this is the case if no external pull-up resistor is connected to the MS_SEL pin due to the presence of the internal pull-down resistor). If MS_SEL is pulled high during power-up or reset, the device is configured by default to prefer master.

If autonegotiation is disabled, MS SEL sets the default master/slave selection. Autonegotiation is enabled by default for the

HARDWARE CONFIGURATION PINS

ADIN1110 and it is strongly recommended that autonegotiation is always enabled.

During autonegotiation, when prefer slave is selected and the remote end is prefer or forced master, the local PHY is set to slave (and remote to master). When the remote end is prefer or forced slave, the local PHY is set to master (and remote to slave).

Transmit Amplitude

The TX2P4 EN hardware configuration pin allows the user to configure the required transmit amplitude mode for the intended application (see Table 15). If TX2P4 EN is pulled low, the ADIN1110 is configured by default to support both 1.0 V p-p and 2.4 V p-p transmit levels, decided by autonegotiation. If TX2P4_EN is pulled high, the ADIN1110 is configured to disable 2.4 V p-p transmit operating mode by default and operate with 1.0 V p-p transmit level only. If the $\overline{TX2P4}$ EN pin is strapped high (1.0 V p-p only), the associated register cannot be changed through the SPI. For example, 2.4 V p-p operation is not possible if the ADIN1110 is hardware pin configured for 1.0 V p-p only.

The 1.0 V p-p transmit operating mode supports the spur use case and can operate at a lower AVDD H supply voltage of 1.8 V. This supports intrinsic safe applications.

The higher transmit operating mode of 2.4 V p-p supports trunk applications and requires a higher AVDD_H supply voltage of 3.3 V. This mode can be used for longer cable lengths in industrial Ethernet environments with high noise levels.

SPI Protocol Configuration

The ADIN1110 allows the use of a generic SPI protocol with or without the use of CRC, or the OPEN Alliance SPI protocol with or without protection.

Table 16. SPI Protocol (Hardware Configuration)

BRINGING UP 10BASE-T1L LINKS

UNMANAGED PHY OPERATION

For an unmanaged PHY application or lightly managed PHY application where there is no software management of the PHY, the hardware configuration pins determine the operating mode. The TX2P4 EN pin configures the PHY to advertise the support of both 1.0 V p-p and 2.4 V p-p transmit level operation or to only advertise support of 1.0 V p-p transmit level operation. The MS_SEL pin is used to configure the PHY to advertise prefer slave or prefer master. The SWPD EN pin must be pulled up at power-up and reset so that the PHY does not enter software power-down mode when it exits reset. Once the PHY exits reset, the ADIN1110 starts autonegotiation and attempts to bring up a link after autonegotiation completes.

A lightly managed PHY can use the hardware configuration pins to determine the operation of the PHY and to bring up a 10BASE-T1L link. Afterwards, software can monitor the operation of the PHY.

MANAGED PHY OPERATION

In a managed PHY application, software is used to configure the PHY operation using the management interface. The hardware configuration pins can be used to set the default values of the registers used to control the transmit amplitude and master/slave setting. The SWPD EN pin must be pulled low at power-up and reset so that the PHY enters software power-down mode when it exits reset. The PHY remains in software power-down mode until the software configures the PHY and takes it out of software power-down mode to start autonegotiation and attempt to bring up a link.

Power-Up and Reset Complete

To confirm that the MAC has exited reset, read the PHY identification register (PHYID). If the reset value of the register (0x283BC91) can be read, the device has exited reset and is ready for configuration.

Next, the host must read the STATUS0 register and confirm that the RESETC field is 1. Note that if the RESETC field is 0, and the SYNC field of the CONFIG0 register is 1, the MAC-PHY is already configured by the host and, therefore, has not been reset. This can indicate that the host is reset, but the MAC-PHY has not been reset.

Write 1 to the RESETC field in the STATUS0 register to clear this field, and the interrupt pin asserts high.

The PHYINT field of the STATUS0 register also asserts. To clear this field, the corresponding status registers, PHY_SUB-SYS_IRQ_STATUS and CRSM_IRQ_STATUS, must be cleared or masked.

The system ready bit (CRSM_SYS_RDY) can also be read to verify that the start-up sequence is complete and the system is ready for normal operation.

The software power-down status bit (CRSM_SFT_PD_RDY) can be read to check if the device is in the software power-down state. This bit is configured by the SWPD EN hardware configuration pin.

MAC Initialization

After power-up or reset, configure the ADIN1110 MAC. Write the IMASK0 and IMASK1 registers to enable interrupts as required.

Write CONFIG0 and CONFIG2 to set up the required functionality of the MAC. For example, set the OPEN Alliance chunk size or enable cut through, if required.

When the MAC is configured, write 1 to the SYNC field in the CON-FIG0 register to indicate that the MAC configuration is complete.

Configuring the Device for Linking

After power-up or reset, configure the ADIN1110 PHY for the desired operation for linking. The ADIN1110 may already be configured as required for linking by the hardware configuration pins, but greater control is available using the management registers.

The autonegotiation process is used to match the operating mode between a local and remote PHY. For example, autonegotiation is used to ensure that the modes agree between the two devices on which PHY operates as master and which as slave. Autonegotiation is also used to match the transmit level between the two PHYs.

Autonegotiation is enabled by default for the ADIN1110, and it is strongly recommended to always keep autonegotiation enabled. Autonegotiation is defined by the IEEE standard and includes a number of mechanisms to ensure robust linking operation between PHYs and is the fastest way to bring up a link.

Advertisement of Transmit Level Operating Mode

The ADIN1110 can support transmit level operation at either 1.0 V p-p or 2.4 V p-p if the 10BASE-T1L high voltage transmit ability read only register bit (B10L TX LVL HI ABLE) is 1 and there is a 3.3 V supply provided on the AVDD_H pins. The higher transmit level can support longer reach but also has higher power consumption. The ADIN1110 can support 1.0 V p-p transmit level operation with a 1.8 V supply on the AVDD H pins at very low power consumption. The 1.0 V p-p transmit level operation is required for intrinsically safe operation.

The ADIN1110 can either be configured to advertise support of both 1.0 V p-p and 2.4 V p-p transmit level operation (if B10L_TX_LVL_HI_ABLE = 1) or to advertise support of only 1.0 V p-p transmit level operation. This is set using the 10BASE-T1L high level transmit operating mode ability bit within the BASE-T1 autonegotiation advertisement register (AN_ADV_B10L_TX_LVL_HI_ABL). 0 = support 1.0 V p-p transmit level only, and $1 =$ support both $1.0 \vee p$ -p and $2.4 \vee p$ -p transmit level.

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The ADIN1110 can also be configured to advertise a request for 2.4 V p-p transmit level operation (if B10L_TX_LVL_HI_ABLE = 1). This is set using the 10BASE-T1L high level transmit operating mode request bit (AN_ADV_B10L_TX_LVL_HI_REQ). 0 = request 1.0 V p-p transmit level, and $1 =$ request 2.4 V p-p transmit level.

The link partner advertised transmit level ability can be read in the link partner 10BASE-T1L high level transmit operating mode ability register bit (AN_LP_ADV_B10L_TX_LVL_HI_ABL). The link partner advertised transmit level request can be read in the link partner 10BASE-T1L high level transmit operating mode request register bit (AN_LP_ADV_B10L_TX_LVL_HI_REQ). These bits are updated during the autonegotiation process and are valid when the autonegotiation complete register bit (AN_COMPLETE) is set.

Operation at the 1.0 V p-p transmit level operation occurs if either the local or remote PHY advertises that it is not capable of transmitting in the high level (2.4 V p-p) transmit operating mode, or if neither the local nor remote PHY advertises a request for high level (2.4 V p-p) transmit operating mode.

Operation at the 2.4 V p-p transmit level occurs if both the local and remote PHY advertises that they are capable of transmitting in the high level (2.4 V p-p) transmit operating mode, and if either the local or remote PHY advertises a request for high level (2.4 V p-p) transmit operating mode.

Therefore, a PHY can ensure it must operate at 1.0 V p-p transmit level, but it can only request operation at the 2.4 V p-p transmit level.

Table 17. Determination of Transmit Level by Autonegotiation¹

¹ X means don't care.

² HI_ABL, HI_REQ, LP_HI_ABL, and LP_HI_REQ refer to the advertisement bits AN_LP_ADV_B10L_TX_LVL_HI_ABL, AN_LP_ADV_B10L_TX_LVL_HI_REQ, AN_ADV_B10L_TX_LVL_HI_ABL, and AN_ADV_B10L_TX_LVL_HI_REQ, respectively.

Advertisement of Master/Slave

The 10BASE-T1L standard uses what is known as a master/slave clock scheme. This scheme is commonly used in full duplex transceiver standards using echo cancellation. One PHY is designated as the master and the other PHY as the slave. Autonegotiation is used to agree which PHY is the master and which is the slave, and it generally doesn't matter which PHY is which.

The ADIN1110 has an internal pull-down resistor on the MS_SEL pin, which results in a default setting of configuring the PHY to advertise prefer slave. It is recommended to either use the default setting of advertise prefer slave or to use a setting of advertise prefer master.

If it is mandatory for the PHY to operate as master, use an advertise forced master configuration. However, this configuration must be used with caution because if remote end is also forced master, there is a configuration fault, autonegotiation fails, and the link is not brought up.

The force master/slave configuration register bit

(AN_ADV_FORCE_MS) is used to configure the PHY to advertise its master/slave configuration as a preference or as a forced value, as follows: 0 = master/slave configuration is a preferred mode, and 1 = master/slave configuration is a forced mode.

The master/slave configuration register bit (AN_ADV_MST) is used to configure the PHY to advertise its master/slave configuration, as follows: $0 =$ slave and $1 =$ master.

The link partner advertised master/slave setting can be read in the link partner force master/slave configuration register bit (AN_LP_ADV_FORCE_MS) and the link partner master/slave configuration register bit (AN_LP_ADV_MST). These bits are updated during the autonegotiation process and are valid when the autonegotiation complete register bit (AN_COMPLETE) is set.

When the local and remote PHY have the same preferred configuration (for example, both slave or both master), a random process is used to determine which is the master and which is the slave. When one PHY has a forced configuration, its master/slave configuration is given priority over a PHY with a preferred setting where both PHYs have the same master/slave configuration. If both PHYs have a forced configuration and the same master/slave configuration, configuration fault occurs and autonegotiation fails.

The resolution of master/slave can be read using the master/slave resolution result register bits (AN_MS_CONFIG_RSLTN). This result indicates if the PHY is configured as a slave or a master or if there was a configuration fault. These bits are updated during the autonegotiation process and are valid when the autonegotiation complete register bit (AN_COMPLETE) is set.

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Table 18. Determination of Master/Slave by Autonegotiation¹

¹ X means don't care.

Completion of Autonegotiation

When autonegotiation completes, the autonegotiation complete indication register bit (AN_LINK_GOOD) is set. This bit indicates completion of the autonegotiation transmission, and that the enabled PHY technology is either bringing up its link or that it has brought up its link.

When autonegotiation has completed and the link is up, the autonegotiation complete register bit (AN_COMPLETE) is set. When this bit is read as 1, the autonegotiation process is complete, the PHY link is up, and the contents of the AN_ADV_ABILITY and AN_LP_ADV_ABILITY register bits are valid.

Link Status

The status of the link can be determined by reading the link status register bit (AN_LINK_STATUS). This bit latches low.

When read as 1, this bit indicates that a valid link is established.

If this bit reads 0, the link has failed since the last time it was read. If the value of this bit is read as 0, this bit must be read a second time to determine the link status (see [Latch Low Registers](#page-41-0) section).

If the link is dropped, the autonegotiation process restarts automatically. Autonegotiation can be restarted by request through a write to the autonegotiation restart bit (AN_RESTART) in the autonegotiation control register (AN_CONTROL).

LOOPBACK MODES

The PHY core within the MAC-PHY provides the following loopback modes:

- ► Physical medium attachment (PMA) loopback
- ► Physical coding sublayer (PCS) loopback
- ► MAC interface loopback
- ► MAC interface remote loopback

These loopback modes test and verify various functional blocks within the PHY. The use of the built-in frame checker and frame generator allows completely self contained in-circuit testing of the digital and analog data paths within the PHY core. A loopback can also be established that includes the MAC portion of the ADIN1110 (not limited to just the PHY core) by implementing the necessary software within the host processor.

PMA Loopback

For PMA loopback, leave the MDI pins open circuit, thereby transmitting into an unterminated connector or cable. For the most accurate results, leave the cable disconnected. The PHY can then operate by receiving the reflection from its own transmission. This loopback is intended as an implementation of IEEE Standard 802.3cg Subclause 146.5.6 PMA local loopback.

Note that for 10BASE-T1L PMA local loopback, the device must be configured in the forced link configuration mode (autonegotiation disabled). Setting the B10L_LB_PMA_LOC_EN bit (B10L_PMA_CNTRL register) enables PMA loopback.

PCS Loopback

PCS loopback mode loops the transmit data back to the receiver within the PCS block at the input stage of the PHY digital block. Setting the B10L_LB_PCS_EN bit (B10L_PCS_CNTRL register) enables PCS loopback.

When the PCS loopback mode is enabled, no signal is transmitted to the MDI pins.

MAC Interface Loopback

MAC interface loopback mode loops the data received on the MAC interface back to the SPI host. Setting the MAC_IF_LB_EN bit (MAC_IF_LOOPBACK register) enables MAC interface loopback. If the MAC_IF_LB_TX_SUP_EN bit within the same register is set (set by default), the transmission of the signal is suppressed to the MDI pins.

MAC Interface Remote Loopback

MAC interface remote loopback requires a link up with a remote PHY and enables looping of the data received from the remote PHY back to the remote PHY. This linking allows a remote PHY to verify a complete link by ensuring that the PHY receives the proper data. Setting the MAC_IF_REM_LB_EN bit (MAC_IF_LOOP-BACK register) enables MAC interface remote loopback. If the MAC_IF_REM_LB_RX_SUP_EN bit within the same register is set (set by default), the data received by the PHY is suppressed and not sent to the MAC.

MAC Loopback

The MAC loopback loops the data received on the MAC transmit channel back to the SPI host. The MAC loopback is enabled with the P1_LOOP register.

Host Processor Loopback

Outside of the loopback modes associated with the PHY core within the ADIN1110, the host processor can be used to create a full MAC loopback. In a full MAC loopback, whatever frame is received from the MAC is transmitted back to the MAC, as shown in Figure 11.

Figure 11. ADIN1110 Loopback Modes

FRAME GENERATOR AND CHECKER

The ADIN1110 can be configured to generate frames and to check received frames (see Figure 12). The frame generator and checker can be used independently to generate frames or check frames, or the frame generator and checker can be used together to simultaneously generate frames and check frames. If frames are looped back at the remote end, the frame checker can be used to check frames generated by the ADIN1110.

When the frame generator is enabled, the source of the data for the PHY comes from the frame generator and not the MAC.

The frame generator control registers configure the type of frames to be sent (for example, random data, all 1s), the frame length, and the number of frames to be generated.

The generation of the requested frames starts by enabling the frame generator (FG_EN). When the generation of the frames completes, the frame generator done bit is set (FG_DONE).

The frame checker is enabled using the frame checker enable bit (FC_EN). The frame checker can be configured to check and analyze received frames from either the MAC interface or the PHY, which is configured using the frame checker transmit select bit (FC_TX_SEL). The frame checker reports the number of frames received, CRC errors, and various other frame errors. The frame checker frame counter register and frame checker error counter register count these events.

The frame checker counts the number of CRC errors and these are reported in the receive error counter register (RX_ERR_CNT). To ensure synchronization between the frame checker error counter and frame checker frame counters, all of the counters are latched when the receive error counter register is read. Therefore, when using the frame checker, read the receive error counter first, and

then read all other frame counters and error counters. A latched copy of the receive frame counter register is available in the FC_FRM_CNT_H register and FC_FRM_CNT_L register.

In addition to CRC errors, the frame checker counts frame length errors, frame alignment errors, symbol errors, oversized frames errors, and undersized frame errors. In addition to the received frames, the frame checker counts frames with an odd number of nibbles in the frame, and counts packets with an odd number of nibbles in the preamble. The frame checker also counts the number of false carrier events, which is a count of the number of times the bad start of the stream delimiter (bad SSD) state is entered.

Frame Generator and Checker Used with Remote Loopback with Two MAC-PHYs

Using two MAC-PHY devices, the user can configure a convenient self contained validation of the PHY core to PHY core connection, or can exercise the full signal chain by using the host processor to perform the loopback at the remote end. Figure 12 shows an overview of how each MAC-PHY is configured. An external cable is connected between both devices, and MAC-PHY 1 is generating frames using the frame generator.

When limiting the test to just the PHY core portions of the ADIN1110, MAC-PHY 2 has MAC interface remote loopback enabled (MAC_IF_REM_LB_EN). The frames issued by MAC-PHY 1 are sent through the cable, through the PHY 2 signal chain returned by the PHY 2 MAC interface remote loopback, back again through the cable, and checked by the MAC-PHY 1 frame checker. Alternatively, the frames from MAC-PHY 1 can be sent all the way to the host processor of the remote device and looped back from there, through the MAC and PHY blocks within MAC-PHY 2, and back to MAC-PHY 1.

Figure 12. Remote Loopback Used Across Two PHYs for Self Check Purposes

TEST MODES

The ADIN1110 provides several test modes that allow testing of the transmitter waveform, distortion, jitter, and droop. These test modes change only the data symbols provided to the transmitter circuitry and do not alter the electrical and jitter characteristics of the transmitter and receiver from the normal operation.

The following test modes included in theADIN1110 are defined in Subclause 146.5.2 from the IEEE 802.3cg:

- ► Test Mode 1. This is a transmitter output voltage and timing jitter test mode. When this mode is selected, the ADIN1110 repeatedly transmits the data symbol sequence (+1, –1).
- ► Test Mode 2. This is a transmitter output droop test mode. In this mode, the ADIN1110 transmits ten +1 symbols followed by ten −1 symbols. This sequence repeats indefinitely.
- ► Test Mode 3. Normal operation in idle mode test mode. When this test mode is selected, the ADIN1110 transmits as in non test operation and in the master data mode with data set to normal interframe idle signals.

In addition to these test modes, the ADIN1110 provides the transmit disable mode defined in Subclause 45.2.1.186a.2. This mode maintains both the receive and transmit path active as in normal operation, but only transmits 0 symbols. This mode can be used to measure the MDI return loss specified in Subclause 146.8.3.

Accessing Test Mode 1 to Test Mode 3

To set the ADIN1110 into Test Mode 1, Test Mode 2, or Test Mode 3, the device must be in software power-down mode (CRSM_SFT_PD). The power-down status of the ADIN1110 can be checked reading the software power-down status bit (CRSM_SFT_PD_RDY).

When the ADIN1110 is in software power-down mode, disable autonegotiation by clearing the autonegotiation enable bit (AN_EN).

With autonegotiation disabled, force the autonegotiation configuration by writing to the autonegotiation forced mode bit (AN_FRC_MODE_EN).

The desired test mode can now be selected by writing the appropriate value to the 10BASE-T1L test mode control register (B10L_TEST_MODE_CNTRL) and exiting the device from powerdown by clearing the software power-down bit (CRSM_SFT_PD).

Accessing Transmit Disable Test Mode

To configure the ADIN1110 in the transmit disable test mode, the device must be in software power-down mode (CRSM_SFT_PD). The power-down status of the ADIN1110 can be checked reading the software power-down status bit (CRSM_SFT_PD_RDY).

When the ADIN1110 is in software power-down mode, disable autonegotiation by clearing the autonegotiation enable bit (AN_EN).

With autonegotiation disabled, force the autonegotiation configuration by writing to the autonegotiation forced mode bit (AN_FRC_MODE_EN).

The transmit disable test mode can now be enabled by setting the B10L_TX_DIS_MODE_EN bits to 1. To exit from the test mode, write 0 to CRSM_SFT_PD.

TIME DOMAIN REFLECTOMETRY (TDR)

Given that the 10BASE-T1L compliant PHY enables communication over long cables, debugging a faulty cable can become costly and difficult without the right tools. To help with this, Analog Devices 10BASE-T1L products provide a TDR engine that enables cable fault detection, distance to fault, and cable length estimation.

The diagnostics solution is the combination of a highly accurate on-chip TDR engine and a set of algorithms that run on a host microcontroller, allowing maximum flexibility for a wide variety of cables and more advanced cable diagnostic capabilities.

Figure 13. ADIN1110 TDR Engine

Fault Detection with the TDR Engine

The Analog Devices algorithm has a time resolution of 8.3 ns, which translates to a length resolution of less than 1 m and a maximum of 1600 m, with an accuracy of 2%.

This fault detector algorithm is capable of finding open and short fault conditions even when the ADIN1110 is physically connected to another PHY through their MDI, which implies that the link partner PHY is potentially transmitting DME pages. Traditional TDR methods struggle to find faults if other signal sources or noise is also present in the same link. This is not the case of the Analog Devices solution, which makes it suitable for debugging when there is no control over the remote end.

The fault detector algorithm is provided as a C-code library containing the high-level functions required for diagnostics. These functions have been optimized to not utilize any advanced processing so that they can be executed by any low-power microcontroller.

A single function call is sufficient to execute the fault detector. The function returns the type of fault and the distance to the fault in meters from the MDI connector.

TDR Offset Calibration

The library includes a function to calibrate the offset of the TDR measurement. This particular function in the library is useful given that different MDI circuits may introduce variable delays in the signal path, which can contribute to the offset of the length measurement. For instance, an isolation transformer on the MDI is highly

likely to introduce a signal delay that corresponds to a couple of meters in length.

This calibration is not required to run the fault detector, and an average value is provided by default. However, it is recommended for short cables if accuracy is required. If this calibration is required, it can be done once in the lab for a specific MDI circuit implementation, and the offset value can then be stored in nonvolatile memory for future use.

To perform this calibration, the MDI port must be left open or shorted. No load or cable can be connected to the MDI port.

Cable Calibration

By default, the algorithm is optimized to support long reach cables compliant with the IEEE 802.3cg standard. However, given the wide variety of cable types, which have different insertion loss, return loss, and signal delay characteristics, the library includes a calibration function that optimizes the algorithm to operate with any cable, and estimates its nominal velocity of propagation (NVP) for more accurate length estimations. The length accuracy mainly depends on the accuracy of the NVP value.

To run this calibration, a cable with a known length must be attached to the MDI port, and its end must be left open or shorted. NVP values are generally between 0.5 and 0.9 and are a property of the construction of the cable. In general, an average NVP value of approximately 0.65 can be assumed. This calibration is not required to run the fault detector, unless higher length accuracy is needed or if nonstandard cables are utilized. This calibration can be done once in the laboratory for a given cable, and the values can be stored in nonvolatile memory.

Refer to the C-code driver for more information related to the usage of these functions.

Length/Distance to Fault Accuracy

The accuracy of the distance to a fault, or length measurements, mainly depends on the NVP value, which is determined by the accuracy of the cable length used to perform the NVP calibration.

Table 19 provides results for induced faults and distance-to-fault measurements for different cables and lengths. In all cases, the algorithm was successful finding the open or short conditions induced during the test. The NVP value for the Profibus PA cable used in this test was roughly estimated, and the same was used for the Cat5E and Cat6 cables.

Table 19. Length Estimation Error for Different Cables

SYSTEM LEVEL POWER MANAGEMENT

Transmit Level = 1.0 V p-p

The transmit mode of 1.0 V p-p can be used for shorter reach applications supporting cable lengths up to 400 m where signal amplitude requirements tend to be lower.

For applications where the ADIN1110 must operate in a 1.0 V p-p transmit operating mode, the TX2P4_EN pin must be tied high via a 4.7 kΩ resistor (see Figure 14). This configuration forces the ADIN1110 to only operate at 1.0 V p-p transmit operating mode and enables the operation of the ADIN1110 from a signal supply voltage, operating at a lower voltage rail (for example, 1.8 V), allowing the user to minimize power dissipation in the system.

Figure 14. Supplies and Capacitors for 1.0 V p-p Transmit Mode

Transmit Level = 2.4 V p-p

For longer reach applications, a higher signal amplitude of 2.4 V p-p is required. The ADIN1110 is designed to operate with long reach cables up to 1000 m in this mode, requiring a higher AVDD H supply voltage of 3.3 V.

For the ADIN1110 to be able to operate in 2.4 V p-p, the $\overline{TX2P4-EN}$ pin must be tied low (no external connection required to achieve this due to the presence of an internal pull-down resistor). This

mode of operation still allows the 1.0 V p-p operating mode to be selected via MDIO or via autonegotiation.

Figure 15 shows an overview of the proposed power configuration. For single supply operation, the same rail can be used to supply the AVDD H, AVDD L, and VDDIO supply rails. The DVDD 1P1 1.1 V rail can be derived internally or alternatively provided by an external 1.1 V rail. Note that this configuration requires that AVDD_H is 3.3 V even if the link is established at 1.0 V p-p transmit operating mode via MDIO or autonegotiation.

Figure 15. Supplies and Capacitors for 2.4 V p-p and 1.0 V p-p Transmit Mode

LED CIRCUIT EXAMPLES

The LED_0 and LED_1 pins can be used in various circuit configu-rations depending on the [LED](#page-96-0) polarity mode selected (see LED [Polarity Register\)](#page-96-0). The example circuits described in this section provide examples for the three polarity modes available for each LED, as follows:

- ► Autosense (default)
- ► Active low
- ► Active high

As described in the [LED Functions](#page-13-0) section, the maximum output current for both LED_0 and LED_1 is 8 mA with a VDDIO =

3.3 V. For higher current requirements, consider using the circuit described in Transistor Controlled LED.

Active High LED Polarity

In the active high configuration, the LED \times pin can drive an external LED from the anode side. Select the R0 and R1 resistors to control the LED current (refer to the selected LED specifications in [Table](#page-4-0) [2](#page-4-0) for information). External pull-down resistors (R_{PDO} , R_{PD1}) with a value of 4.7 kΩ are recommended.

Figure 16. Recommended Active High LED Circuit

Active Low LED Polarity

In active low configuration, the LED x pin can drive an external LED from the cathode side. Select the R0 and R1 resistors to control the LED current (refer to the selected LED specifications in [Table 2](#page-4-0) for information). External pull-up resistors $(R_{P_{U0}}, R_{P_{U1}})$ with a value of 4.7 kΩ are recommended.

Figure 17. Recommended Active Low LED Circuit

Transistor Controlled LED

Figure 18 displays a typical configuration where the LED current required is higher than what the LED_0 and LED_1 pins can supply.

The circuit operates using the active high LED mode. An external transistor such as an N-channel metal-oxide semiconductor field effect transistor (MOSFET) can be used. The transistor must be selected so the gate input capacitance is not sinking current above the maximum rating of the LED during the actuation (refer to the transistor technical specifications for information). If required, the inrush current can be reduced by placing a resistance between the transistor gate and the ADIN1110 pin, and/or adding a parallel capacitor between the GND and the LED_x pin. The additional resistor and capacitor values must be defined based on the transistor selection.

Select the R0 and R1 resistors to control the LED current. External pull-down resistors (R_{PD0}, R_{PD1}) with a value of 4.7 kΩ are recommended.

VCC can be set to match the LED power requirements.

Figure 18. Recommended Transistor Controlled LED Circuit

Autosense Polarity

In autosense mode, the polarity of the LED is automatically detected during power-up, hardware reset, or software reset.

LED 0 (internal pull-up) and LED 1 (internal pull-down) have different autosense behaviors due to their internal pull-up and pull-down configurations. Use one of the configurations described in the Active High LED Polarity, Active Low LED Polarity, and Transistor Controlled LED sections so that the two LED_x pins can be controlled the same way.

COMPONENT RECOMMENDATIONS

The ADIN1110 requires an external 25 MHz clock, which can be sourced from an external crystal oscillator or an external singleended clock.

The signal voltage on the XTAL_I/CLK_IN pin $(V_{CLK|N})$ must be a sine or filtered square wave signal with a peak-to-peak voltage range from 0.8 V to 2.5 V. For the single-ended clock option, a V_{C-K} IN with a 1.0 V p-p swing is recommended to achieve best performance.

Various circuit configurations are proposed in the following sections. A common circuit topology can be used across these options with a change to the passive component values.

Note that during normal operation, a 25 MHz reference clock generated from the external clock source input (a crystal or 25 MHz external single-ended clock) is provided on the CLK25_REF output pin. This pin can be used as a reference clock for other circuits, such as another 10BASE-T1L device. CLK25 REF is disabled in reset mode.

External Crystal Oscillator

The typical connection for an external crystal (XTAL) is shown in Figure 19.

To ensure minimum current consumption and minimize stray capacitance, make connections between the crystal, capacitors, and ground as close to the ADIN1110 as possible. Consult individual crystal vendors for recommended load information and crystal performance specifications.

The crystal load capacitance (C_L) is defined by the crystal vendor. C_{PCB1} and C_{PCB2} are the parasitic capacitance between the XTAL_I/CLK_IN and XTAL_O pins and the ground plan beneath, respectively. C_{X1} and C_{X2} are the two external load capacitors required for the oscillator to operate.

Assuming the following:

 \triangleright C_{PCB1} \approx C_{PCB2} \approx C_{PCBx}

$$
\blacktriangleright C_{X1} \approx C_{X2} \approx C_{Xx}
$$

Then, $C_{Xx} = 2 \times C_1 - C_{PCBx} - 3$ pF

Choose precision capacitors for $C_{\text{X}x}$ with low appreciable temperature coefficient to minimize frequency errors.

To ensure minimum current consumption and to minimize stray capacitance, make the connections between the crystal, capacitors, and ground as close to the ADIN1110 as possible.

Figure 19. Crystal Oscillator Connection

External 25 MHz Clock Input

The clock source must be a dc-coupled with the ADIN1110 XTAL_I/ CLK_IN pin input, and the XTAL_O pin must be left open circuit.

With 0.8 V \leq V_{CLK} IN p-p \leq 2.5 V, the following results:

- ► For 0.8 V ≤ V_S p-p ≤ 1.0 V, the following is true:
	- \blacktriangleright R1 = 50 Ω
	- ► R2 is not required
- \triangleright For 1.0 V < V_S p-p < 1.8 V, the following is true:
	- ► For best performance, set $V_{CLK IN}$ to 1.0 V p-p
	- ► 500 Ω ≤ R1 ≤ 2 kΩ
	- ► 1 kΩ ≤ R2 ≤ 2 kΩ

$$
\triangleright \ \mathsf{V}_{\mathsf{S}} \ \mathsf{p}\text{-}\mathsf{p} - \mathsf{V}_{\mathsf{CLK_IN}} \ \mathsf{p}\text{-}\mathsf{p} \geq 0.2 \ \mathsf{V}
$$

$$
R2 = \frac{V_{CLK_IN}p - p \times R1}{V_{Sp} - p - V_{CLK_IN}p - p}
$$

► For 1.8 V ≤ V_S p-p, the following is true:

$$
\blacktriangleright R1 = 2 k\Omega
$$

 \triangleright R2 = 2 kΩ

Figure 20. External 25 MHz Clock Input Circuit

802.1AS SUPPORT

Typically, any device operating in an 802.1AS network executes the following operations periodically:

- ► Generate peer delay request and handle response
- ► Receive peer delay request and generate response
- ► Receive synchronization frame (slave clock)
- ► Transmit synchronization frame (master clock)

These features require that the MAC is capable of time stamping specific incoming and outgoing frames.

To assist with these features, the ADIN1110 MAC provides the following hardware:

- ► Internal free-running counter
- ► Syntonized counter
- ► Waveform generation on TS_TIMER output

Internal Free Running Counter

The ADIN1110 has an internal free running counter running at 120 MHz. This counter provides an accuracy of 8.333 ns. When using this counter, there is a period of approximately 35 s. This period ensures that the clock does not wrap during any of the necessary operations, for example, between receipt of peer delay request and transmission of the response.

To enable the free running counter, the TS_EN bits must be set to 1.

When the free running counter is enabled, the ADIN1110 captures the time stamp for all received frames, and it is appended before each data frame received. The time stamps of transmitted frames are captured when requested. See [Time Stamp Capture](#page-30-0) for more details.

The value of the free running counter can be captured using the input capture signal (TS_CAPT), capturing the value of the counter in the TS_FREECNT_CAPT register.

Syntonized Counter

The syntonized counter is a 64-bit counter in which the lower 32 bits represent nanoseconds with 1 LSB = 1 ns. When the lower 32 bits reach the value stored in TS_1SEC_CMP, these bits clear and the upper 32 bits increment representing seconds.

To enable the syntonized counter, the TS_EN bits must be set to 1.

Three modes are supported for capturing time stamps for transmitted and received frames, as follows:

- **1.** Capture a 2-bit sec and a 30-bit ns time stamp as defined in the OPEN Alliance Specification Section 7.8. See the [Configuration](#page-47-0) [Register 0](#page-47-0) section.
- **2.** Capture a 32-bit sec and a 30-bit ns time stamp as defined in the OPEN Alliance Specification Section 7.8. See the [Configu](#page-47-0)[ration Register 0](#page-47-0) section.
- **3.** Capture the 32-bit free running counter. See the [Timer Configu](#page-60-0)[ration Register](#page-60-0) section.

To enable capturing of time stamps for transmitted and received frames, set FTSE (CONFIG0 register) to 1.

Waveform Generation on TS_TIMER Output

The ADIN1110 can generate an output signal (TS_TIMER) that uses two counters to generate repeating waveforms driven by the syntonized time. These two counters, TS_TIMER_HI and TS_TIM-ER_LO, specify the high and low period of the TS_TIMER signal and need to be programmed with multiples of 16 because they are driven by the syntonized time.

It is possible to specify a time with respect to the 64-bit syntonized timer to start the generation of the TS_TIMER output. The TS TIMER START register can be programmed with a value that is compared with the nanoseconds portion of the syntonized counter to generate a one-shot start.

The sequence to enable the TS_TIMER output is as follows:

- **1.** If required, change the default value of the TS_TIMER output from 0 to 1 by writing to the TS_TIMER_DEF bits.
- **2.** Write the values required for the high and low times for the TS_TIMER output to the TS_TIMER_HI and TS_TIMER_LO registers.
- **3.** Write the value required for the quantization error correction to the TS_TIMER_QE_CORR register.
- **4.** Write a start time to the TS_TIMER_START registers. When the nanoseconds part of the syntonized counter matches this value, TS TIMER starts toggling.

The TS TIMER output can be stopped by writing 1 to the TS TIM-ER_STOP bits. When the TS_TIMER output is stopped, the output goes back to the default value specified in TS_TIMER_DEF.

ELECTROMAGNETIC COMPATIBILITY (EMC) AND ELECTROMAGNETIC IMMUNITY (EMI)

The ADIN1110 was tested at the system level for EMC and EMI. Table 21 summarizes the results.

Table 21. EMC/EMI Tests Conducted on ADIN1110 at System Level

SPI

The ADIN1110 register interface is via a 4-wire SPI consisting of the following pins: SCLK, CS, SDI, and SDO/SPI_CFG0.

The possible access permissions of the registers are as follows:

- ► R/W: read/write
- ► R: read only
- ► W: write only
- ► R/W1C: read/write 1 to clear

The ADIN1110 also allows access to the PHY registers via an SPI to MDIO master bridge. See the [SPI Access to the PHY Registers](#page-39-0) section.

The following registers have additional access permissions:

- ► R LL: read only, latch low
- ► R LH: read only, latch high
- ► R/W SC: read/write, self clear

Generic SPI Protocol

The generic SPI protocol is detailed in Table 22 to [Table 29.](#page-30-0) The protocol is determined by the hardware configuration pins. The register map is organized as a 32-bit map, and all accesses are in multiples of 32-bit words. Both single and burst access in multiples

Table 22. Control Write Transaction

of 32-bit words are supported. The MSB of the data is transmitted first.

The R/W and TA fields are defined as follows:

- ► R/W: read/write
	- ► 0: read
	- ► 1: write
- ► TA: turn around

Burst writes and reads must be in multiples of 4 bytes. The last word (4 bytes) written can contain between 1 byte and 4 bytes of valid data. However, TX_FSIZE is still written with the original frame size $+2$ bytes for the frame header (see [Figure 21](#page-31-0)). For example, to transmit a 65-byte frame that is prepended with a 2-byte header, 67 is written to TX_FSIZE, but 68 bytes are transferred over SDI. The last byte is not used.

It is possible to enable a CRC on the SPI protocol via a hardware configuration pin on power-up. This 8-bit CRC uses the polynomial $x^8 + x^2 + x + 1$ seeded with 0x0, and provides up to 3-bit error detection. The 8-bit CRC is included for every control and data transaction after the ADDR bits, and then after every 32-bit data word for every control transaction. There is no 8-bit CRC after each 32-bit data word in data transactions because Ethernet frames include their own 32-bit CRC.

Table 26. Control Write Transaction with CRC

Table 27. Control Read Transaction with CRC

Table 28. Data Write Transaction with CRC

Table 29. Data Read Transaction with CRC

The generic SPI protocol is half duplex. Therefore, it is not possible to write frame data into the MAC_TX register and read from the MAC RX register at the same time. Because of this, the SPI SCLK frequency must be 25 MHz to achieve full duplex transmissions on Ethernet at 10 Mbps.

MAC Frame: Transmit and Receive

The 2-byte frame header shown in [Table 30](#page-32-0) is appended to all transmitted and received frames. This always precedes the frame data (see [Figure 21](#page-31-0)).

Time Stamp Capture

On receive, if TIME_STAMP_PRESET is asserted, an additional 4-byte or 8-byte time stamp is provided after the 2-byte header in [Table 30](#page-32-0) and before the data frame. This time stamp can then be stored or discarded by software when reading the receive FIFO.

On transmit, if EGRESS_CAPTURE is set other than 00, the ADIN1110 captures the time stamp of the transmitted frame into the respective TTSCxH and TTSCxL registers.

To capture time stamps, enable the counter by setting TS_EN (TS_CFG register) to 1 and setting FTSE (CONFIG0 register) to 1.

Transmit Frame over SPI

The following sequence must be followed when using the generic SPI protocol in store and forward mode:

- **1.** The device defaults to operating in store and forward mode.
- **2.** Verify that there is space for the frame by reading the transmit FIFO space register. The MAC internally appends a 2-byte size field to the frame in the FIFO, so ensure that there is sufficient space for the Ethernet frame plus 2-byte header plus 2-byte size field.
- **3.** Write the size of the frame in bytes including the 2-byte header to the MAC transmit frame size register. If the host has appended a frame check sequence (FCS) to the frame, this is also included in the size.
- **4.** Write the frame data including the 2-byte frame header to the transmit FIFO using MAC transmit register. The first byte for transmission is written to TXD, Bits[31:24]. The full frame can be written in a single burst or split up into multiple smaller burst writes. The burst write data must always be in multiples of 4 bytes, that is, the last word (4 bytes) can contain between 1 byte and 4 bytes of valid data.
- **5.** When the end of frame (EOF) byte of a frame is read from the transmit FIFO, the bit transmit ready asserts, and an interrupt triggers if the TX_RDY_MASK is set.

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MAC SPI

TRANSMIT: 2-BYTE FRAME HEADER TO THE TX REGISTER IN FRONT OF THE FRAME

ENTIRE FRAME CAN BE PASSED TO THE MAC-PHY IN A SINGLE SPI TRANSACTION OR MAY BE SPLIT INTO MULTIPLE TRANSACTIONS $\frac{8}{9}$

Figure 21. MAC Frame: Transmit

RECEIVE: 2-BYTE FRAME HEADER READ FIRST FROM THE P1_RX REGISTER

Figure 22. MAC Frame: Receive

Table 30. Frame Header

See the following definitions:

- ► Priority: indicates which priority queue the frame was received from. Not used on transmit. Set 0 in transmitted frames.
- ► EGRESS CAPTURE: capture an egress time stamp into the host readable egress time registers, as follows:
	- ► 00: no action.
	- ► 01: capture in the pair of TTSCAL and TTSCAH registers. The TTSCAA bits in the STATUS0 register assert when captured.
	- ► 10: capture in the pair of TTSCBL and TTSCBH registers. The TTSCAB bits in the STATUS0 register assert when captured.
	- ► 11: capture in the pair of TTSCCL and TTSCCH registers. The TTSCAC bits in the STATUS0 register assert when captured.
- ► TIME_STAMP_PARITY: odd parity for the appended time stamp. Not used on transmit. Set to 0 in transmitted frames.
- ► TIME_STAMP_PRESENT: on receive, the first 4 bytes or 8 bytes of data contain the time stamp for the frame. Not used on transmit. Set to 0 in transmitted frames.
- ► Reserved: always set to 0.

Receive Frame over SPI

The following procedure must be followed to receive an Ethernet frame when using the generic SPI protocol in store and forward mode:

- **1.** The device defaults to operating in store and forward mode.
- **2.** Set the P1_RX_RDY_MASK bit to 0 to enable an interrupt when a full frame is received.
- **3.** If the P1_RX_RDY bit is asserted, read the MAC receive frame size register to determine the size of the received frame.
- **4.** Read the frame via the MAC receive register. It is possible to burst read the entire frame or split it up into multiple smaller burst reads. The first byte of the received frame is returned in P1_RX, Bits[31:24]. The burst read transaction must be a multiple of 4 bytes. Some of the last 4 bytes are padded with 0s if the frame is not a multiple of 4 bytes in size.
- **5.** Read P1 RX RDY again. If the value of the bit is 1, another frame is available to read. Repeat from Step 3.

Cut Through

The generic SPI protocol supports cut through mode for transmit operations.

Before transmitting any frames, write 1 to the transmit cut through enable bits.

The threshold at which the frame transmit starts can be modified via the host transmit start threshold (see the [Transmit Threshold Regis](#page-55-0)[ter](#page-55-0) section). This register has a default value of 1. Therefore, by

default, transmit starts immediately on writing to the host transmit FIFO.

To ensure that the frame transmission does not under run, the host transmit FIFO has to be written at a rate greater than 10 Mbps. If the frame under runs, the host transmit under run error bit asserts.

Generic SPI Errors Generic SPI CRC Error

If an SPI CRC error occurs on a write to a register, the register is not written.

If the write is to the transmit register, the transmit FIFO is missing data and must be cleared by the host. Similarly, a read of the receive register has missing data in the receive frame, and the FIFO must be cleared.

If the errored transaction was a write to a configuration register, the SPI host must issue the write again. If the software does not know which configuration, the MAC must be reset by writing the RST_MAC_ONLY keys to the SOFT_RST register.

Generic SPI Transmit Protocol Error (TXPE)

TXPE asserts when the TX_FSIZE register is written, but the MAC still expects further writes to the transmit register related to the previous frame size written to the TX_FSIZE register. This error does not occur in normal operation and indicates an issue with the software driver, for example, two consecutive writes to the TX_FSIZE register without any writes to the transmit register.

In response to the assertion of TXPE, the host must clear the transmit FIFO.

OPEN Alliance SPI Protocol

The OPEN Alliance SPI protocol Version 1.0 can transfer data over the SPI using full duplex operation, achieving 10 Mbps bidirectional frame transfer with an SPI clock frequency in the region of 12 MHz to 16 MHz or greater.

The ADIN1110 supports the following OPEN Alliance SPI capabil-ities (see the [Supported Capabilities Register](#page-46-0) section for more details):

- ► Transmit FCS validation
- ► Cut through
- ► IEEE 1588 time stamp capture on transmit and receive
- \triangleright Minimum supported chunk size is 8 bytes

The OPEN Alliance SPI protocol defines two types of transactions: data transactions for Ethernet frame transfers and control transactions for register read/write operations.

A chunk is the basic element of data transactions, and they are composed of 4 bytes of overhead plus the configured payload size.

Data transactions consist of an equal number of transmit and receive chunks. Chunks in both transmit and receive directions may or may not contain valid frame data independent from each other, allowing for the simultaneous transmission and reception of different length frames. The data header of the chunk in transmit

frames, and the data footer in receive frames, indicate which bytes of the payload contain valid frame data. For full information on the OPEN Alliance SPI protocol used by the ADIN1110, refer to OPEN Alliance 10BASE-T1x MAC-PHY Serial Interface v1.0.

Note that \overline{CS} has to be deasserted between data transactions and control transactions, as shown in Figure 23.

Figure 23. Ethernet Data Frame Transfer Followed by Control Transfer

Data Chunks

Transmit data chunks consist of a 4-byte header followed by the transmit data chunk payload, as shown in Figure 24.

Figure 24. Transmit Data Chunk

Receive data chunks consist of the receive data chunk payload followed by a 4-byte footer, as shown in Figure 25

CHUNK PAYLOAD BYTES AND 4 BYTES RECEIVER HEADER *Figure 25. Receive Data Chunk*

The default size of the data chunk payload is 64 bytes. This size can be configured to 8 bytes, 16 bytes, 32 bytes, or 64 bytes via the chunk payload selector bits. The data chunk size must be configured before enabling data transmission or reception. Therefore, when the data chunk size is configured, it must not be changed without resetting the MAC-PHY.

Data Chunk Transactions

Data transactions consist of 1 to N chunks on SDO and SDI. The 4-byte data header occurs at the beginning of each transmit data chunk on SDO, and the 4-byte data footer occurs at the end of each data chunk on SDI. These headers and footers contain the information needed to determine the validity and location of the transmitted and received frames within the data chunk payload. The Ethernet frames start at any 32-bit aligned word within the payloads, as shown in Figure 26 and Figure 27.

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Transmit Data Header

Table 31. Transmit Data Header

See the following definitions:

- ► SEQ: data chunk sequence. The sequence functionality is not supported by the ADIN1110. This bit must be set to 0.
- ► NORX: no receive flag. The SPI host can set this bit to indicate to the MAC-PHY that it does not process receive frame data that is in the current receive data chunk. For normal operation, set NORX to 0 to indicate that it accepts and process any receive frame data within the current chunk.
- ► VS: vendor specific bits. This two bits need to be set to 00 by the host.
- ► DV: data valid flag. The SPI host uses this bit to indicate if the current chunk contains valid transmit data (DV = 1) or not. When this bit is 0, the MAC-PHY ignores the chunk payload.
- \triangleright SV: start valid flag. When this bit is 1, the beginning of an Ethernet frame is present in the current transmit data chunk payload. The SV bit is not to be confused with the start of frame delimiter (SFD) byte described in IEEE Standard 802.3.
- ► SWO: start word offset. When SV is 1, this field contains the 32-bit word offset into the transmit data chunk payload that points to the start of the new Ethernet frame. If SV is 0, the host must write this field as 0.
- \triangleright EV: end valid flag. When this bit is 1, the end of an Ethernet frame is present in the current transmit data chunk payload.
- ► EBO: end byte offset. When EV is 1, this field contains the byte offset into the transmit data chunk payload that points to the last byte of the Ethernet frame to transmit. If EV is 0, the host must write this field as 0.
- ► TSC: time stamp capture. Request a time stamp capture when the frame is transmitted onto the network. See the following: \blacktriangleright 00: no action.
	- ► 01: capture in the pair of TTSCAL and TTSCAH registers. The TTSCAA bits in the STATUS0 register assert when captured.
	- ► 10: capture in the pair of TTSCBL and TTSCBH registers. The TTSCAB bits in the STATUS0 register assert when captured.
	- ► 11: capture in the pair of TTSCCL and TTSCCH registers. The TTSCAC bits in the STATUS0 register assert when captured.
- ► P: parity. Parity bit calculated over the transmit data header. Method is odd parity.
- ► RSVD: reserved. Always set to 0.
Receive Data Footer

Table 32. Receive Data Footer

See the following definitions:

- ► EXST: extended status. This bit is set when any bit in the STATUS0 or STATUS1 registers are set and not masked.
- ► HDRB: received header bad. When this bit is set, the MAC-PHY has received a control or data header with a parity error.
- ► SYNC: configuration synchronized flag. This field reflects the state of the SYNC bit in the CONFIG0 register. When 0, this bit indicates that the MAC-PHY configuration may not be as expected by the SPI host. Following configuration, the SPI host sets the corresponding bit in the configuration register, which is reflected in this field.
- ► RCA: receive chunks available. The RCA field indicates the minimum number of additional receive data chunks of frame data that are available for reading beyond the current one. This field is 0 when there is no more receive frame data pending in the buffer of the MAC-PHY to be read.
- ► VS: vendor specific.
	- ► VS[1]: priority of the received frame.
		- \triangleright 0: frame received via the low priority queue.
		- \blacktriangleright 1: frame received via the high priority queue.
	- ► VS[0]: reserved.
- ► DV: data valid flag. The SPI host uses this bit to indicate if the current chunk contains valid transmit data ($DV = 1$) or not. When this bit is 0, the SPI host ignores the chunk payload.
- ► SV: start valid flag. When this bit is 1, the beginning of an Ethernet frame is present in the current transmit data chunk payload. The SV bit is not to be confused with the SFD byte described in IEEE Standard 802.3.
- ► SWO: start word offset. When SV is 1, this field contains the 32-bit word offset into the receive data chunk payload that points to the start of the new Ethernet frame. When a receive time stamp is added to the beginning of the received frame (RTSA = 1), SWO points to the most significant byte of the time stamp. If SV is 0, the host must write this field as 0.
- ► FD: frame drop. When set, this bit indicates that the MAC has detected a condition for which the SPI host must drop the received Ethernet frame. This bit is only valid at the end of a received frame $(EV = 1)$, and must be 0 at all other times.
- ► EV: end valid flag. When this bit is 1, the end of an Ethernet frame is present in the current receive data chunk payload.
- ► EBO: end byte offset. When EV is 1, this field contains the byte offset into the receive data chunk payload that points to the last byte of the received Ethernet frame. This field is 0 when EV = 0.
- ► RTSA: receive time stamp added. This bit is set when a 32-bit or 64-bit time stamp is added to the beginning of the SPI frame. This bit must be 0 when $SV = 0$.
- \blacktriangleright TXC: transmit credits. This field contains the minimum number of transmit data chunks of frame data that the SPI host can write in a single transaction without incurring a transmit buffer overflow.
- ► P: parity. Parity bit calculated over the receive data header. Method is odd parity.

OPEN Alliance SPI Cut Through Mode

If cut through from or to the host is enabled, the method to transfer frames remains the same as when using store and forward mode. However, the frame receive starts when sufficient frame data to fill a chunk is received, and the frame transmit starts when a configured transmit threshold is reached (see the [Transmit Threshold Register](#page-55-0) section).

The cut through mode can be enabled via the receive cut through enable bits and transmit cut through enable bits (see the [Configura](#page-47-0)[tion Register 0](#page-47-0) section).

On receive, the MAC returns data as it becomes available. Unlike in store and forward mode, there may be empty chunks $(DV = 0)$ between a start of frame (SOF) chunk and an end of frame (EOF) chunk.

If the host does not read frames fast enough to keep the receive FIFO empty, the frames are then buffered in the receive FIFO as if it is operating in store and forward mode. When all the frames are read, the FIFO returns to operating in cut through mode.

On transmit, the host must provide frame data at a rate fast enough (>10 Mbps) to ensure that the frame does not under run on transmit. If the MAC under runs, TXBUE in the STATUS0 register asserts and the MAC stops transmitting the frame in progress and appends a bad CRC to the frame.

Cut Through Transmit Latency

The time interval between the time the start of an SPI data transaction with a transmit header SWO of 0 (frame starts immediately in the chunk), and the time TX EN rises on the MII with an SPI frequency of 16 MHz and TX $THRESH = 1$ is 4 μs. The PHY transmit latency is 3.2 μs. This makes a total transmit latency of 7.2 μs.

Cut Through Receive Latency

The receive latency varies based on the chunk size and the SPI frequency. [Table 33](#page-37-0) indicates the latency for an SPI frequency of 16 MHz and all supported chunk sizes.

 1 Enough frame data to fill a chunk must be received before a transfer starts on the SPI. The time to receive the frame preamble is also included in this.

 $²$ Assuming that the μC is not waiting for an interrupt and that it is providing back-to-back OPEN Alliance data transactions on the SPI. The frame transfers start in the middle</sup> of the chunk on average.

 3 Realistically, the µC cannot use the data until it receives the receive header at the end of the chunk.

Control Transactions

Table 34. Control Command Header

Control transactions consist of one or more control commands. These commands are used by the SPI host to read and write registers within the MAC-PHY, and each one is composed of a 32-bit control command header followed by register data. See Table 34.

See the following definitions:

- ► HDRB: received header bad. When set by the MAC-PHY, HDRB indicates that a header was received with a parity error. The SPI host must always clear this bit. The MAC-PHY ignores this value.
- ► WNR: write not read. If 1, data is to be written to registers. Otherwise, data is to be read.
- ► AID: address increment disable. When clear, the address is automatically post-incremented by one following each register read or write.
- ► MMS: memory map selector. This field selects the specific register memory map to access. See Table 35.
- ► ADDR: address of the first register within the selected memory map to access.
- ► LEN: length. Specifies the number of registers to read/write. This field is interpreted as the number of registers − 1. Therefore, a length of 0 reads or writes a single register.
- ► P: parity. Parity bit calculated over the control command header. Method used is odd parity.

Control Write

The MAC-PHY ignores the final 32 bits of data from the SPI host at the end of the control write command. The write command and data is also echoed from the MAC-PHY back to the SPI so it can identify which register write failed in the case of any bus errors.

Control write commands can write either a single or multiple registers. When multiple registers are written, the address is automatically post incremented.

When a control write command is followed by another control command, the new control header must immediately follow the last word of the echoed register write data. The SPI host must deassert $\overline{\text{CS}}$ following the last word of the echoed register write data when the write command is the last command of the transaction.

Data Sheet **[ADIN1110](http:/www.analog.com/ADIN1110.html)**

MAC SPI

Control Read

The MAC-PHY ignores all data from the SPI host following the control header for the rest of the control read command. Control read commands can read either a single or multiple registers. When multiple registers are read, the address is automatically post-incremented according to the address increment disable bit in the control header.

Figure 29. Control Read Transaction

OPEN Alliance SPI Errors

See the following OPEN Alliance SPI errors:

► SPI Header Parity Error. If a parity error is detected on a transmit header and there is a transmit frame in the process of being transferred over SPI, this frame is dropped. If the MAC is operating in cut through mode, the frame transmit stops and a bad CRC is appended to the frame.

The MAC-PHY returns a fixed value of 0x4000_0000 in every word until CS goes high. The MAC-PHY responds with DV = 1, $EV = 1$, $EBO = 0$, and $FD = 1$ in the first data footer following a new assertion of CS.

If there is a parity error on a control transaction, the operation does not complete. Software can determine which transaction caused the error as the MAC-PHY returns a fixed 0x4000_0000 on SDO for the duration of the SPI transaction. Software can then resend the corrupted control transaction after clearing the header error bit.

► Transmit Protocol Error. Occurs when the MAC-PHY detects protocol errors in the transfer of transmit data chunks. These errors are usually due to SPI host firmware issues and do not occur in normal operation. The transmit protocol error bit is set when a data header received by the MAC-PHY indicates data valid (DV = 1) without a prior start of frame indication (SV = 1), in which case the data chunk is ignored. Or, when the MAC-PHY receives two data headers indicating a start of frame (SV = 1) without and end of frame $(EV = 1)$, the MAC-PHY drops the frame data from the previous start of frame indicator and begins accepting the frame data from the second start of frame indicator.

- ► Transmit Buffer Overflow. Occurs when attempting to write transmit frame data to the MAC-PHY when there is no transmit buffer space available as indicated by the transmit credit field (TXC) of the previous data footer. In this condition, the MAC-PHY ignores the transmit data chunk and sets the host transmit FIFO overflow bit, and the frame data already in the buffer is dropped.
- ► Transmit Buffer Under Run. This error can only occur in cut through mode. The SPI host must always send frame data to the MAC-PHY faster than the network to avoid this error. When this error occurs, the host transmit FIFO under run error bit is set, and the MAC-PHY terminates the frame being transmitted in a way that invalidates the frame. Additionally, the MAC-PHY ignores any additional frame data received from the SPI host until it receives an end of frame indication ($EV = 1$).
- \triangleright Loss of Framing Error. This error occurs when the $\overline{\text{CS}}$ signal is deasserted before the expected end of the data chunk or control command. The MAC-PHY and the loss of frame error is set, any transmit frame in progress is dropped, and any receive frame in progress of being sent to the SPI host is terminated.
- ► Receive Buffer Overflow. This error occurs when the SPI host does not read frame data from the MAC-PHY fast enough. This error can occur both in store and forward and cut through modes. When this error occurs, the MAC-PHY terminates the frame being received from the PHY. In store and forward mode, no portion of the frame is transferred to the SPI host. In cut through mode, the MAC-PHY terminates the frame (EV = 1) with frame drop set $(FD = 1)$.
- ► Control Data Protection Error. The control data protection error (CDPE) and the loss of frame error (LOFE) bits assert when protection is enabled on the OPEN Alliance SPI and there is an error on write data received from the host. The write does not complete in this case.

If possible, the software executes the write again. If software does not know which configuration register was written, the device might not be configured properly. In this case, the MAC must be reset by writing the RST_MAC_ONLY keys to the software reset register.

SPI Access to the PHY Registers

The ADIN1110 provides indirect access using the SPI to access the PHY management registers. The 8 registers MDIOACCn in the SPI register map are used to access the PHY management registers. Each MDIOACCn register corresponds to an MDIO transaction.

The MDC default speed is 2.5 MHz. Either 2.5 MHz or 4.166 MHz MDC frequency can be selected via the MSPEED bits in the CONFIG2 register.

The MDIO master polls in round robin mode the TRDONE bits of the eight MDIOACCn registers. When the MDIO master detects that one of the TRDONE fields is 0, an MDIO transaction is started by the MDIO master. When the MDIO transaction completes, the

TRDONE bits are set to 1, and the master proceeds to check the TRDONE bits of the next MDIOACCn register.

Note that MDIO_DEVAD is always written with the device ID of the register being accessed, MDIO_PRTAD is always written to 0x1, and MDIO ST is written to 0x0 for Clause 45 access (this applies to all of the following examples).

Example write to PHY Register XYZ:

- **1.** Write MDIOACC0 with MDIO_DATA = the address of Register XYZ, MDIO_DEVAD = the device ID of Register XYZ, MDIO PRTAD = 0x1, MDIO OP = 0x0(ADDR), MDIO ST = $0x0$, and TRDONE = $0x0$.
- **2.** Write MDIOACC1 with MDIO_DATA = the value to be written to Register XYZ, MDIO $OP = 0x1(WR)$ and TRDONE = 0x0.
- **3.** Optionally, poll MDIOACC0.TRDONE = 0x1 to determine that the write address operation has completed.
- **4.** Poll MDIOACC1.TRDONE = 0x1 to determine that the write data operation has completed.

Example read of PHY Register XYZ:

- **1.** Write MDIOACC0 with MDIO_DATA = the address of Register XYZ, MDIO $OP = 0x0(ADDR)$, and TRDONE = 0x0.
- **2.** Write MDIOACC1 with MDIO OP = 0x3(RD) and TRDONE = 0x0.
- **3.** Poll MDIOACC1. TRDONE= 0x1 to determine that the write data operation has completed. MDIOACC1. MDIO_DATA reflects the content of MDIO Register XYZ.

Example write operation followed by a read to verify the write operation:

- **1.** Write MDIOACC0 with MDIO_DATA = the address of register ABC and TRDONE = 0x0.
- **2.** Write MDIOACC1 with MDIO_DATA = the value to be written to register ABC, MDIO_OP = $0x1(WR)$, and TRDONE = $0x0$.
- **3.** Write MDIOACC2 MDIO $OP = 0x3(RD)$ and TRDONE = 0x0.
- **4.** Poll MDIOACC2.TRDONE = 0x1 to verify that all operations are completed. MDIO_DATA reflects the content of register ABC.

Example of four consecutive writes. It is possible to write a command to all 8 register before checking any.

- **1.** Write MDIOACC0 with MDIO_DATA = the address of register ABC and TRDONE = 0x0.
- **2.** Write MDIOACC1 with the write data for register ABC, MDIO $OP = 0x1$, and TRDONE = 0x0.
- **3.** Write MDIOACC2 with MDIO_DATA = the address of register DEF and TRDONE = 0x0.
- **4.** Write MDIOACC3 with the write data for register DEF, MDIO $OP = 0x1$, and TRDONE = 0x0.
- **5.** Write MDIOACC4 with MDIO_DATA = the address of register GHJ and TRDONE = 0x0.
- **6.** Write MDIOACC5 with the write data for register GHJ, MDIO $OP = 0x1$, and TRDONE = 0x0.

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MAC SPI

- **7.** Write MDIOACC6 with MDIO_DATA = the address of Register XYZ and TRDONE = 0x0.
- **8.** Write MDIOACC7 with the write data for Register XYZ, MDIO $OP = 0x1$, and TRDONE = 0x0.
- **9.** Host polls MDIOACC7. TRDONE = 0x1 to verify that all write data operations are complete.

Example burst read starting from Register XYZ:

- **1.** Write MDIOACC0 with MDIO_DATA = the address of the Register XYZ, MDIO_OP = $0x0(ADDR)$, and TRDONE = $0x0$
- **2.** Write MDIOACC1 with MDIO OP = 0x2(INC_RD) and $TROONE = 0x0.$
- **3.** Write MDIOACC2 with MDIO_OP = 0x2(INC_RD) and TRDONE = 0x0.
- **4.** Write MDIOACC3 with MDIO_OP = 0x2(INC_RD) and TRDONE = 0x0.
- **5.** Write MDIOACC4 with MDIO_OP = 0x2(INC_RD) and TRDONE = 0x0.
- **6.** Write MDIOACC5 with MDIO_OP = 0x2(INC_RD) and TRDONE = 0x0.
- **7.** Write MDIOACC6 with MDIO_OP = 0x2(INC_RD) and TRDONE = 0x0.
- **8.** Write MDIOACC7 with MDIO_OP = 0x2(INC_RD) and $TROONE = 0x0.$
- **9.** Poll MDIOACC7. TRDONE = 1 to verify that all read data operations are complete.
- **10.** Read MDIOACC1. MDIO_DATA, reflects the content of Register XYZ.
- **11.** Read MDIOACC2. MDIO_DATA, reflects the content of register at address XYZ. ADDR + 1.
- **12.** Read MDIOACC3. MDIO_DATA, reflects the content of register at address XYZ. ADDR + 2.
- **13.** Read MDIOACC4. MDIO_DATA, reflects the content of register at address XYZ. ADDR + 3.
- **14.** Read MDIOACC5. MDIO_DATA, reflects the content of register at address XYZ. ADDR + 4.
- **15.** Read MDIOACC6. MDIO_DATA, reflects the content of register at address XYZ. ADDR + 5.
- **16.** Read MDIOACC7. MDIO_DATA, reflects the content of register at address XYZ. ADDR + 6.

Example of Clause 22 write of Register XYZ:

- **1.** Write MDIOACC0 with MDIO_DATA = write data, MDIO DEV $AD =$ the address of the Register XYZ, MDIO PRTAD = 0x1, MDIO OP = 0x1(WR), MDIO ST = $0x1$ (Clause 22), and TRDONE = $0x0$.
- **2.** Poll MDIOACC0. TRDONE= 0x1 to determine that the write data operation is complete.

Example of Clause 22 read of Register XYZ:

- **1.** Write MDIOACC0 with MDIO_DEV_AD = the address of the Register XYZ, MDIO PRTAD = 0x1, MDIO OP = 0x3(RD), MDIO $ST = 0x1$ (Clause 22), and TRDONE = 0x0.
- **2.** Poll MDIOACC0. TRDONE = 0x1 to determine that the read operation is complete. MDIO_DATA reflects the contents of MDIO Register XYZ.

Example of Clause 22 write and read back of Register XYZ:

- **1.** Write MDIOACC0 with MDIO_DATA = write data, MDIO DEV $AD =$ the address of the Register XYZ, MDIO \overline{PRTAD} = 0x1, MDIO \overline{OP} = 0x1(WR), MDIO \overline{ST} = $0x1$ (Clause 22), and TRDONE = $0x0$.
- **2.** Write MDIOACC1 with MDIO_DEV_AD = the address of the Register XYZ, MDIO_PRTAD = 0x1, MDIO_OP = 0x3(RD), MDIO_ST = 0x1(Clause 22), and TRDONE = 0x0.
- **3.** Poll MDIOACC1. TRDONE = 0x1 to determine that the read operation is complete. MDIO_DATA reflects the contents of MDIO Register XYZ.

MDIO PHY Address Determination

The MDIO PHY address for the ADIN1110 PHY is 0x1.

PHY Registers Contents

The PHY registers provide access to control and status information in the management registers.

The registers of the PHY Clause 45 register map are made up of four device address groupings (see Table 36) based on the MDIO manageable device (MMD). Within each device address space, IEEE standard registers are located in register addresses between 0x0000 and 0x7FFF, and vendor specific registers are located in register addresses from 0x8000 to 0xFFFF.

Table 36. Clause 45 Register Groupings

Clause 45 can access to up to 32 PHYs consisting of up to 32 MMDs through a single MDIO interface.

The default value of some of the registers are determined by the value of the hardware configuration pins, which are read just after the RESET pin is deasserted. In these cases, the reset value in the register table is listed as pin dependent, which allows the default operation of the ADIN1110 to be configured without having to write to it over the SPI. This method is useful in unmanaged applications where the desired operation of the PHY is configured from the hardware configuration pins without any software intervention. For unmanaged applications, do not configure the PHY to enter software power-down mode after reset to ensure that the PHY

immediately attempts to bring up links as configured by the other hardware configuration pins. In managed applications, software is available to configure the PHY via the management interface. In this case, it is possible to use the hardware configuration pins to configure the PHY to enter software power-down mode after reset, such that the PHY can be configured before linking is attempted.

Recommended Register Operation

Many of the PHY registers in the ADIN1110 are defined in the IEEE Standard 802.3, and the exact behavior of these registers follows the standard. This behavior may not always be obvious and is described in this section, including the recommended operation and use of the registers.

Latch Low Registers

The IEEE Standard 802.3-2018 requires certain MDIO accessible registers to exhibit latch low behavior. The idea is to allow software that only intermittently reads these registers to detect conditions that can be transitory or short lived. For example, the AN_LINK_STATUS bit is required to latch low. When the device exits from a reset or power-down state, the latching condition is not active and the value of the AN_LINK_STATUS bit reflects the current status of the link. However, if the link comes up and drops, the latching condition is active. In this case, the AN_LINK_STATUS bit reads as 0 even if the link has come back up again in the interim. The latching condition is only cleared when the AN_LINK_STA-TUS bit is read, ensuring the software has had the opportunity to observe that the link dropped.

One implication of this latch low behavior is that, if software wishes to determine the current status of the link, it must perform two reads of the AN_LINK_STATUS bit back to back. The first read is needed to clear any active latching condition.

Another implication is that it is important that software take account of the interaction between MDIO accessible bits that share a register address. For example, the AN_PAGE_RX bits and AN_LINK_STATUS bits reside at the same register address. As a result, reading the AN_PAGE_RX bits clears any active latching condition associated with the AN_LINK_STATUS bits.

IEEE Duplicated Registers

The IEEE Standard 802.3-2018 covers a very wide range of standards and speeds, from 10 Mbps to 40 Gbps and higher, and includes a very large number of clauses. There are registers associated with many clauses, and different PHYs can include different clauses and combinations of clauses. Therefore, registers for common functions like software reset, software power-down, loopback, and so on, tend to be implemented in multiple clauses.

In the ADIN1110, the physical implementation of these registers is in a single location, but they can be accessed at multiple addresses. For example, the software reset bit, can be read or written in

all the following IEEE MMD locations and vendor specific register locations:

- ► PMA_SFT_RST
- ► B10L_PMA_SFT_RST
- ► PCS_SFT_RST
- ► B10L_PCS_SFT_RST
- ► CRSM_SFT_RST

In this example, these are the PMA/PMD, PCS, autonegotiation, and Vendor Specific MMD 1 device address locations (per [Table](#page-40-0) [36](#page-40-0)).

Having multiple address locations for the same register makes the use of the device more complex than necessary, particularly in relation to registers that have latch low or self clear access permissions. This is an unavoidable consequence of the IEEE standard.

The ADIN1110 data sheet only calls out a single recommended address location for each of these IEEE registers to simplify the operation and use of the device. In general, the registers introduced in the 802.3cg (10BASE-T1L) section of the standard are recommended over older (equivalent) registers. Often, registers in a vendor specific address are recommended, particularly where a register brings a number of useful IEEE register bits into a single register address. The ADIN1110 responds to register accesses to all the IEEE register address locations covered by the 10BASE-T1L standard when the start-up is complete after a power-on reset, hardware reset, or software reset.

Read Modify Write Operation

All register write operations must be performed as read modify write operations. If this process is not followed, the value of the register bits can inadvertently change.

MAC

Frame Filtering on Receive

By default, the device filters all frames received. To receive frames, set up the address filtering table, or the default operation for all received frames can be changed.

The device can be configured to filter up to 16 different MAC addresses based on the destination MAC address (DA).

To receive frames with a particular DA, that DA has to be programmed to one of the 16 ADDR FILT x registers. Each register is 32 bits wide. Therefore, for example, to program a DA of 0800 005A 646B to ADDR_FILT[0], write the following:

- **1.** 0x0800 to ADDR_FILT_UPR0.
- **2.** 0x005A6468 to ADDR_FILT_LWR0.

To forward frames with this DA to the host, set the TO_HOST bit within the ADDR FILT UPRn to 1. To apply this rule, set the APPLY2PORT1 bit to 1.

MAC addresses can be masked using the ADDR_MSK_x registers. For example, to receive all the MAC addresses in the range 0x8000 005A 64xx, write:

- **1.** 0xFFFF to ADDR_MSK_UPR0.
- **2.** 0XFFFFFF00 to ADDR_MSK_LWR0.

Frames that do not match any of the 16 ADDR_FILT_x registers are dropped by default. If the P1_FWD_UNK2HOST bits within the CONFIG2 register are set to 1, all frames that do not match a DA are forwarded to the host. Figure 30 shows the filtering algorithm.

Figure 30. Filtering Algorithm

Frames received with a bad CRC or with RX_ER asserted from the PHY, as well as runt and jabber frames, are dropped and counted.

Receive Priority Queues

There are two different FIFOs on receive: a high priority FIFO and a low priority FIFO.

By default, the low priority FIFO is configured to 12 kB, and the high priority FIFO is configured to 8 kB. The sizes of these FIFOs can be changed before receiving or transmitting any frames via the P1_RX_LO_SIZE and P1_RX_HI_SIZE fields in the FIFO_SIZE register.

Frames are always returned from the high priority FIFO first.

Statistics Counters

There are 15 32-bit counters on the receive port that increment on each frame transmit and receive.

Receive Drop FIFO Full Counter

Before the first byte of a received frame is written into the appropriate receive FIFO, the space in the FIFO is checked. If there is no space for at least 256 bytes, the frame is dropped and the P1_RX_DROP_FULL_CNT counter increments. If there is space for at least 256 bytes in the FIFO, the logic commences writing the frame to the receive FIFO. If the received frame exceeds 256 bytes and the receive FIFO fills, the frame is dropped and the P1_RX_DROP_FULL_CNT counter increments.

Frame Receive and Transmit Errors

By default, all received errored frames are dropped and counted. Received errored frames do not generate interrupts. Instead, they are dropped and counted, and the software must monitor the statistics counters.

SRAM ECC Error

When writing a frame to the FIFO, the size of the frame is inserted in a 16-bit word at the front of the frame and written to the FIFO. A 5-bit error correction code (ECC) is placed alongside the size field.

When this location is read from static random-access memory (SRAM), the ECC is checked. If a double bit error is detected, the RX_ECC_ERR or TX_ECC_ERR bits of the STATUS1 register assert. If a double bit error is detected on reading a frame header from the receive FIFO, the frame is not transmitted.

In response to an ECC error, a FIFO automatically clears. All frames in the FIFO are lost, transmission stops, and a bad CRC

is appended to the frame that was transmitted. The next frame received is written to a FIFO.

SPI REGISTER DETAILS

Table 38. SPI Register Map

Table 38. SPI Register Map (Continued)

Identification Version Register

Address: 0x00, Reset: 0x00000010, Name: IDVER

Table 39. Bit Descriptions for IDVER

PHY Identification Register

Address: 0x01, Reset: 0x0283BC91, Name: PHYID

Table 40. Bit Descriptions for PHYID

Supported Capabilities Register

Address: 0x02, Reset: 0x000006C3, Name: CAPABILITY

Table 41. Bit Descriptions for CAPABILITY

Table 41. Bit Descriptions for CAPABILITY (Continued)

Reset Control and Status Register

Address: 0x03, Reset: 0x00000000, Name: RESET

Table 42. Bit Descriptions for RESET

Configuration Register 0

Address: 0x04, Reset: 0x00000006, Name: CONFIG0

Table 43. Bit Descriptions for CONFIG0

Table 43. Bit Descriptions for CONFIG0 (Continued)

Configuration Register 2

Address: 0x06, Reset: 0x00000800, Name: CONFIG2

Vendor specific.

Table 44. Bit Descriptions for CONFIG2

Status Register 0

Address: 0x08, Reset: 0x00000040, Name: STATUS0

Table 45. Bit Descriptions for STATUS0

Table 45. Bit Descriptions for STATUS0 (Continued)

Status Register 1

Address: 0x09, Reset: 0x00000000, Name: STATUS1

Table 46. Bit Descriptions for STATUS1

Table 46. Bit Descriptions for STATUS1 (Continued)

Buffer Status Register

Address: 0x0B, Reset: 0x00007700, Name: BUFSTS

Table 47. Bit Descriptions for BUFSTS

Interrupt Mask Register 0

Address: 0x0C, Reset: 0x00001FBF, Name: IMASK0

Table 48. Bit Descriptions for IMASK0

Table 48. Bit Descriptions for IMASK0 (Continued)

Mask Bits for Driving the Interrupt Pin Register

Address: 0x0D, Reset: 0x43FA1F1A, Name: IMASK1

Table 49. Bit Descriptions for IMASK1

Transmit Time Stamp Capture Register A (High)

Address: 0x10, Reset: 0x00000000, Name: TTSCAH

This field contains the upper 32 bits of the captured time stamp for when the requested frame was transmitted.

Table 50. Bit Descriptions for TTSCAH

Transmit Time Stamp Capture Register A (Low)

Address: 0x11, Reset: 0x00000000, Name: TTSCAL

This field contains the lower 32 bits of the captured time stamp for when the requested frame was transmitted.

Table 51. Bit Descriptions for TTSCAL

Transmit Time Stamp Capture Register B (High)

Address: 0x12, Reset: 0x00000000, Name: TTSCBH

This field contains the upper 32 bits of the captured time stamp for when the requested frame was transmitted.

Table 52. Bit Descriptions for TTSCBH

Transmit Time Stamp Capture Register B (Low)

Address: 0x13, Reset: 0x00000000, Name: TTSCBL

This field contains the lower 32 bits of the captured time stamp for when the requested frame was transmitted.

Table 53. Bit Descriptions for TTSCBL

Transmit Time Stamp Capture Register C (High)

Address: 0x14, Reset: 0x00000000, Name: TTSCCH

This field contains the upper 32 bits of the captured time stamp for when the requested frame was transmitted.

Table 54. Bit Descriptions for TTSCCH

Transmit Time Stamp Capture Register C (Low)

Address: 0x15, Reset: 0x00000000, Name: TTSCCL

This field contains the lower 32 bits of the captured time stamp for when the requested frame was transmitted.

Table 55. Bit Descriptions for TTSCCL

MDIO Access Registers

Address: 0x20 to 0x27 (Increments of 1), Reset: 0x8C000000, Name: MDIOACCn

Use this register to access the PHY registers via the SPI to MDIO bridge.

Table 56. Bit Descriptions for MDIOACCn

MAC Tx Frame Size Register

Address: 0x30, Reset: 0x00000000, Name: TX_FSIZE

Table 57. Bit Descriptions for TX_FSIZE

MAC Transmit Register

Address: 0x31, Reset: 0x00000000, Name: TX

The transmit FIFO is written via this register.

Table 58. Bit Descriptions for TX

Tx FIFO Space Register

Address: 0x32, Reset: 0x00000FFF, Name: TX_SPACE

Table 59. Bit Descriptions for TX_SPACE

Transmit Threshold Register

Address: 0x34, Reset: 0x00000041, Name: TX_THRESH

Table 60. Bit Descriptions for TX_THRESH

MAC FIFO Clear Register

Address: 0x36, Reset: 0x00000000, Name: FIFO_CLR

Table 61. Bit Descriptions for FIFO_CLR

Scratch Registers

Address: 0x37 to 0x3A (Increments of 1), Reset: 0x00000000, Name: SCRATCHn

Table 62. Bit Descriptions for SCRATCHn

MAC Reset Status Register

Address: 0x3B, Reset: 0x00000003, Name: MAC_RST_STATUS

If this register returns 0x00000000_00000001 when read, the oscillator clock is active, but the 25 MHz crystal clock is not active.

If this register returns 0x00000000_00000003 when read, both the oscillator clock and the 25 MHz crystal clock are active.

If this register returns 0x00000000_0000000 (SDO output pad is enabled while \overline{CS} is low), the SPI slave and MAC core are both still in reset.

Only single SPI reads of this register are supported. An SPI burst read must not increment into this register.

Table 63. Bit Descriptions for MAC_RST_STATUS

Software Reset Register

Address: 0x3C, Reset: 0x00000000, Name: SOFT_RST

Table 64. Bit Descriptions for SOFT_RST

Inject an Error on MISO from the DUT Register

Address: 0x3D, Reset: 0x00000000, Name: SPI_INJ_ERR

Table 65. Bit Descriptions for SPI_INJ_ERR

FIFO Sizes Register

Address: 0x3E, Reset: 0x00000464, Name: FIFO_SIZE

Before modifying the FIFO sizes, frame reception and transmission must be stopped and the FIFOs must be empty.

Configure the forwarding rules to drop all frames and set P1_UNK2HOST to 0 to ensure all received frames are dropped.

Use RXF CLR & TXF CLR to reset the FIFOs. Then, the FIFO sizes can be modified.

The total FIFO size must be less than or equal to 28 kB.

Table 66. Bit Descriptions for FIFO_SIZE

Tx FIFO Frame Count Register

Address: 0x3F, Reset: 0x00000000, Name: TFC

For debug only. Number of frames in the transmit FIFO.

Table 67. Bit Descriptions for TFC

Tx FIFO Valid Half Words Register

Address: 0x40, Reset: 0x00000000, Name: TXSIZE

Number of Valid Half Words (16 Bit) in the Host Tx FIFO.

Table 68. Bit Descriptions for TXSIZE

Host Tx Frames Dropped Due to FIFO Overflow Register

Address: 0x41, Reset: 0x00000000, Name: HTX_OVF_FRM_CNT

Address of a Detected ECC Error in Memory Register

Address: 0x42, Reset: 0x00000000, Name: MECC_ERR_ADDR

Table 70. Bit Descriptions for MECC_ERR_ADDR

Corrected ECC Error Counters Register

Address: 0x43 to 0x49 (Increments of 1), Reset: 0x00000000, Name: CECC_ERRn

Table 71. Bit Descriptions for CECC_ERRn

MAC Address Rule and DA Filter Upper 16 Bits Registers

Address: 0x50 to 0x6E (Increments of 2), Reset: 0x00000000, Name: ADDR_FILT_UPRn

Contains the upper 16 bits of a MAC address and the filtering rule associated with the MAC address.

When writing the ADDR_FILT_x registers, two register locations must be written in order for a given table entry.

For example, to write table entry 0, the registers must be written in the following order:

1. ADDR_FILT_UPR0.

2. ADDR_FILT_LWR0.

Table 72. Bit Descriptions for ADDR_FILT_UPRn

Table 72. Bit Descriptions for ADDR_FILT_UPRn (Continued)

MAC Address DA Filter Lower 32 Bits Registers

Address: 0x51 to 0x6F (Increments of 2), Reset: 0x00000000, Name: ADDR_FILT_LWRn

Contains the lower 32 bits of a MAC address in the DA filter table.

A write to one of these registers must be preceded by a write to the corresponding ADDR_FILT_UPRn register.

Table 73. Bit Descriptions for ADDR_FILT_LWRn

Upper 16 Bits of the MAC Address Mask Register

Address: 0x70 to 0x72 (Increments of 2), Reset: 0x0000FFFF, Name: ADDR_MSK_UPRn

The upper 16 bits of a MAC address mask in the DA mask table.

When writing the ADDR, MSK, x registers, all two register locations must be written in order for a given table entry. They must be written in order with the UPR register written first and the LWR register written last.

Table 74. Bit Descriptions for ADDR_MSK_UPRn

Lower 32 Bits of the MAC Address Mask Register

Address: 0x71 to 0x73 (Increments of 2), Reset: 0xFFFFFFFF, Name: ADDR_MSK_LWRn

The lower 32 bits of a MAC address mask in the DA mask table.

When writing the ADDR_MSK_x registers, all two register locations must be written in order for a given table entry. The register locations must be written in order with the UPR register written first and the LWR register written last.

Table 75. Bit Descriptions for ADDR_MSK_LWRn

Time Stamp Accumulator Addend Register

Address: 0x80, Reset: 0x85555555, Name: TS_ADDEND

Table 76. Bit Descriptions for TS_ADDEND

Timer Update Compare Register

Address: 0x81, Reset: 0x3B9ACA00, Name: TS_1SEC_CMP

Seconds Counter Register

Address: 0x82, Reset: 0x00000000, Name: TS_SEC_CNT

Use this register to write to the seconds counter.

Table 78. Bit Descriptions for TS_SEC_CNT

Nanoseconds Counter Register

Address: 0x83, Reset: 0x00000000, Name: TS_NS_CNT

Use this register to write to the nanoseconds counter.

Table 79. Bit Descriptions for TS_NS_CNT

Timer Configuration Register

Address: 0x84, Reset: 0x00000000, Name: TS_CFG

Table 80. Bit Descriptions for TS_CFG

Table 80. Bit Descriptions for TS_CFG (Continued)

High Period for TS_TIMER Register

Address: 0x85, Reset: 0x00000000, Name: TS_TIMER_HI

Table 81. Bit Descriptions for TS_TIMER_HI

Low Period for TS_TIMER Register

Address: 0x86, Reset: 0x00000000, Name: TS_TIMER_LO

Table 82. Bit Descriptions for TS_TIMER_LO

Quantization Error Correction Register

Address: 0x87, Reset: 0x00000000, Name: TS_TIMER_QE_CORR

Table 83. Bit Descriptions for TS_TIMER_QE_CORR

TS_TIMER Counter Start Time Register

Address: 0x88, Reset: 0x00000000, Name: TS_TIMER_START

Point in Time at Which to Start the TS_TIMER Counter.

Table 84. Bit Descriptions for TS_TIMER_START

Table 84. Bit Descriptions for TS_TIMER_START (Continued)

TS_CAPT Pin 0 Time Stamp Register

Address: 0x89, Reset: 0x00000000, Name: TS_EXT_CAPT0

Time stamp captured on the assertion of the TS_CAPT pin.

Table 85. Bit Descriptions for TS_EXT_CAPT0

TS_CAPT Pin 1 Time Stamp Register

Address: 0x8A, Reset: 0x00000000, Name: TS_EXT_CAPT1

Time stamp captured on the assertion of the TS_CAPT pin.

Table 86. Bit Descriptions for TS_EXT_CAPT1

TS_CAPT Free Running Counter Register

Address: 0x8B, Reset: 0x00000000, Name: TS_FREECNT_CAPT

Capture of the free running counter when TS_CAPT asserts.

Table 87. Bit Descriptions for TS_FREECNT_CAPT

P1 MAC Rx Frame Size Register

Address: 0x90, Reset: 0x00000000, Name: P1_RX_FSIZE

Table 88. Bit Descriptions for P1_RX_FSIZE

P1 MAC Receive Register

Address: 0x91, Reset: 0x00000000, Name: P1_RX

The receive FIFO is read via this register.

It is possible to burst read data from the Rx FIFO over SPI.

Table 89. Bit Descriptions for P1_RX

P1 Rx Frame Count Register

Address: 0xA0, Reset: 0x00000000, Name: P1_RX_FRM_CNT

Table 90. Bit Descriptions for P1_RX_FRM_CNT

P1 Rx Broadcast Frame Count Register

Address: 0xA1, Reset: 0x00000000, Name: P1_RX_BCAST_CNT

Table 91. Bit Descriptions for P1_RX_BCAST_CNT

P1 Rx Multicast Frame Count Register

Address: 0xA2, Reset: 0x00000000, Name: P1_RX_MCAST_CNT

Table 92. Bit Descriptions for P1_RX_MCAST_CNT

P1 Rx Unicast Frame Count Register

Address: 0xA3, Reset: 0x00000000, Name: P1_RX_UCAST_CNT

Table 93. Bit Descriptions for P1_RX_UCAST_CNT

P1 Rx CRC Errored Frame Count Register

Address: 0xA4, Reset: 0x00000000, Name: P1_RX_CRC_ERR_CNT

Table 94. Bit Descriptions for P1_RX_CRC_ERR_CNT

P1 Rx Align Error Count Register

Address: 0xA5, Reset: 0x00000000, Name: P1_RX_ALGN_ERR_CNT

Table 95. Bit Descriptions for P1_RX_ALGN_ERR_CNT

P1 Rx Long/Short Frame Error Count Register

Address: 0xA6, Reset: 0x00000000, Name: P1_RX_LS_ERR_CNT

Table 96. Bit Descriptions for P1_RX_LS_ERR_CNT

P1 Rx PHY Error Count Register

Address: 0xA7, Reset: 0x00000000, Name: P1_RX_PHY_ERR_CNT

Table 97. Bit Descriptions for P1_RX_PHY_ERR_CNT

P1 Tx Frame Count Register

Address: 0xA8, Reset: 0x00000000, Name: P1_TX_FRM_CNT

Table 98. Bit Descriptions for P1_TX_FRM_CNT

P1 Tx Broadcast Frame Count Register

Address: 0xA9, Reset: 0x00000000, Name: P1_TX_BCAST_CNT

Table 99. Bit Descriptions for P1_TX_BCAST_CNT

P1 Tx Multicast Frame Count Register

Address: 0xAA, Reset: 0x00000000, Name: P1_TX_MCAST_CNT

Table 100. Bit Descriptions for P1_TX_MCAST_CNT

P1 Tx Unicast Frame Count Register

Address: 0xAB, Reset: 0x00000000, Name: P1_TX_UCAST_CNT

Table 101. Bit Descriptions for P1_TX_UCAST_CNT

P1 Rx Frames Dropped Due to FIFO Full Register

Address: 0xAC, Reset: 0x00000000, Name: P1_RX_DROP_FULL_CNT

Table 102. Bit Descriptions for P1_RX_DROP_FULL_CNT

P1 Rx Frames Dropped Due to Filtering Register

Address: 0xAD, Reset: 0x00000000, Name: P1_RX_DROP_FILT_CNT

Table 103. Bit Descriptions for P1_RX_DROP_FILT_CNT

Frame Received on Port 1 with IFG Errors Register

Address: 0xAE, Reset: 0x00000000, Name: P1_RX_IFG_ERR_CNT

Table 104. Bit Descriptions for P1_RX_IFG_ERR_CNT

P1 Transmit Interframe Gap Register

Address: 0xB0, Reset: 0x0000000B, Name: P1_TX_IFG

Table 105. Bit Descriptions for P1_TX_IFG

P1 MAC Loopback Enable Register

Address: 0xB3, Reset: 0x00000000, Name: P1_LOOP

Table 106. Bit Descriptions for P1_LOOP

P1 CRC Check Enable on Receive Register

Address: 0xB4, Reset: 0x00000001, Name: P1_RX_CRC_EN

Table 107. Bit Descriptions for P1_RX_CRC_EN

Table 107. Bit Descriptions for P1_RX_CRC_EN (Continued)

P1 Receive Interframe Gap Register

Address: 0xB5, Reset: 0x0000000A, Name: P1_RX_IFG

Table 108. Bit Descriptions for P1_RX_IFG

P1 Max Receive Frame Length Register

Address: 0xB6, Reset: 0x00000618, Name: P1_RX_MAX_LEN

Maximum receive frame length in bytes.

Table 109. Bit Descriptions for P1_RX_MAX_LEN

P1 Min Receive Frame Length Register

Address: 0xB7, Reset: 0x00000040, Name: P1_RX_MIN_LEN

Minimum receive frame length in bytes.

Table 110. Bit Descriptions for P1_RX_MIN_LEN

P1 Rx Low Priority FIFO Frame Count Register

Address: 0xB8, Reset: 0x00000000, Name: P1_LO_RFC

The number of frames in the receive FIFO.

Table 111. Bit Descriptions for P1_LO_RFC

P1 Rx High Priority FIFO Frame Count Register

Address: 0xB9, Reset: 0x00000000, Name: P1_HI_RFC

The number of frames in the receive FIFO.

Table 112. Bit Descriptions for P1_HI_RFC

P1 Low Priority Rx FIFO Valid Half Words Register

Address: 0xBA, Reset: 0x00000000, Name: P1_LO_RXSIZE

Number of valid half words (16 bits) in the low priority Rx FIFO.

Table 113. Bit Descriptions for P1_LO_RXSIZE

P1 High Priority Rx FIFO Valid Half Words Register

Address: 0xBB, Reset: 0x00000000, Name: P1_HI_RXSIZE

Number of valid half words (16 bits) in the high priority Rx FIFO.

Table 114. Bit Descriptions for P1_HI_RXSIZE

PHY CLAUSE 22 REGISTER DETAILS

Table 115. PHY Clause 22 Register Summary

MII Control Register

Address: 0x0, Reset: 0x1100, Name: MI_CONTROL

This address corresponds to the MII control register specified in Clause 22.2.4.1 of Standard 802.3.

Table 116. Bit Descriptions for MI_CONTROL (Continued)

MII Status Register

Address: 0x1, Reset: 0x1009, Name: MI_STATUS

This address corresponds to the MII status register specified in Clause 22.2.4.2 of Standard 802.3.

Table 117. Bit Descriptions for MI_STATUS

Table 117. Bit Descriptions for MI_STATUS (Continued)

PHY Identifier 1 Register

Address: 0x2, Reset: 0x0283, Name: MI_PHY_ID1

The PHY Identifier 1 address allows 16 bits of the OUI to be observed.

Table 118. Bit Descriptions for MI_PHY_ID1

PHY Identifier 2 Register

Address: 0x3, Reset: 0xBC91, Name: MI_PHY_ID2

The PHY Identifier 2 address allows six bits of the OUI, the model number, and revision number to be observed.

Table 119. Bit Descriptions for MI_PHY_ID2

MMD Access Control Register

Address: 0xD, Reset: 0x0000, Name: MMD_ACCESS_CNTRL

This address corresponds to the MMD access control register specified in Clause 22.2.4.3.11 of IEEE Standard 802.3-2018.

Table 120. Bit Descriptions for MMD_ACCESS_CNTRL

MMD Access Register

Address: 0xE, Reset: 0x0000, Name: MMD_ACCESS

This address corresponds to the MMD access address data register specified in Clause 22.2.4.3.12 of IEEE Standard 802.3-2018.

The MMD_ADDR_DATA register is used with the MMD_ACCESS_CNTRL register to provide access to the MMD address space using the interface and mechanisms defined in Clause 22.2.4.

Table 121. Bit Descriptions for MMD_ACCESS

PHY CLAUSE 45 REGISTER DETAILS

Table 122. PHY Clause 45 Register Summary (Continued)

Table 122. PHY Clause 45 Register Summary (Continued)

Table 122. PHY Clause 45 Register Summary (Continued)

PMA/PMD Control 1 Register

Device Address: 0x01; Register Address: 0x0000, Reset: 0x0000, Name: PMA_PMD_CNTRL1

This address corresponds to the PMA/PMD Control Register 1 specified in Clause 45.2.1.1 of Standard 802.3. Note that the reset value of this register is dependent on the hardware configuration pin settings.

Table 123. Bit Descriptions for PMA_PMD_CNTRL1

PMA/PMD Status 1 Register

Device Address: 0x01; Register Address: 0x0001, Reset: 0x0002, Name: PMA_PMD_STAT1

This address corresponds to the PMA/PMD Status Register 1 specified in Clause 45.2.1.2 of Standard 802.3.

Table 124. Bit Descriptions for PMA_PMD_STAT1

PMA/PMD MMD Devices in Package 1 Register

Device Address: 0x01; Register Address: 0x0005, Reset: 0x008B, Name: PMA_PMD_DEVS_IN_PKG1

Table 125. Bit Descriptions for PMA_PMD_DEVS_IN_PKG1

PMA/PMD MMD Devices in Package 2 Register

Device Address: 0x01; Register Address: 0x0006, Reset: 0xC000, Name: PMA_PMD_DEVS_IN_PKG2

Table 126. Bit Descriptions for PMA_PMD_DEVS_IN_PKG2

PMA/PMD Control 2 Register

Device Address: 0x01; Register Address: 0x0007, Reset: 0x003D, Name: PMA_PMD_CNTRL2

Table 127. Bit Descriptions for PMA_PMD_CNTRL2

Table 127. Bit Descriptions for PMA_PMD_CNTRL2 (Continued)

Bits	Bit Name	Description	Reset	Access
		0100000: TS_40GBASE_KR4_PMA_PMD.		
		0100001: TS_40GBASE_CR4_PMA_PMD.		
		0100010: TS 40GBASE SR4 PMA PMD.		
		0100011: TS_40GBASE_LR4_PMA_PMD.		
		0100100: TS 40GBASE FR PMA PMD.		
		0100101: TS_40GBASE_ER4_PMA_PMD.		
		0100110: TS_40GBASE_T_PMA.		
		0101000: TS_100GBASE_CR10_PMA_PMD.		
		0101001: TS_100GBASE_SR10_PMA_PMD.		
		0101010: TS_100GBASE_LR4_PMA_PMD.		
		0101011: TS 100GBASE ER4 PMA PMD.		
		0101100: TS 100GBASE KP4 PMA PMD.		
		0101101: TS_100GBASE_KR4_PMA_PMD.		
		0101110: TS_100GBASE_CR4_PMA_PMD.		
		0101111: TS_100GBASE_SR4_PMA_PMD.		
		0110000: TS 2 5GBASE T PMA.		
		0110001: TS 5GBASE T PMA.		
		0110010: TS 10GPASS XR D PMA PMD.		
		0110011: TS 10GPASS XR U PMA PMD.		
		0110100: TS BASE H PMA PMD.		
		0110101: TS_25GBASE_LR_PMA_PMD.		
		0110110: TS_25GBASE_ER_PMA_PMD.		
		0110111: TS 25GBASE T PMA.		
		0111000: TS_25GBASE_CR_OR_25GBASE_CR_S_PMA_PMD.		
		0111001: TS_25GBASE_KR_OR_25GBASE_KR_S_PMA_PMD.		
		0111010: TS 25GBASE SR PMA PMD.		
		0111101: TS BASE T1 PMA PMD.		
		1010011: TS_200GBASE_DR4_PMA_PMD.		
		1010100: TS_200GBASE_FR4_PMA_PMD.		
		1010101: TS_200GBASE_LR4_PMA_PMD.		
		1011001: TS_400GBASE_SR16_PMA_PMD.		
		1011010: TS_400GBASE_DR4_PMA_PMD.		
		1011011: TS 400GBASE FR8 PMA PMD.		
		1011100: TS 400GBASE LR8 PMA PMD.		

PMA/PMD Status 2 Register

Device Address: 0x01; Register Address: 0x0008, Reset: 0x8301, Name: PMA_PMD_STAT2

PMA/PMD Transmit Disable Register

Device Address: 0x01; Register Address: 0x0009, Reset: 0x0000, Name: PMA_PMD_TX_DIS

This address corresponds to the PMD transmit disable register specified in Clause 45.2.1.8 of Standard 802.3.

Table 129. Bit Descriptions for PMA_PMD_TX_DIS

PMA/PMD Extended Abilities Register

Device Address: 0x01; Register Address: 0x000B, Reset: 0x0800, Name: PMA_PMD_EXT_ABILITY

PMA/PMD extended abilities.

Table 130. Bit Descriptions for PMA_PMD_EXT_ABILITY

BASE-T1 PMA/PMD Extended Ability Register

Device Address: 0x01; Register Address: 0x0012, Reset: 0x0004, Name: PMA_PMD_BT1_ABILITY

This address corresponds to the BASE-T1 PMA/PMD extended ability register specified in Clause 45.2.1.16 of Standard 802.3. This register is read only, and writes have no effect.

Table 131. Bit Descriptions for PMA_PMD_BT1_ABILITY

BASE-T1 PMA/PMD Control Register

Device Address: 0x01; Register Address: 0x0834, Reset: 0x8002, Name: PMA_PMD_BT1_CONTROL

This address corresponds to the BASE-T1 PMA/PMD control register specified in Clause 45.2.1.185 of Standard 802.3.

Table 132. Bit Descriptions for PMA_PMD_BT1_CONTROL

Table 132. Bit Descriptions for PMA_PMD_BT1_CONTROL (Continued)

10BASE-T1L PMA Control Register

Device Address: 0x01; Register Address: 0x08F6, Reset: 0x0000, Name: B10L_PMA_CNTRL

This address corresponds to the 10BASE-T1L PMA control register specified in Clause 45.2.1.186a of Standard 802.3cg.

Table 133. Bit Descriptions for B10L_PMA_CNTRL

10BASE-T1L PMA Status Register

Device Address: 0x01; Register Address: 0x08F7, Reset: 0x2800, Name: B10L_PMA_STAT

This address corresponds to the 10BASE-T1L PMA status register specified in Clause 45.2.1.186b of Standard 802.3cg.

Bits	Bit Name	Description	Reset	Access
[15:14]	RESERVED	Reserved.	0x0	R
13	B10L LB PMA LOC ABLE	10BASE-T1L PMA Loopback Ability. This bit always reads as 1 because the PMA has loopback ability	0x1	R
12	B10L TX LVL HI ABLE	10BASE-T1L High Voltage Tx Ability. Indicates that the PHY supports 10BASE-T1L high voltage (2.4 V p-p) transmit level operating mode.	Pin Dependent	R
11	B10L PMA SFT PD ABLE	PMA Supports Power-Down. Indicates that the PMA supports software power-down.	0x1	R
10	B10L EEE ABLE	10BASE-T1L EEE Ability. Indicates if the PHY supports 10BASE-T1L EEE.	0x0	R

Table 134. Bit Descriptions for B10L_PMA_STAT

Table 134. Bit Descriptions for B10L_PMA_STAT (Continued)

10BASE-T1L Test Mode Control Register

Device Address: 0x01; Register Address: 0x08F8, Reset: 0x0000, Name: B10L_TEST_MODE_CNTRL

This address corresponds to the 10BASE-T1L PMA test mode control register specified in Clause 45.2.1.186c of Standard 802.3cg. The default value of this register selects normal operation without management intervention as the initial state of the device.

Frequency Offset Saturation Threshold for CR Stability Check Register

Device Address: 0x01; Register Address: 0x8015, Reset: 0x0008, Name: CR_STBL_CHK_FOFFS_SAT_THR

Table 136. Bit Descriptions for CR_STBL_CHK_FOFFS_SAT_THR

Slave IIR Filter Change Echo Acquisition Clock Recovery Proportional Gain Register

Device Address: 0x01; Register Address: 0x81E7, Reset: 0x0400, Name: SLV_FLTR_ECHO_ACQ_CR_KP

Table 137. Bit Descriptions for SLV_FLTR_ECHO_ACQ_CR_KP

10BASE-T1L PMA Link Status Register

Device Address: 0x01; Register Address: 0x8302, Reset: 0x0000, Name: B10L_PMA_LINK_STAT

This address can be read to determine the 10BASE-T1L PMA link status. Reading B10L PMA_LINK_STAT clears the latching condition of these bits.

Table 138. Bit Descriptions for B10L_PMA_LINK_STAT

Table 138. Bit Descriptions for B10L_PMA_LINK_STAT (Continued)

Bits	Bit Name	Description	Reset	Access
9	B10L REM RCVR STAT OK LL	10BASE-T1L Remote Receiver Status OK Latch Low, Latched low version of B10L_REM_RCVR_STAT_OK.	0x0	RLL
8	B10L REM RCVR STAT OK	10BASE-T1L Remote Receiver Status OK. When read as 1, this bit indicates that the remote receiver status is OK.	0x0	R
$\overline{7}$	B10L LOC RCVR STAT OK LL	10BASE-T1L Local Receiver Status OK Latch Low. Latched low version of B10L LOC RCVR STAT OK.	0x0	R LL
6	B10L LOC RCVR STAT OK	10BASE-T1L Local Receiver Status OK. When read as 1, this bit indicates that the local receiver status is OK.	0x0	R
5	B10L DSCR STAT OK LL	BASE-T1L Descrambler Status OK Latch Low. When read as 1, this bit indicates that the descrambler status is OK.	0x0	R LL
4	B10L DSCR STAT OK	10BASE-T1L Descrambler Status OK. When read as 1, this bit indicates that the descrambler status is OK.	0x0	R
$[3:2]$	RESERVED	Reserved.	0x0	R
	B10L LINK STAT OK LL	Link Status OK Latch Low. When read as 1, this bit indicates that the link status is OK.	0x0	R LL
0	B10L LINK STAT OK	Link Status OK. When read as 1, this bit indicates that the link status is OK.	0x0	R

MSE Value Register

Device Address: 0x01; Register Address: 0x830B, Reset: 0x0000, Name: MSE_VAL

Table 139. Bit Descriptions for MSE_VAL

PCS Control 1 Register

Device Address: 0x03; Register Address: 0x0000, Reset: 0x0000, Name: PCS_CNTRL1

This address corresponds to the PCS Control Register 1 specified in Clause 45.2.3.1 of Standard 802.3.

Table 140. Bit Descriptions for PCS_CNTRL1

PCS Status 1 Register

Device Address: 0x03; Register Address: 0x0001, Reset: 0x0002, Name: PCS_STAT1

Table 141. Bit Descriptions for PCS_STAT1

PCS MMD Devices in Package 1 Register

Device Address: 0x03; Register Address: 0x0005, Reset: 0x008B, Name: PCS_DEVS_IN_PKG1

Table 142. Bit Descriptions for PCS_DEVS_IN_PKG1

PCS MMD Devices in Package 2 Register

Device Address: 0x03; Register Address: 0x0006, Reset: 0xC000, Name: PCS_DEVS_IN_PKG2

Vendor Specific Device 1 and Vendor Specific Device 2 MMDs are present.

Table 143. Bit Descriptions for PCS_DEVS_IN_PKG2

PCS Status 2 Register

Device Address: 0x03; Register Address: 0x0008, Reset: 0x8000, Name: PCS_STAT2

Table 144. Bit Descriptions for PCS_STAT2

10BASE-T1L PCS Control Register

Device Address: 0x03; Register Address: 0x08E6, Reset: 0x0000, Name: B10L_PCS_CNTRL

This address corresponds to the 10BASE-T1L PCS control register specified in Clause 45.2.3.68a of Standard 802.3cg.

Table 145. Bit Descriptions for B10L_PCS_CNTRL

10BASE-T1L PCS Status Register

Device Address: 0x03; Register Address: 0x08E7, Reset: 0x0000, Name: B10L_PCS_STAT

This address corresponds to the 10BASE-T1L PCS status register specified in Clause 45.2.3.68b of Standard 802.3cg.

Table 146. Bit Descriptions for B10L_PCS_STAT

Autonegotiation MMD Devices in Package 1 Register

Device Address: 0x07; Register Address: 0x0005, Reset: 0x008B, Name: AN_DEVS_IN_PKG1

Clause 22 registers and PMA/PMD, PCS, and autonegotiation MMDs are present.

Table 147. Bit Descriptions for AN_DEVS_IN_PKG1

Autonegotiation MMD Devices in Package 2 Register

Device Address: 0x07; Register Address: 0x0006, Reset: 0xC000, Name: AN_DEVS_IN_PKG2

Vendor Specific Device 1 and Vendor Specific Device 2 MMDs are present.

Table 148. Bit Descriptions for AN_DEVS_IN_PKG2

BASE-T1 Autonegotiation Control Register

Device Address: 0x07; Register Address: 0x0200, Reset: 0x1000, Name: AN_CONTROL

This address corresponds to the BASE-T1 autonegotiation control register specified in Clause 45.2.7.19 of Standard 802.3.

Table 149. Bit Descriptions for AN_CONTROL

BASE-T1 Autonegotiation Status Register

Device Address: 0x07; Register Address: 0x0201, Reset: 0x0008, Name: AN_STATUS

This address corresponds to the BASE-T1 autonegotiation status register specified in Clause 45.2.7.20 of Standard 802.3.

Table 150. Bit Descriptions for AN_STATUS

Table 150. Bit Descriptions for AN_STATUS (Continued)

BASE-T1 Autonegotiation Advertisement Register, Bits[15:0]

Device Address: 0x07; Register Address: 0x0202, Reset: 0x0001, Name: AN_ADV_ABILITY_L

This address corresponds to the BASE-T1 autonegotiation advertisement register, Bits[15:0] specified in Clause 45.2.7.21 of Standard 802.3.

Table 151. Bit Descriptions for AN_ADV_ABILITY_L

BASE-T1 Autonegotiation Advertisement Register, Bits[31:16]

Device Address: 0x07; Register Address: 0x0203, Reset: 0x4000, Name: AN_ADV_ABILITY_M

This address corresponds to the BASE-T1 autonegotiation advertisement register, Bits[31:16], specified in Clause 45.2.7.21 of Standard 802.3.

Table 152. Bit Descriptions for AN_ADV_ABILITY_M

BASE-T1 Autonegotiation Advertisement Register, Bits[47:32]

Device Address: 0x07; Register Address: 0x0204, Reset: 0x0000, Name: AN_ADV_ABILITY_H

This address corresponds to the BASE-T1 autonegotiation advertisement register, Bits[47:32], specified in Clause 45.2.7.21 of Standard 802.3.

Table 153. Bit Descriptions for AN_ADV_ABILITY_H

BASE-T1 Autonegotiation Link Partner Base Page Ability Register, Bits[15:0]

Device Address: 0x07; Register Address: 0x0205, Reset: 0x0000, Name: AN_LP_ADV_ABILITY_L

This address corresponds to the link partner's BASE-T1 autonegotiation base page ability register, Bits[15:0], specified in Clause 45.2.7.22 of Standard 802.3. Note that the value of the AN_LP_ADV_ABILITY_M and AN_LP_ADV_ABILITY_H registers is latched when AN_LP_ADV_ABILITY_L is read.

Table 154. Bit Descriptions for AN_LP_ADV_ABILITY_L

BASE-T1 Autonegotiation Link Partner Base Page Ability Register, Bits[31:16]

Device Address: 0x07; Register Address: 0x0206, Reset: 0x0000, Name: AN_LP_ADV_ABILITY_M

This address corresponds to the link partner's BASE-T1 autonegotiation base page ability register, Bits[31:16], specified in Clause 45.2.7.22 of Standard 802.3. Note that the value of this register is latched when the AN_LP_ADV_ABILITY_L register is read. Reading this register returns the latched value rather than the current value.

Table 155. Bit Descriptions for AN_LP_ADV_ABILITY_M

BASE-T1 Autonegotiation Link Partner Base Page Ability Register, Bits[47:32]

Device Address: 0x07; Register Address: 0x0207, Reset: 0x0000, Name: AN_LP_ADV_ABILITY_H

This address corresponds to the link partner's BASE-T1 autonegotiation base page ability register, Bits[47:32], specified in Clause 45.2.7.22 of Standard 802.3. Note that the value of this register is latched when the AN_LP_ADV_ABILITY_L register is read. Reading this register returns the latched value rather than the current value.

BASE-T1 Autonegotiation Next Page Transmit Register, Bits[15:0]

Device Address: 0x07; Register Address: 0x0208, Reset: 0x2001, Name: AN_NEXT_PAGE_L

This address corresponds to the BASE-T1 autonegotiation next page transmit register, Bits[15:0], specified in Clause 45.2.7.23 of Standard 802.3. On power-up or autonegotiation reset, this register contains the default value, which represents a message page with the message code set to null. Write AN_NEXT_PAGE_M and AN_NEXT_PAGE_H before AN_NEXT_PAGE_L.

Table 157. Bit Descriptions for AN_NEXT_PAGE_L

BASE-T1 Autonegotiation Next Page Transmit Register, Bits[31:16]

Device Address: 0x07; Register Address: 0x0209, Reset: 0x0000, Name: AN_NEXT_PAGE_M

This address corresponds to the BASE-T1 autonegotiation next page transmit register, Bits[31:16], specified in Clause 45.2.7.23 of Standard 802.3. On power-up or autonegotiation reset, this register contains the default value, which represents a message page with the message code set to null. Write AN_NEXT_PAGE_M and AN_NEXT_PAGE_H before AN_NEXT_PAGE_L.

Table 158. Bit Descriptions for AN_NEXT_PAGE_M

BASE-T1 Autonegotiation Next Page Transmit Register, Bits[47:32]

Device Address: 0x07; Register Address: 0x020A, Reset: 0x0000, Name: AN_NEXT_PAGE_H

This address corresponds to the BASE-T1 autonegotiation next page transmit register, Bits[47:42], specified in Clause 45.2.7.23 of Standard 802.3. On power-up or autonegotiation reset, this register contains the default value, which represents a message page with the message code set to null. Write AN_NEXT_PAGE_M and AN_NEXT_PAGE_H before AN_NEXT_PAGE_L.

Table 159. Bit Descriptions for AN_NEXT_PAGE_H

BASE-T1 Autonegotiation Link Partner Next Page Ability Register, Bits[15:0]

Device Address: 0x07; Register Address: 0x020B, Reset: 0x0000, Name: AN_LP_NEXT_PAGE_L

This address corresponds to the BASE-T1 autonegotiation link partner's next page ability register, Bits[15:0], specified in Clause 45.2.7.24 of Standard 802.3. The values of AN_LP_NEXT_PAGE_M and AN_LP_NEXT_PAGE_H are latched when this register is read.

Table 160. Bit Descriptions for AN_LP_NEXT_PAGE_L

BASE-T1 Autonegotiation Link Partner Next Page Ability Register, Bits[31:16]

Device Address: 0x07; Register Address: 0x020C, Reset: 0x0000, Name: AN_LP_NEXT_PAGE_M

This address corresponds to the BASE-T1 autonegotiation next page ability register, Bits[31:16], of the link partner specified in Clause 45.2.7.24 of Standard 802.3. The values of this register are latched when AN_LP_NEXT_PAGE_L is read. Reading this register returns the latched value rather than the current value.

Table 161. Bit Descriptions for AN_LP_NEXT_PAGE_M

BASE-T1 Autonegotiation Link Partner Next Page Ability Register, Bits[47:32]

Device Address: 0x07; Register Address: 0x020D, Reset: 0x0000, Name: AN_LP_NEXT_PAGE_H

This address corresponds to the BASE-T1 autonegotiation link partner's next page ability register, Bits[47:32], specified in Clause 45.2.7.24 of Standard 802.3. The values of this register are latched when AN_LP_NEXT_PAGE_L is read. Reading this register returns the latched value rather than the current value.

Table 162. Bit Descriptions for AN_LP_NEXT_PAGE_H

10BASE-T1 Autonegotiation Control Register

Device Address: 0x07; Register Address: 0x020E, Reset: 0x8000, Name: AN_B10_ADV_ABILITY

This address corresponds to the 10BASE-T1 autonegotiation control register specified in Clause 45.2.7.25 of Standard 802.3cg.

Table 163. Bit Descriptions for AN_B10_ADV_ABILITY

Table 163. Bit Descriptions for AN_B10_ADV_ABILITY (Continued)

10BASE-T1 Autonegotiation Status Register

Device Address: 0x07; Register Address: 0x020F, Reset: 0x0000, Name: AN_B10_LP_ADV_ABILITY

This address corresponds to the 10BASE-T1 autonegotiation status register specified in Clause 45.2.7.26 of Standard 802.3cg.

Table 164. Bit Descriptions for AN_B10_LP_ADV_ABILITY

Autonegotiation Force Mode Enable Register

Device Address: 0x07; Register Address: 0x8000, Reset: 0x0000, Name: AN_FRC_MODE_EN

Note that the effect of this register is superseded by the AN_EN bit, which enables the autonegotiation process. If autonegotiation is disabled (AN_EN = 0) and AN_FRC_MODE_EN is 1, forced mode is enabled.

Table 165. Bit Descriptions for AN_FRC_MODE_EN

Extra Autonegotiation Status Register

Device Address: 0x07; Register Address: 0x8001, Reset: 0x0000, Name: AN_STATUS_EXTRA

This register is provided in addition to AN_STATUS.

Table 166. Bit Descriptions for AN_STATUS_EXTRA

PHY Instantaneous Status Register

Device Address: 0x07; Register Address: 0x8030, Reset: 0x0010, Name: AN_PHY_INST_STATUS

This register address provides access to instantaneous status indications. These values are not latched, and the set of indications returned here are a consistent set, that is, a set of values in effect at the time the register address is read.

Vendor Specific MMD 1 Device Identifier High Register

Device Address: 0x1E; Register Address: 0x0002, Reset: 0x0283, Name: MMD1_DEV_ID1

This address corresponds to the Vendor Specific MMD 1 device identifier register specified in Clause 45.2.11.1 of Standard 802.3 and allows 16 bits of the organizationally unique identifier (OUI) to be observed.

Table 168. Bit Descriptions for MMD1_DEV_ID1

Vendor Specific MMD 1 Device Identifier Low Register

Device Address: 0x1E; Register Address: 0x0003, Reset: 0xBC91, Name: MMD1_DEV_ID2

This address corresponds to the Vendor Specific MMD 1 device identifier register specified in Clause 45.2.11.1 of Standard 802.3 and allows six bits of the OUI along with the model number and revision number to be observed.

Table 169. Bit Descriptions for MMD1_DEV_ID2

Vendor Specific 1 MMD Devices in Package Register

Device Address: 0x1E; Register Address: 0x0005, Reset: 0x008B, Name: MMD1_DEVS_IN_PKG1

Clause 22 registers and PMA/PMD, PCS, and autonegotiation MMDs are present.

Table 170. Bit Descriptions for MMD1_DEVS_IN_PKG1

Device Address: 0x1E; Register Address: 0x0006, Reset: 0xC000, Name: MMD1_DEVS_IN_PKG2

Vendor-specific device 1 and Vendor-specific device 2 MMDs present

Table 171. Bit Descriptions for MMD1_DEVS_IN_PKG2

Vendor Specific MMD 1 Status Register

Device Address: 0x1E; Register Address: 0x0008, Reset: 0x8000, Name: MMD1_STATUS

This address corresponds to the Vendor Specific MMD 1 status register specified in Clause 45.2.11.2 of Standard 802.3.

Table 172. Bit Descriptions for MMD1_STATUS

System Interrupt Status Register

Device Address: 0x1E; Register Address: 0x0010, Reset: 0x1000, Name: CRSM_IRQ_STATUS

This address can be used to check which interrupt requests have triggered since the last time it was read. Each bit goes high when the associated event occurs and then latches high until it is unlatched by reading. The bits of CRSM_IRQ_STATUS go high even when the associated interrupts are not enabled. A reserved interrupt being triggered indicates a fatal error in the system.

Table 173. Bit Descriptions for CRSM_IRQ_STATUS

System Interrupt Mask Register

Device Address: 0x1E; Register Address: 0x0020, Reset: 0x1FFE, Name: CRSM_IRQ_MASK

Controls whether or not the interrupt signal is asserted in response to various events.

Table 174. Bit Descriptions for CRSM_IRQ_MASK

Software Reset Register

Device Address: 0x1E; Register Address: 0x8810, Reset: 0x0000, Name: CRSM_SFT_RST

Table 175. Bit Descriptions for CRSM_SFT_RST

Software Power-Down Control Register

Device Address: 0x1E; Register Address: 0x8812, Reset: 0x0000, Name: CRSM_SFT_PD_CNTRL

Table 176. Bit Descriptions for CRSM_SFT_PD_CNTRL

PHY Subsystem Reset Register

Device Address: 0x1E; Register Address: 0x8814, Reset: 0x0000, Name: CRSM_PHY_SUBSYS_RST

Table 177. Bit Descriptions for CRSM_PHY_SUBSYS_RST

PHY MAC Interface Reset Register

Device Address: 0x1E; Register Address: 0x8815, Reset: 0x0000, Name: CRSM_MAC_IF_RST

Table 178. Bit Descriptions for CRSM_MAC_IF_RST

System Status Register

Device Address: 0x1E; Register Address: 0x8818, Reset: 0x0000, Name: CRSM_STAT

Table 179. Bit Descriptions for CRSM_STAT

CRSM Power Management Control Register

Device Address: 0x1E; Register Address: 0x8819, Reset: 0x0000, Name: CRSM_PMG_CNTRL

Table 180. Bit Descriptions for CRSM_PMG_CNTRL

CRSM Diagnostics Clock Control Register

Device Address: 0x1E; Register Address: 0x882C, Reset: 0x0002, Name: CRSM_DIAG_CLK_CTRL

CRSM diagnostics clock control.

Table 181. Bit Descriptions for CRSM_DIAG_CLK_CTRL

Package Configuration Values Register

Device Address: 0x1E; Register Address: 0x8C22, Reset: 0x0000, Name: MGMT_PRT_PKG

The MGMT_CFG_VAL address allows reading of the package configuration values.

Table 182. Bit Descriptions for MGMT_PRT_PKG

MDIO Control Register

Device Address: 0x1E; Register Address: 0x8C30, Reset: 0x0000, Name: MGMT_MDIO_CNTRL

Table 183. Bit Descriptions for MGMT_MDIO_CNTRL

Pin Mux Configuration 1 Register

Device Address: 0x1E; Register Address: 0x8C56, Reset: 0x00FE, Name: DIGIO_PINMUX

Table 184. Bit Descriptions for DIGIO_PINMUX

Table 184. Bit Descriptions for DIGIO_PINMUX (Continued)

Bits	Bit Name	Description	Reset	Access
		0: assert high.		
		1: assert low.		

LED_0 On/Off Blink Time Register

Device Address: 0x1E; Register Address: 0x8C80, Reset: 0x3636, Name: LED0_BLINK_TIME_CNTRL

LED on blink time = LED0 ON N4MS \times 4 ms.

LED off blink time = LED0 OFF $N4MS \times 4$ ms.

If LEDx MODE = 0 and LEDx FUNCTION is set to blink, the LED activity starts with an LED off sequence, followed by an LED on sequence, and then repeats.

If LEDx MODE = 1 and LEDx FUNCTION is set to blink, the LED activity starts with an LED on sequence, followed by an LED off sequence, and then repeats.

If LEDx OFF_N4MS = LEDx_ON_N4MS = 0, this is a special case whereby the internal activity signal as selected by LEDx_FUNCTION can be monitored live.

If LEDx FUNCTION is programmed to a combination of a link and activity signal, the LED is on while the link is up and with no activity. The LED switches off for either loss of link or receipt of activity.

If LED_X FUNCTION is programmed to an activity signal, the LED is off with no activity. The LED switches on upon receipt of activity.

Table 185. Bit Descriptions for LED0_BLINK_TIME_CNTRL

LED 1 On/Off Blink Time Register

Device Address: 0x1E; Register Address: 0x8C81, Reset: 0x3636, Name: LED1_BLINK_TIME_CNTRL

LED on blink time = LED1 ON N4MS \times 4 ms.

LED off blink time = LED1 OFF N4MS \times 4 ms.

If LEDx MODE = 0 and LEDx FUNCTION is set to blink, the LED activity starts with an LED off sequence followed by an LED on sequence, and then repeats.

If LEDx MODE = 1 and LEDx FUNCTION is set to blink, the LED activity starts with an LED on sequence, followed by an LED Off sequence, and then repeats.

If LEDx $OFN4MS = LEDx ON N4MS = 0$, this is a special case whereby the internal activity signal as selected by LEDx $FUNCTION can be$ monitored live.

If LEDx FUNCTION is programmed to a combination of a link and activity signal, the LED is on while the link is up and with no activity. The LED switches off for either loss of link or receipt of activity.

If LED_X FUNCTION is programmed to an activity signal, the LED is off with no activity. The LED switches on upon receipt of activity.

Table 186. Bit Descriptions for LED1_BLINK_TIME_CNTRL

LED Control Register

Device Address: 0x1E; Register Address: 0x8C82, Reset: 0x8480, Name: LED_CNTRL

LED control register.

Table 187. Bit Descriptions for LED_CNTRL

Table 187. Bit Descriptions for LED_CNTRL (Continued)

Table 187. Bit Descriptions for LED_CNTRL (Continued)

LED Polarity Register

Device Address: 0x1E; Register Address: 0x8C83, Reset: 0x0000, Name: LED_POLARITY

Allows the LED polarity to be automatically sensed by the internal logic or allows reconfiguration by the user.

Table 188. Bit Descriptions for LED_POLARITY

Vendor Specific MMD 2 Device Identifier High Register

Device Address: 0x1F; Register Address: 0x0002, Reset: 0x0283, Name: MMD2_DEV_ID1

Table 189. Bit Descriptions for MMD2_DEV_ID1

Vendor Specific MMD 2 Device Identifier Low Register

Device Address: 0x1F; Register Address: 0x0003, Reset: 0xBC91, Name: MMD2_DEV_ID2

Table 190. Bit Descriptions for MMD2_DEV_ID2

Vendor Specific 2 MMD Devices in Package Register

Device Address: 0x1F; Register Address: 0x0005, Reset: 0x008B, Name: MMD2_DEVS_IN_PKG1

Clause 22 registers and PMA/PMD, PCS, and autonegotiation MMDs are present.

Table 191. Bit Descriptions for MMD2_DEVS_IN_PKG1

Device Address: 0x1F; Register Address: 0x0006, Reset: 0xC000, Name: MMD2_DEVS_IN_PKG2

Vendor Specific Device 1 and Vendor Specific Device 2 MMDs are present.

Table 192. Bit Descriptions for MMD2_DEVS_IN_PKG2

Vendor Specific MMD 2 Status Register

Device Address: 0x1F; Register Address: 0x0008, Reset: 0x8000, Name: MMD2_STATUS

This address corresponds to the Vendor Specific MMD 2 status register.

Table 193. Bit Descriptions for MMD2_STATUS

PHY Subsystem Interrupt Status Register

Device Address: 0x1F; Register Address: 0x0011, Reset: 0x0000, Name: PHY_SUBSYS_IRQ_STATUS

This address can be read to check which interrupt events have occurred since the last time it was read. Each bit goes high when the associated event occurs and then latches high until it is unlatched by reading. The bits of PHY_SUBSYS_IRQ_STATUS go high even when the associated bits in PHY_SUBSYS_IRQ_MASK are not set. A reserved interrupt being triggered indicates a fatal error in the system.

Table 194. Bit Descriptions for PHY_SUBSYS_IRQ_STATUS (Continued)

PHY Subsystem Interrupt Mask Register

Device Address: 0x1F; Register Address: 0x0021, Reset: 0x2402, Name: PHY_SUBSYS_IRQ_MASK

Controls whether or not the interrupt signal is asserted in response to various events.

Table 195. Bit Descriptions for PHY_SUBSYS_IRQ_MASK

Frame Checker Enable Register

Device Address: 0x1F; Register Address: 0x8001, Reset: 0x0001, Name: FC_EN

This register is used to enable the frame checker. The frame checker analyzes the received frames from either the MAC interface or the PHY (see the FC_TX_SEL register) to report the number of frames received, CRC errors, and various other frame errors. The frame checker frame and error counter registers count these events.

Table 196. Bit Descriptions for FC_EN

Frame Checker Interrupt Enable Register

Device Address: 0x1F; Register Address: 0x8004, Reset: 0x0001, Name: FC_IRQ_EN

This register is used to enable the frame checker interrupt. An interrupt is generated when a receive error occurs. Enable the frame checker/generator interrupt in the PHY_SUBSYS_IRQ_MASK register. Set the MAC_IF_FC_FG_IRQ_EN bit.

The status can be read via the MAC_IF_FC_FG_IRQ_LH bit in the PHY_SUBSYS_IRQ_STATUS register.

Frame Checker Transmit Select Register

Device Address: 0x1F; Register Address: 0x8005, Reset: 0x0000, Name: FC_TX_SEL

This register is used to select the transmit side or receive side for frames to be checked. If set, frames received from the MAC interface to be transmitted are checked. The frame checker can be used to verify that correct data is received over the MAC interface and is also useful if remote loopback is enabled (see the MAC_IF_REM_LB_EN bit in the MAC_IF_LOOPBACK register) because it can be used to check the received data after it is looped back at the MAC interface.

Table 198. Bit Descriptions for FC_TX_SEL

Receive Error Count Register

Device Address: 0x1F; Register Address: 0x8008, Reset: 0x0000, Name: RX_ERR_CNT

The receive error counter register is used to access the receive error counter associated with the frame checker in the PHY.

Table 199. Bit Descriptions for RX_ERR_CNT

Frame Checker Count High Register

Device Address: 0x1F; Register Address: 0x8009, Reset: 0x0000, Name: FC_FRM_CNT_H

This register is a latched copy of Bits[31:16] of the 32-bit of the receive frame counter register. When the receive error counter (RX_ERR_CNT) is read, the receive frame counter register is latched so that the error count and the receive frame count are synchronized.

Table 200. Bit Descriptions for FC_FRM_CNT_H

Frame Checker Count Low Register

Device Address: 0x1F; Register Address: 0x800A, Reset: 0x0000, Name: FC_FRM_CNT_L

This register is a latched copy of Bits[15:0] of the 32-bit receive frame counter register. When the receive error counter (RX_ERR_CNT) is read, the receive frame counter register is latched so that the error count and receive frame count are synchronized.

Table 201. Bit Descriptions for FC_FRM_CNT_L

Frame Checker Length Error Count Register

Device Address: 0x1F; Register Address: 0x800B, Reset: 0x0000, Name: FC_LEN_ERR_CNT

This register is a latched copy of the frame length error counter register. This register is a count of received frames with a length error status. When the receive error counter (RX_ERR_CNT) is read, the frame length error counter register is latched, which ensures that the frame length error count and receive frame count are synchronized.

Table 202. Bit Descriptions for FC_LEN_ERR_CNT

Frame Checker Alignment Error Count Register

Device Address: 0x1F; Register Address: 0x800C, Reset: 0x0000, Name: FC_ALGN_ERR_CNT

This register is a latched copy of the frame alignment error counter register. This register is a count of received frames with an alignment error status. When the receive error counter (RX_ERR_CNT) is read, the alignment error counter is latched, which ensures that the frame alignment error count and the receive frame count are synchronized.

Table 203. Bit Descriptions for FC_ALGN_ERR_CNT

Frame Checker Symbol Error Count Register

Device Address: 0x1F; Register Address: 0x800D, Reset: 0x0000, Name: FC_SYMB_ERR_CNT

This register is a latched copy of the symbol error counter register. This register is a count of received frames with both RX_ER and RX_DV set. When the receive error counter (RX_ERR_CNT) is read, the symbol error count is latched, which ensures that the symbol error count and the frame receive count are synchronized.

Table 204. Bit Descriptions for FC_SYMB_ERR_CNT

Frame Checker Oversized Frame Count Register

Device Address: 0x1F; Register Address: 0x800E, Reset: 0x0000, Name: FC_OSZ_CNT

This register is a latched copy of the oversized frame error counter register. This register is a count of receiver frames with a length greater than specified in frame checker maximum frame size (FC_MAX_FRM_SIZE). When the receive error counter (RX_ERR_CNT) is read, the oversized frame counter register is latched, which ensures that the oversized error count and the receive frame count are synchronized.

Table 205. Bit Descriptions for FC_OSZ_CNT

Frame Checker Undersized Frame Count Register

Device Address: 0x1F; Register Address: 0x800F, Reset: 0x0000, Name: FC_USZ_CNT

This register is a latched copy of the undersized frame error counter register. This register is a count of received frames with less than 64 bytes. When the receive error counter (RX_ERR_CNT) is read, the undersized frame error counter is latched, which ensures that the undersized frame error count and the receive frame count are synchronized.

Table 206. Bit Descriptions for FC_USZ_CNT

Frame Checker Odd Nibble Frame Count Register

Device Address: 0x1F; Register Address: 0x8010, Reset: 0x0000, Name: FC_ODD_CNT

This register is a latched copy of the odd nibble frame register. This register is a count of received frames with an odd number of nibbles in the frame. When the receive error counter (RX_ERR_CNT) is read, the odd nibble frame counter register is latched, which ensures that the odd nibble frame count and the receive frame count are synchronized.

Table 207. Bit Descriptions for FC_ODD_CNT

Frame Checker Odd Preamble Packet Count Register

Device Address: 0x1F; Register Address: 0x8011, Reset: 0x0000, Name: FC_ODD_PRE_CNT

This register is a latched copy of the odd preamble packet counter register. This register is a count of received packets with an odd number of nibbles in the preamble. When the receive error counter (RX_ERR_CNT) is read, the odd preamble packet counter register is latched, which ensures that the odd preamble packet count and the receive frame count are synchronized.

Table 208. Bit Descriptions for FC_ODD_PRE_CNT

Frame Checker False Carrier Count Register

Device Address: 0x1F; Register Address: 0x8013, Reset: 0x0000, Name: FC_FALSE_CARRIER_CNT

This register is a latched copy of the false carrier events counter register. This is a count of the number of times the bad SSD state is entered. When the receive error counter (RX_ERR_CNT) is read, the false carrier events counter register is latched, which ensures that the false carrier events count and the receive frame count are synchronized.

Table 209. Bit Descriptions for FC_FALSE_CARRIER_CNT

Frame Generator Enable Register

Device Address: 0x1F; Register Address: 0x8020, Reset: 0x0000, Name: FG_EN

This register is used to enable the frame generator. When the frame generator is enabled, the source of data for the PHY comes from the frame generator and not the MAC interface. To use the frame generator, the diagnostic clock must also be enabled (DIAG_CLK_EN).

Table 210. Bit Descriptions for FG_EN

Frame Generator Control/Restart Register

Device Address: 0x1F; Register Address: 0x8021, Reset: 0x0001, Name: FG_CNTRL_RSTRT

This register controls the frame generator. The FG_CNTRL bit field specifies data field type used by the frame generator, for example, random all zeros. The FG_RSTRT bit restarts the frame generator.

Table 211. Bit Descriptions for FG_CNTRL_RSTRT

Frame Generator Continuous Mode Enable Register

Device Address: 0x1F; Register Address: 0x8022, Reset: 0x0000, Name: FG_CONT_MODE_EN

This register is used to put the frame generator into continuous mode. The default mode of operation is burst mode, where the number of frames generated is specified by the FG_NFRM_H and FG_NFRM_L registers.

Table 212. Bit Descriptions for FG_CONT_MODE_EN

Frame Generator Interrupt Enable Register

Device Address: 0x1F; Register Address: 0x8023, Reset: 0x0000, Name: FG_IRQ_EN

This register is used to enable the frame generator interrupt. An interrupt is generated when the requested number of frames is generated. Enable the frame checker/generator interrupt in the PHY_SUBSYS_IRQ_MASK register. Set the MAC_IF_FC_FG_IRQ_EN bit.

The interrupt status can be read via the MAC_IF_FC_FG_IRQ_LH bit in the PHY_SUBSYS_IRQ_STATUS register.

Table 213. Bit Descriptions for FG_IRQ_EN

Frame Generator Frame Length Register

Device Address: 0x1F; Register Address: 0x8025, Reset: 0x006B, Name: FG_FRM_LEN

This register specifies the data field frame length in bytes. In addition to the data field, six bytes are added for the source address, six bytes for the destination address, two bytes for the length field, and four bytes for the frame check sequence (FCS). The total length is the data field length plus 18.

Table 214. Bit Descriptions for FG_FRM_LEN

Frame Generator Interframe Gap Length Register

Device Address: 0x1F; Register Address: 0x8026, Reset: 0x000C, Name: FG_IFG_LEN

This register specifies the length in bytes of the interframe gap to be inserted between frames by the frame generator.

Table 215. Bit Descriptions for FG_IFG_LEN

Frame Generator Number of Frames High Register

Device Address: 0x1F; Register Address: 0x8027, Reset: 0x0000, Name: FG_NFRM_H

This register is Bits[31:16] of a 32-bit register that specifies the number of frames to be generated each time the frame generator is enabled or restarted.

Table 216. Bit Descriptions for FG_NFRM_H

Frame Generator Number of Frames Low Register

Device Address: 0x1F; Register Address: 0x8028, Reset: 0x0100, Name: FG_NFRM_L

This register is Bits[15:0] of a 32-bit register that specifies the number of frames to be generated each time the frame generator is enabled or restarted.

Table 217. Bit Descriptions for FG_NFRM_L

Frame Generator Done Register

Device Address: 0x1F; Register Address: 0x8029, Reset: 0x0000, Name: FG_DONE

This register is used to indicate that the frame generator has completed the generation of the number of frames requested in the FG_NFRM_H and FG_NFRM_L registers.

Table 218. Bit Descriptions for FG_DONE

Table 218. Bit Descriptions for FG_DONE (Continued)

MAC Interface Loopbacks Configuration Register

Device Address: 0x1F; Register Address: 0x8055, Reset: 0x000A, Name: MAC_IF_LOOPBACK

MAC interface loopbacks configuration.

Table 219. Bit Descriptions for MAC_IF_LOOPBACK

MAC Start of Packet (SOP) Generation Control Register

Device Address: 0x1F; Register Address: 0x805A, Reset: 0x001B, Name: MAC_IF_SOP_CNTRL

Table 220. Bit Descriptions for MAC_IF_SOP_CNTRL

PCB LAYOUT RECOMMENDATIONS

This section details the key areas of interest for the placement and layout of the PHY and corresponding support components.

PACKAGE LAYOUT

The LFCSP has an exposed pad underneath the package that must be soldered to the PCB ground for mechanical, electrical, and thermal reasons. For thermal impedance performance and to maximize heat removal, the use of a 4 × 4 array of thermal vias beneath the exposed ground pad is recommended. The PCB land pattern must incorporate the exposed ground pad with these vias in the footprint. The EVAL-ADIN1110EBZ uses an array of 4 × 4 filled vias on a 1.00 mm grid arrangement. The via pad diameter dimension is 0.02 inches (0.5015 mm).

COMPONENT PLACEMENT AND ROUTING

Prioritization of the critical traces and components helps simplify the routing exercise. Place and orient the critical traces and components first to ensure an effective layout. The critical components are the crystal and load capacitors, the CEXT_2 and CEXT_3 capacitors, and all bypass capacitors local to the ADIN1110 device. Prioritize these components for placement and routing.

Follow these recommendations:

- ► Place the decoupling capacitors as close as possible to their respective input pin.
- ► Minimize trace turns and use 45° corners.
- ► Avoid traces crossing power planes on adjacent layers.
- ► Avoid stubs.
- ► Avoid vias on high speed signals. If vias are required, place ground vias next to the signal vias to improve the return current path.

CRYSTAL PLACEMENT AND ROUTING

Particular attention is required on the crystal placement and routing to ensure minimum current consumption, reduce stray capacitance, and improve noise immunity.

Follow these recommendations:

- ► Place the crystal and its capacitors as close as possible to the XTAL I and XTAL O pins.
- ► Place the load capacitors close to each other.
- ► Use a local GND plane (copper island) for the crystal and load capacitor with a single point connection to the main GND.
- ► Reduce parasitic capacitance by keeping the XTAL I and XTAL_O traces away from each other.
- ► Adding a copper keep out on the layer beneath the crystal can also reduce the parasitic capacitance.

PCB STACK

Follow these recommendations:

- \triangleright Use a PCB stack with a minimum of 4 layers.
	- ► Consider 6 layers or more with external layers used as ground planes to improve EMI issues (optional).
- ► Define copper layer thickness based on the application and power requirements.
- ► Use internal layers for power and ground planes.
- ► Use external layers for signals.
- ► Use via stitching to improve ground and reduce EMI. Stitching pattern and via to via gaps to be defined based on the application.