RENESAS

ISL15100

Single Port, PLC Differential Line Driver

FN8577

Rev 0.00 September 19, 2013

DATASHEET

The ISL15100 is a single port differential line driver developed for Power Line Communication (PLC) applications. The device is designed to drive heavy line loads while maintaining the high level of linearity required in OFDM PLC modem links. With 15.5dBm of total transmit signal power into 100 Ω line load, the driver achieves -43dB average MTPR distortion across the output spectrum up to 86MHz.

The ISL15100 has two bias current control pins (C0, C1) to allow for four power settings (disable, low, medium, high). In disable mode, the line driver outputs maintain a high impedance in the presence of high receive signal amplitude, so it doesn't affect TDM receive signal integrity.

The ISL15100 is available in the thermally-enhanced 16 Ld QFN and is specified for operation over the full -40 $^\circ$ C to +85 $^\circ$ C ambient temperature range.

Features

- Single differential driver
- 100MHz Broadband PLC G.hn, EOC, HomePlug AV2
- Control pins for enable/disable and supply current selection
- · High output impedance when disabled for TDM operation
- · -43dBc average MTPR distortion at full line power
- Single +12V or bipolar ±6V nominal supplies
- High surge current handling capability

Applications

- Power Line Communication differential driver
- Pin compatible upgrade to ISL1571IRZ

Related Literature

• AN1325 "Choosing and Using Bypass Capacitors"

TABLE 1. ALTERNATE SOLUTIONS

| PART # | NOMINAL ±V _S (V) | BANDWIDTH (MHz) | APPLICATIONS |
|---------|--------------------------------|--------------------|--------------|
| ISL1571 | ±6, +12 | 200 | HomePlug AV1 |

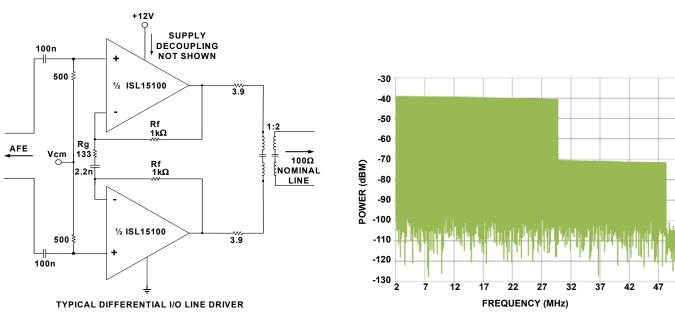
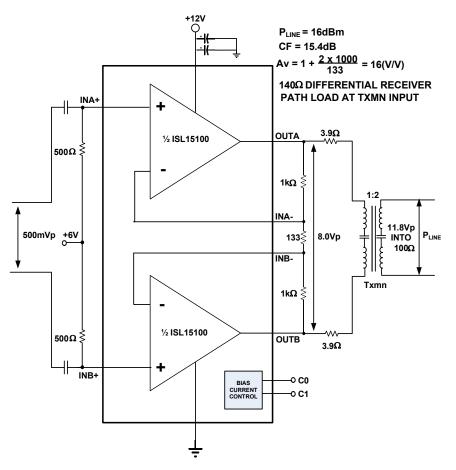


FIGURE 1. TYPICAL APPLICATION CIRCUIT



Connection Diagram





Ordering Information

| PART NUMBER (Notes 2, 3) | PART MARKING | TEMP RANGE (°C) | PACKAGE (Pb-free) | PKG. DWG. # |
|--------------------------------|------------------|--------------------|----------------------|----------------|
| ISL15100IRZ | 151 00IRZ | -40 to +85 | 16 Ld QFN | L16.4x4H |
| ISL15100IRZ-T7 (Note 1) | 151 00IRZ | -40 to +85 | 16 Ld QFN | L16.4x4H |
| ISL15100IRZ-T13 (Note 1) | 151 00IRZ | -40 to +85 | 16 Ld QFN | L16.4x4H |
| ISL15100EVAL1Z | Evaluation Board | | | · · · |

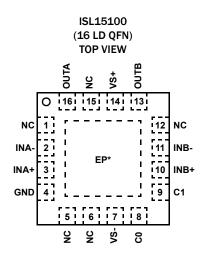
NOTES:

1. Please refer to $\underline{\mathsf{TB347}}$ for details on reel specifications.

- These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
- 3. For Moisture Sensitivity Level (MSL), please see device information page for ISL15100. For more information on MSL please see tech brief TB363.



Pin Configuration



'EXPOSED THERMAL PAD CONNECTS TO MOST NEGATIVE SUPPLY

Pin Descriptions

| PIN NUMBER | PIN NAME | FUNCTION |
|----------------------------|-----------------------------|--|
| EP | THERMAL PAD | Connect to the Most Negative Supply |
| 1 | NC | No Internal Connection |
| 2 | INA- | Amplifier A Inverting Input |
| 3 | INA+ | Amplifier A Non-Inverting Input |
| 4 | GND | Ground |
| 5 | NC | No Internal Connection |
| 6 | NC | No Internal Connection |
| 7 | V _S - | Negative Supply Voltage (-6V for split supplies, GND for single supply operation) |
| 8 | CO | Digital Control Pin |
| 9 | C1 | Digital Control Pin |
| 10 | INB+ | Amplifier B Non-Inverting Input |
| 11 | INB- | Amplifier B Inverting Input |
| 12 | 2 NC No Internal Connection | |
| 13 OUTB Amplifier B Output | | Amplifier B Output |
| 14 | V _S + | Positive Supply Voltage (+6V for split supplies, +12V for single supply operation) |
| 15 | NC | No Internal Connection |
| 16 | OUTA | Amplifier A Output |

CO, C1 Truth Table

| C1 | CO | FUNCTION |
|----|----|-------------------------------|
| 0 | 0 | High Bias Setting |
| 0 | 1 | Medium Bias Setting |
| 1 | 0 | Low Bias Setting |
| 1 | 1 | Outputs Disabled (Power Down) |



Absolute Maximum Ratings $(T_A = +25^{\circ}C)$

| $V_{S}\text{+}$ Voltage to $V_{S}\text{-}$ or GND |
|--|
| INA+, INB+ Voltage GND to V _S + |
| C ₀ , C ₁ Voltage to GND0.3V to V _S + |
| Current into any Input 8mA |
| Continuous Output Current for Long Term Reliability 50mA |
| Latch-up (Tested per JESD78D, Class II) 100mA |
| ESD Rating |
| Human Body Model (Tested per JESD22-A114F) 4kV |
| Charge Device Model (Tested per JESD22-C101E) 1.5kV |
| Machine Model (Tested per JESD22-A115C) |
| |

Thermal Information

| Thermal Resistance (Typical) | θ _{JA} (°C/W | /) θ _{JC} (°C/W) |
|---|-----------------------|---------------------------|
| 16 Ld QFN Package (Notes 4, 5) | 53 | 16.5 |
| Maximum Junction Temperature (Plastic Pac | kage) | +150°C |
| Storage Temperature Range | | -40°C to +150°C |
| Pb-Free Reflow Profile | | see link below |
| http://www.intersil.com/pbfree/Pb-FreeRe | eflow.asp | |

Operating Conditions

| Ambient Temperature Range | -40°C to +85°C |
|----------------------------|----------------|
| Junction Temperature Range | 40°C to +150°C |

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

NOTES:

- 4. θ_{JA} is measured in free air with the component mounted on a high effective thermal conductivity test board with "direct attach" features. See Tech Brief <u>TB379</u>.
- 5. For θ_{JC} the "case temp" location is the center of the exposed metal pad on the package underside.

Electrical Specifications $V_{S^+} = \pm 12V$, $V_{S^-} = GND = 0V$, see Figure 3, Full Bias ($C_0 = C_1 = 0V$), $T_A = \pm 25$ °C, unless otherwise specified.

| PARAMETER | DESCRIPTION | CONDITIONS | MIN (Note 6) | ТҮР | MAX (Note 6) | UNIT |
|--|------------------------------------|--|-----------------|------|-----------------|--------|
| DYNAMIC PERFORMANCE | 1 | | | | | |
| -3dB Bandwidth | BW | Figure 3, $2V_{P-P}$ differential output at pins | | 180 | | MHz |
| Slew Rate | SR | Differential V_OUT (V_OUTA - V_OUTB) from -5V to +5V (10V_{P-P)} | | 1200 | | V/µs |
| Total Harmonic Distortion | THD, Low Frequency, Light Load | 200kHz, differential 12V _{P-P} , across ≥350Ω differential load | | -88 | -67 | dBc |
| | THD, Low Frequency, Heavy Load | 200kHz, differential 12V _{P-P} , across 29Ω differential load | | -72 | -68 | dBc |
| | THD, High Frequency, Light Load | 4MHz, differential 12V _{P-P} , across ≥350Ω differential load | | -64 | -58 | dBc |
| | THD, High Frequency, Heavy Load | 4MHz, differential 12V _{P-P} , across 29Ω differential load | | -51 | -48 | dBc |
| Avg. Multi-Tone Power Ratio | MTPR | 2MHz to 50MHz, 25kHz tone spacing, P _{LINE} = 15.5dBm, CF = 15dB | | -43 | | dBc |
| Off State Multi-Tone Power Ratio | MTPR-OFF | 2MHz to 50MHz, 25kHz tone spacing, P _{LINE} = 15.5dBm, CF = 15dB | | -55 | | dBc |
| Non-inverting Input Spot Voltage Noise | Eni | F > 1MHz, spot noise voltage on INA+ and INB+ inputs separately | | 6 | | nV/√Hz |
| Non-inverting Input Spot Current Noise | Ini+ | F > 1MHz, spot noise current on INA+ and INB+ inputs separately | | 13 | | pA∕√Hz |
| Inverting Input Spot Current Noise | Ini- | F > 1MHz, spot noise current on INA- and INB- inputs separately | | 50 | | pA∕√Hz |
| DC AND INPUT CHARACTERISTICS | | | | | | |
| Non-Inverting Input Bias Current | I _{B+} | Non-inverting inputs, INA+ and INB+, at mid-supply voltage (Note 7) | -7 | 2 | 7 | μA |
| Non-Inverting Input Bias Current Mismatch | IB+DM | Difference between the INA+ and INB+ bias currents | -0.5 | 0 | 0.5 | μA |
| Inverting Input Bias Current | I _{B-} | Inverting inputs, INA- and INB-, at mid supply voltage (Note 7) | -90 | -30 | 55 | μA |



Electrical Specifications V_{S} + = +12V, V_{S} = GND = 0V, see Figure 3, Full Bias ($C_0 = C_1 = 0V$), $T_A = +25$ °C, unless otherwise specified.

| PARAMETER | DESCRIPTION | CONDITIONS | MIN (Note 6) | ТҮР | MAX (Note 6) | UNIT |
|--|---|--|-----------------|------|-----------------|------|
| Inverting Input Bias Current Mismatch | I _{B-DM} | Difference between the INA- and INB- input bias currents | -35 | 0 | 35 | μA |
| Inverting Input Bias Current Common Mode | I _{B-CM} | Average inverting input bias currents (Note 7) | -90 | -30 | 55 | μA |
| Input Offset Voltage | V _{IOA} , V _{IOB} | Voltage difference from INA+ to INA- and from INB+ to INB- | -85 | 0 | 85 | mV |
| Input Offset Voltage Mismatch | V _{IODM} | V _{IOA} - V _{IOB} | -5 | 0 | 5 | mV |
| Input Offset Voltage Common Mode | V _{IOCM} | Average offset voltage across the two inputs | -80 | 20 | 80 | mV |
| Differential Mode Output Offset Voltage | V _{OSDM} | Output referred total effect of all differential DC error terms | -7.8 | 0 | 7.8 | mV |
| Common Mode Output Offset Voltage | V _{OSCM} | Output referred total effect of all common mode DC errors | -105 | 40 | 145 | mV |
| Input Headroom to Positive Supply | (V _S +) - V _{IN(MAX)} | INA+ and INB+ required margin to V _S + supply | | 3 | | v |
| Input Headroom to Negative Supply | V _{IN(MIN)} - (V _S -) | INA+ and INB+ required margin to V _S - supply | | 3 | | v |
| OUTPUT CHARACTERISTICS | | | | | | |
| Output Swing | V _{O-OPEN} | $V_S = \pm 6V$, Differential $R_{LOAD} \ge 1k\Omega$, each output pin voltage range | ±4.85 | ±5.0 | | v |
| | V _{O-LOADED} | $V_S = \pm 6V$, V_O in linear region, Differential $R_{LOAD} = 29\Omega$, each output pin voltage range. | | ±4.6 | | v |
| | | $V_S = \pm 6V$, V_O driven into the rail, differential $R_{LOAD} = 29\Omega$, each output pin voltage range. | ±4.2 | ±4.7 | | v |
| Output Current | Io | Linear output current (not short circuit) | ±300 | ±400 | | mA |
| POWER SUPPLY | | | | | | |
| Bipolar Supply Range | ±VS | Symmetric supply, pin 4 at GND for logic reference | ±4 | ±6 | ±6.6 | v |
| Single Supply Range | V _S + | Single supply with $\mathrm{V}_{\mathrm{S}^{\mathrm{-}}}$ and pin 4 at GND | 8 | 12 | 13.2 | v |
| Positive Supply Currents | I _S + (Full bias) | $V_{O(DIFF)} = 0V, C_0 = C_1 = 0V$ | 27 | 32 | 37 | mA |
| | I _S + (Medium bias) | $V_{O(DIFF)} = 0V, C_0 = 3.3V, C_1 = 0V$ | 19 | 23 | 26 | mA |
| | I _S + (Low bias) | $V_{O(DIFF)} = 0V, C_0 = 0V, C_1 = 3.3V$ | 12 | 15 | 18 | mA |
| | I _S + (Power down) | $C_0 = C_1 = 3.3V$ | 5.5 | 7 | 8.5 | mA |
| C ₀ , C ₁ Input High Current | I _{INH} , C ₀ or C ₁ | $C_0 = C_1 = 3.3V \text{ (Note 7)}$ | -150 | -90 | -30 | μA |
| C ₀ , C ₁ Input Low Current | I _{INL} , C ₀ or C ₁ | $C_0 = C_1 = 0V (Note 7)$ | -1.5 | 1 | 1.5 | μA |
| C ₀ , C ₁ Logic High Voltage | V _{INH} | Pin 4 at GND, logic reference pin | 2 | 3.3 | 5.5 | v |
| C ₀ , C ₁ Logic Low Voltage | V _{INL} | Pin 4 at GND, logic reference pin | -0.3 | 0 | 0.8 | v |

NOTES:

6. Compliance to data sheet limits is assured by one or more methods: production test, characterization and/or design.

7. Positive currents flow out of the pin.



Applications Information

Product Description

The ISL15100 is a differential operational amplifier designed for line driving in power line communications (PLC). It is a low distortion, current mode feedback amplifier that draws moderately low supply current. Due to the current feedback architecture, the ISL15100 closed-loop -3dB bandwidth is dependent on the value of the feedback resistor. The desired bandwidth is selected by choosing the feedback resistor, R_F , and then the gain is set by picking the gain resistor, R_G .

Feedback Resistor Values

The ISL15100 has been designed and specified with $R_F = 1 k \Omega$ for $A_V = +16$. As is the case with all current feedback amplifiers, wider bandwidth at the expense of slight peaking, can be obtained by reducing the value of the feedback resistor. Inversely, larger values of the feedback resistor will cause rolloff to occur at a lower frequency.

Quiescent Current vs Temperature

The ISL15100 was designed to have the quiescent current increase with temperature, which maintains good distortion performance at high temperatures.

Supply Voltage Range

The ISL15100 operates with bipolar supply voltages from $\pm 4.0V$ to $\pm 6.6V$ ($\pm 6.65V$ maximum). Optimum bandwidth, slew rate, and video characteristics are obtained at higher supply voltages.

Single Supply Operation

If a single supply is desired, values from +8.0V to +13.2V (+13.3V maximum) can be used as long as the input common mode range is not exceeded. When using a single supply, be sure to either:

- **1**. DC bias the inputs at an appropriate common mode voltage and AC-couple the signal.
- 2. Ensure the driving signal is within the common mode range of the ISL15100.

Multi-Tone Power Ratio (MTPR)

G.hn PLC uses OFDM modulation to digitally encode data for communication. A carrier spacing of 24.41kHz is used in power lines, and 48.82kHz is used in phone lines.

In multi-tone signaling, linearity is shown in the MTPR measurement. MTPR measures the difference in power of a carrier tone vs. a missing tone.

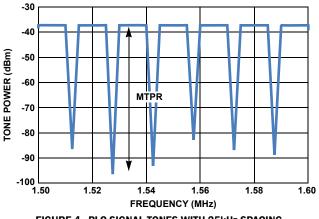


FIGURE 4. PLC SIGNAL TONES WITH 25kHz SPACING

Figure 4 shows ISL15100's MTPR performance for a narrow frequency span.

Disable Linearity

Unlike DSL, communication in PLC systems is half duplex meaning only one device can transmit at a time. When the line driver is not transmitting, it is disabled and the receiver is enabled. Figure 5 shows the shared transmit and receive signal path of two ends. When Txa is transmitting, optimal MTPR is achieved if Txb is removed. Since Txb cannot be removed, the best MTPR occurs if the line driver is a very high impedance when disabled.

 R_{BM} are resistors to limit fault currents, and to provide a driving impedance to the transformer, thus setting its frequency span. R_{BM} is typically low in value (<10 Ω).

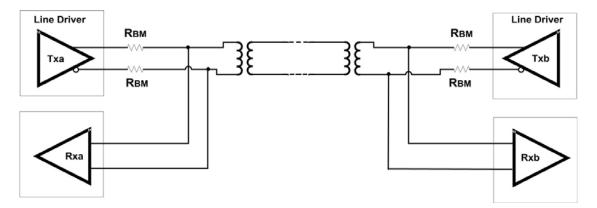


FIGURE 5. TX AND RX SIGNAL PATH. CASE1:[Txa: ON, Rxa: OFF, Txb: OFF, Rxb: ON]. CASE2:[Txa: OFF, Rxa: ON, Txb: ON, Rxb: OFF]

PC Board Design Recommendation

To minimize parasitic capacitance in the ISL15100 design, consider laying out short output traces. Also, select low capacitance protection devices, and use line transformers with low interwinding capacitance in the signal path.

The supply decoupling capacitors must be close to the supply pins to minimize parasitic inductance in the supply paths. High frequency load currents are pulled through these capacitors, so placement of the 0.1μ F capacitors close to the supply pin(s) improves dynamic performance. The higher value 4.7μ F capacitors provide low frequency decoupling, so they can be placed farther from the supply pins.

The ISL15100's thermal pad (EP) should be connected to V_{S^-} (ground in single supply applications). For good thermal control, include a thermal pad in the layout footprint, as shown in the "Typical Recommended Land Pattern" on the "Package Outline Drawing" page. Adding vias to this thermal pad helps dissipate heat away from the package. The ISL15100 evaluation board uses four 10mil (hole size) vias with 20mil diameter pads.

Thermal Resistance and Power Dissipation

Thermal resistance for junction to ambient, T_{JA} , is +53 °C/W. The power dissipation at 12V supply is 600mW. The ambient temperature allowed given the maximum junction temperature of +150 °C is:

$$\begin{split} T_{A} &= T_{J} - \theta_{JA} \times Pd \ & (\text{EQ. 1}) \\ T_{A} &= +150^{\circ}\text{C} - 53^{\circ}(\text{C}/\text{W})*600\text{mW} = +118^{\circ}\text{C} \end{split}$$

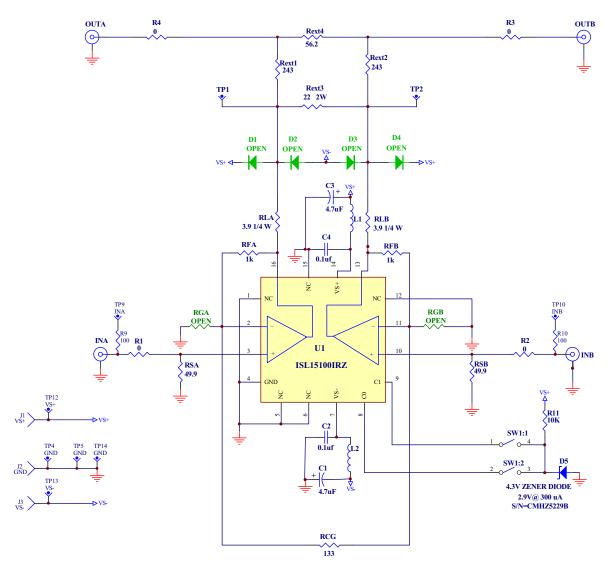


FIGURE 6. ISL15100 EVALUATION BOARD



Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest revision.

| Sentember 19 2013 EN8577.0 Initial Release | DATE | REVISION | CHANGE |
|--|-------------------|----------|------------------|
| | September 19 2013 | FN8577.0 | Initial Release. |

About Intersil

Intersil Corporation is a leader in the design and manufacture of high-performance analog, mixed-signal and power management semiconductors. The company's products address some of the largest markets within the industrial and infrastructure, personal computing and high-end consumer markets. For more information about Intersil, visit our website at <u>www.intersil.com</u>.

For the most updated datasheet, application notes, related documentation and related parts, please see the respective product information page found at <u>www.intersil.com</u>. You may report errors or suggestions for improving this datasheet by visiting <u>www.intersil.com/en/support/ask-an-expert.html</u>. Reliability reports are also available from our website at <u>http://www.intersil.com/en/support/qualandreliability.html#reliability</u>

© Copyright Intersil Americas LLC 2013. All Rights Reserved. All trademarks and registered trademarks are the property of their respective owners.

For additional products, see www.intersil.com/en/products.html

Intersil products are manufactured, assembled and tested utilizing ISO9001 quality systems as noted in the quality certifications found at <u>www.intersil.com/en/support/qualandreliability.html</u>

Intersil products are sold by description only. Intersil may modify the circuit design and/or specifications of products at any time without notice, provided that such modification does not, in Intersil's sole judgment, affect the form, fit or function of the product. Accordingly, the reader is cautioned to verify that datasheets are current before placing orders. Information furnished by Intersil is believed to be accurate and reliable. However, no responsibility is assumed by Intersil or its subsidiaries for its use; nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of Intersil or its subsidiaries.

For information regarding Intersil Corporation and its products, see www.intersil.com

