# **IVIXI/VI** 1575MHz/1900MHz Variable-IP3 Low-Noise Amplifiers

### **General Description**

The MAX2654/MAX2655/MAX2656 high third-order intercept point (IP3), low-noise amplifiers (LNAs) are designed for applications in GPS, PCS, WLL, and satellite phone systems. The MAX2654/MAX2655/MAX2656 incorporate on-chip internal output matching to  $50\Omega$ , eliminating the need for external matching components. A shutdown feature in the MAX2654/MAX2655 reduces operating current to 0.1µA, eliminating the need for an external supply switch.

The MAX2654 operates in the GPS frequency of 1575MHz with 15.1dB of gain, 1.5dB noise figure, and only consumes 5.8mA. The MAX2655 is designed with high-input IP3 to improve operation in cellular applications where the cellular power amplifier leaks into the GPS receiver. The MAX2656 is designed for PCS phone applications with 13.5dB of gain in high-gain mode and 0.8dB of gain in low-gain mode (selected by a logic control) and 1.9dB noise figure.

The IP3 of MAX2655/MAX2656 is adjustable by a single external bias resistor (R\_{BIAS}), allowing supply current to be optimized for a specific application.

The MAX2654/MAX2655/MAX2656 operate from +2.7V to +5.5V single supply and are available in the miniature 6-pin SC70 package.

#### **Applications**

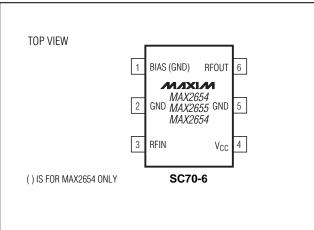
**GPS** Receivers

GPS Receivers in Cell Phones

DCS/PCS Cell Phones

Satellite Phones

Wireless Local Loop



### Pin Configuration

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# For price, delivery, and to place orders, please contact Maxim Distribution at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

#### **Features**

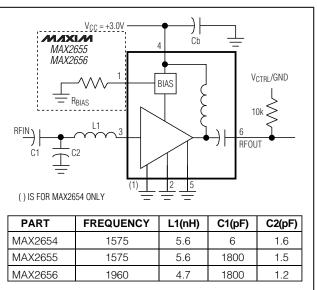
- Low Noise Figure MAX2654: 1.5dB at 1575MHz MAX2655: 1.45dB at 1575MHz MAX2656: 1.9dB at 1960MHz
- High Gain MAX2654: 15.1dB at 1575MHz MAX2655: 14.1dB at 1575MHz MAX2656: 13.5dB at 1960MHz
- + 12.7dB Gain Step (MAX2656 only)
- Integrated 50Ω Output Matching
- Variable IP3 Set by One Bias Resistor (MAX2655/MAX2656 only)
- 0.1µA Shutdown Mode (MAX2654/MAX2655 only)
- + +2.7V to +5.5V Single-Supply Operation
- Ultra-Small 6-Pin SC70 Package

### \_Ordering Information

PART	TEMP. RANGE	PIN- PACKAGE	sot Top- Mark
MAX2654EXT-T	-40°C to +85°C	6 SC70-6	AAI
MAX2655EXT-T	-40°C to +85°C	6 SC70-6	AAJ
MAX2656EXT-T	-40°C to +95°C	6 SC70-6	AAK

### **Typical Operating Circuit**

Maxim Integrated Products 1



#### **ABSOLUTE MAXIMUM RATINGS**

V <sub>CC</sub> to GND	0.3V to +6V
RF Input Power	+5dBm
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
6-Pin SC70 (derate 3.1mW/°C above +70°C)	245mW

Operating Temperature Range	40°C to +85°C
Maximum Junction Temperature	+150°C
Storage Temperature	
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +2.7V \text{ to } +5.5V, R_{BIAS} = 511\Omega \pm 1\%, T_A = -40^{\circ}C \text{ to } +85^{\circ}C. \text{ No RF signal applied. RFIN is AC-coupled and terminated to 50\Omega. RFOUT is unconnected. Typical values are at V_{CC} = +3V, T_A = +25^{\circ}C, unless otherwise noted.) (Note 1)$ 

PARAMETER		CONDITIONS	MIN	ТҮР	MAX	UNIT
Supply Voltage			2.7		5.5	V
		$T_A = +25^{\circ}C$		5.8	8.2	
Operating Supply Current	VRFOUT = VCC	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$			9.2	mA
(MAX2654 only)		$T_A = +25^{\circ}C$		0.1	1	
	V <sub>RFOUT</sub> = GND	$T_{A} = -40^{\circ}C \text{ to } +85^{\circ}C$			2.1	μA
		$R_{BIAS} = 511\Omega \pm 1\%$		8.3	10	
		$R_{BIAS} = 698\Omega \pm 1\%$		5.9		
	$V_{RFOUT} = V_{CC}$ $T_A = +25^{\circ}C$	$R_{BIAS} = 357\Omega \pm 1\%$		10.1		mA
Operating Supply Current (MAX2655 only)	TA = +23 0	$R_{BIAS} = 511\Omega \pm 1\%,$ $T_{A} = -40^{\circ}C \text{ to } +85^{\circ}C$			11.1	
	V <sub>RFOUT</sub> = GND	$T_{A} = +25^{\circ}C$		0.1	1	,
		$T_{A} = -40^{\circ}C \text{ to } +85^{\circ}C$			2.2	μA
	V <sub>RFOUT</sub> = GND T <sub>A</sub> = +25°C	$R_{BIAS} = 511\Omega \pm 1\%$		11.5	15.2	mA
		$R_{BIAS} = 715\Omega \pm 1\%$		8.5		
Operating Supply Current (MAX2656 only)		$\begin{aligned} R_{\text{BIAS}} &= 511\Omega \pm 1\%, \\ T_{\text{A}} &= +85^{\circ}\text{C} \end{aligned}$		13.6		
		$T_A = +25^{\circ}C$		12.3		
	VRFOUT = VCC	$T_A = +85^{\circ}C$		14.6		mA
Input Logic High at V <sub>RFOUT</sub>	(Note 2)		2			V
Input Logic Low at VRFOUT	(Note 3)				0.6	V
Input Logic High Current	MAX2654/MAX2655				15.6	
at V <sub>RFOUT</sub> (Note 4)	MAX2656				71	μA
Input Logic Low Current	MAX2654/MAX2655				1	
at V <sub>RFOUT</sub> (Note 5)	MAX2656				-24	<u>24</u> μΑ

### AC ELECTRICAL CHARACTERISTICS (MAX2654)

(MAX2654 Evaluation Kit,  $V_{CC}$  = +3.0V,  $P_{IN}$  = -30dBm,  $f_{IN}$  = 1575MHz,  $V_{RFOUT}$  =  $V_{CC}$  through a 10k $\Omega$  resistor,  $T_A$  = +25°C. Typical values are at  $V_{CC}$  = +3V,  $T_A$  = +25°C, unless otherwise noted.) (Note 6)

PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNIT
Operating Frequency Range (Note 7)		1400		1700	MHz
Gain	(Note 8)	12.7	15.1		dB
Gain Variation Over Temperature	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$		0.15	1	dB
Input Third-Order Intercept Point (Note 9)			-7.2		dBm
Input 1dB Compression Point			-18		dBm
Noise Figure (Note 10)			1.5	1.8	dB
Input Return Loss			9.7		dB
Output Return Loss			8.4		dB
Reverse Isolation			30		dB

### **AC ELECTRICAL CHARACTERISTICS (MAX2655)**

(MAX2655 Evaluation Kit,  $V_{CC} = +3.0V$ ,  $P_{IN} = -30dBm$ ,  $f_{IN} = 1575MHz$ ,  $V_{RFOUT} = V_{CC}$  through a  $10k\Omega$  resistor,  $R_{BIAS} = 511\Omega \pm 1\%$ ,  $T_A = +25^{\circ}C$ . Typical values are at  $V_{CC} = +3V$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.) (Note 6)

PARAMETER	CONDITIONS	MIN	ТҮР	МАХ	UNIT
Operating Frequency Range (Note 7)		1400		1700	MHz
Gain	(Note 8)	12	14.1		dB
Gain Variation Over Temperature	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$		0.6	1.1	dB
Input Third-Order Intercept Point	$R_{BIAS} = 511\Omega \pm 1\%$		2.8		
	$R_{BIAS} = 698\Omega \pm 1\%$		2.2		dBm
(Note 9)	$R_{BIAS} = 357\Omega \pm 1\%$		3.8		
Input 1dB Compression Point			-12.2		dBm
Noise Figure (Note 10)			1.45	1.9	dB
Input Return Loss			16.1		dB
Output Return Loss			15.5		dB
Reverse Isolation			32		dB

### **AC ELECTRICAL CHARACTERISTICS (MAX2656)**

(MAX2656 Evaluation Kit,  $V_{CC} = +3V$ ,  $P_{IN} = -30dBm$ ,  $f_{IN} = 1960MHz$ ,  $V_{RFOUT} = GND$  through a  $12k\Omega$  resistor,  $R_{BIAS} = 511\Omega \pm 1\%$ ,  $T_A = +25^{\circ}C$ . Typical values are at  $V_{CC} = +3V$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.) (Note 6)

PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNIT	
Operating Frequency Range (Note 7)		1800		2000	MHz	
Gain, High-Gain Mode	(Note 11)	12	13.5		dB	
Gain, Low-Gain Mode	(Note 12)		0.8		dB	
Gain Variation Over Temperature	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$		0.3	1.2	dB	
Gain Step			12.7		dB	
	$R_{BIAS} = 511\Omega \pm 1\%$		1.5			
Input Third-Order Intercept Point (Note 13)	$R_{BIAS} = 715\Omega \pm 1\%$		-3		dBm	
	Low-gain mode (Note 12)		7.2			
Input 1dB Compression Doint	High-gain mode (Note 11)		-7		d Duos	
Input 1dB Compression Point	Low-gain mode (Note 12)		-1.2		dBm	
Naisa Figura (Nata 10)	High-gain mode (Note 11)		1.9	2.4		
Noise Figure (Note 10)	Low-gain mode (Note 12)		10.8		dB	
Incut Datura Laga	High-gain mode (Note 11)		14.4		dB	
Input Return Loss	Low-gain mode (Note 12)		19.3		uв	
Outraut Datura Laga	High-gain mode (Note 11)		10.7			
Output Return Loss	Low-gain mode (Note 12)		7.3		dB	
Reverse Isolation	High-gain mode (Note 11)		28		dD	
	Low-gain mode (Note 12)		25		dB	
Gain Step Response Time			3.2		μs	

**Note 1:** Devices are production tested at  $T_A = +25^{\circ}$ C. Minimum and maximum values are guaranteed by design and characterization over temperature and supply voltages to ±6 sigma.

Note 2: Minimum DC voltage through a 10kΩ resistor that sets the MAX2654/MAX2655 to operate in normal mode and MAX2656 in low-gain mode.

Note 3: Maximum DC voltage through a 10kΩ resistor that sets the MAX2654/MAX2655 to operate in shutdown mode and MAX2656 in high-gain mode.

Note 4: DC current required when  $V_{RFOUT}$  is connected to  $V_{CC}$  through a 10k $\Omega$  resistor.

Note 5: DC current required when  $V_{RFOUT}$  is connected to GND through a 10k $\Omega$  resistor.

**Note 6:** Guaranteed by design and characterization to ±3 sigma.

**Note 7:** The part has been characterized at the specified frequency range. Operation outside of this range is possible, but not guaranteed.

**Note 8:** Production tested at  $T_A = +25^{\circ}C$ .

Note 9: Measured with two input tones,  $f_1 = 1570$ MHz,  $f_2 = 1580$ MHz, both at -30dBm per tone.

Note 10: Excludes PC board losses of 0.2dB for MAX2654/MAX2655 and 0.25dB for MAX2656.

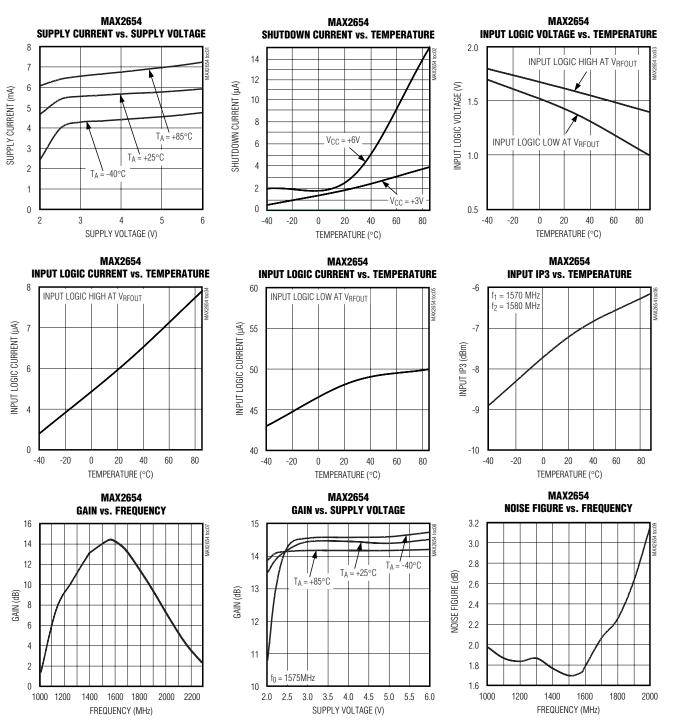
Note 11: High-gain mode is set up by connecting RFOUT to GND through a  $12k\Omega$  resistor.

Note 12: Low-gain mode is set up by connecting RFOUT to V\_{CC} through a 12k  $\!\Omega$  resistor.

Note 13: Measured with two input tones,  $f_1 = 1955MHz$ ,  $f_2 = 1965MHz$ , both at -30dBm per tone.

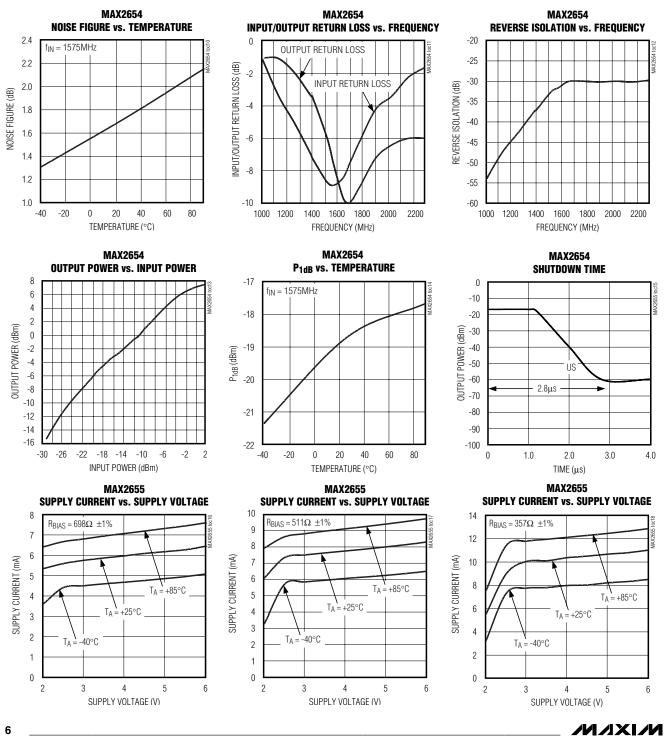
### **Typical Operating Characteristics**

 $(P_{IN} = -30 \text{dBm}, \text{ input} \text{ and output} \text{ are terminated to } 50\Omega, V_{CC} = +3V, \text{ high-gain and low-gain modes are applicable only to the MAX2656, } T_A = +25^{\circ}C, \text{ unless otherwise noted.})$ 



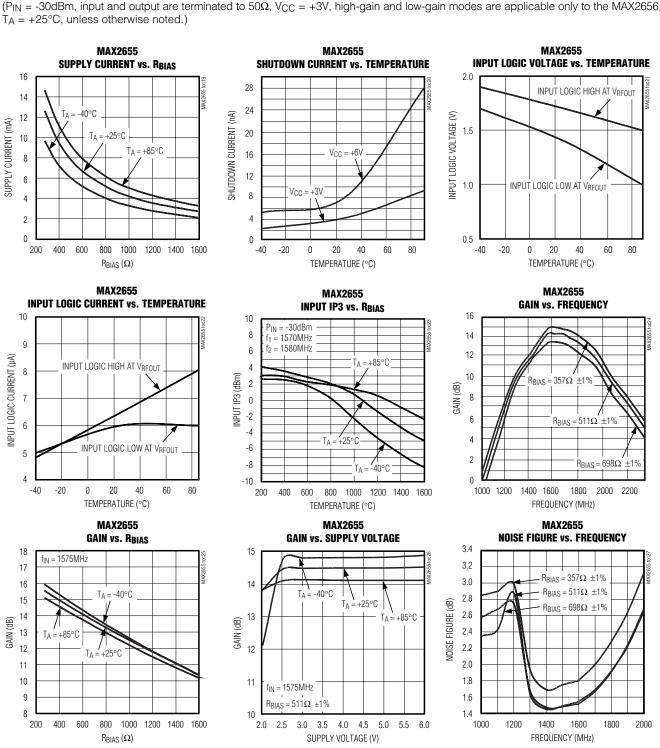
### Typical Operating Characteristics (continued)

(PIN = -30dBm, input and output are terminated to 50Ω, V<sub>CC</sub> = +3V, high-gain and low-gain modes are applicable only to the MAX2656,  $T_A = +25^{\circ}C$ , unless otherwise noted.)



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Typical Operating Characteristics (continued)



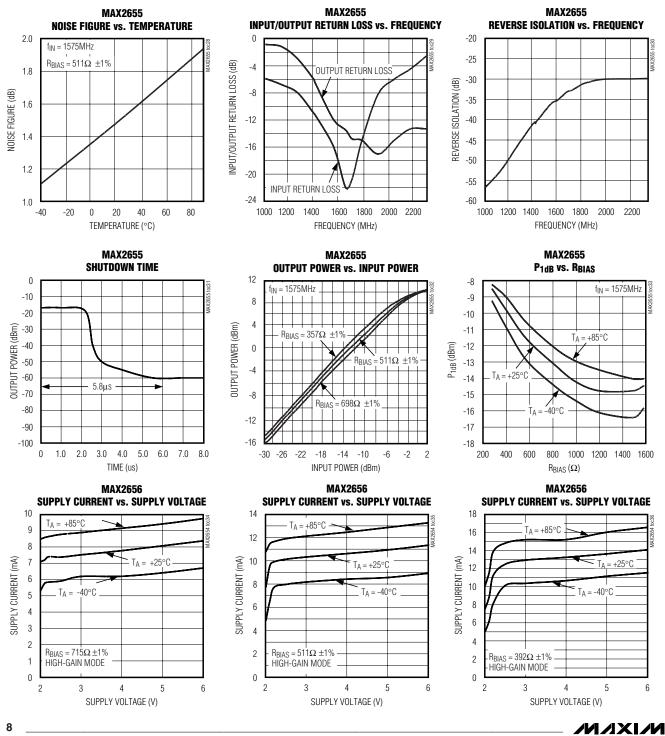
(PIN = -30dBm, input and output are terminated to  $50\Omega$ , V<sub>CC</sub> = +3V, high-gain and low-gain modes are applicable only to the MAX2656.

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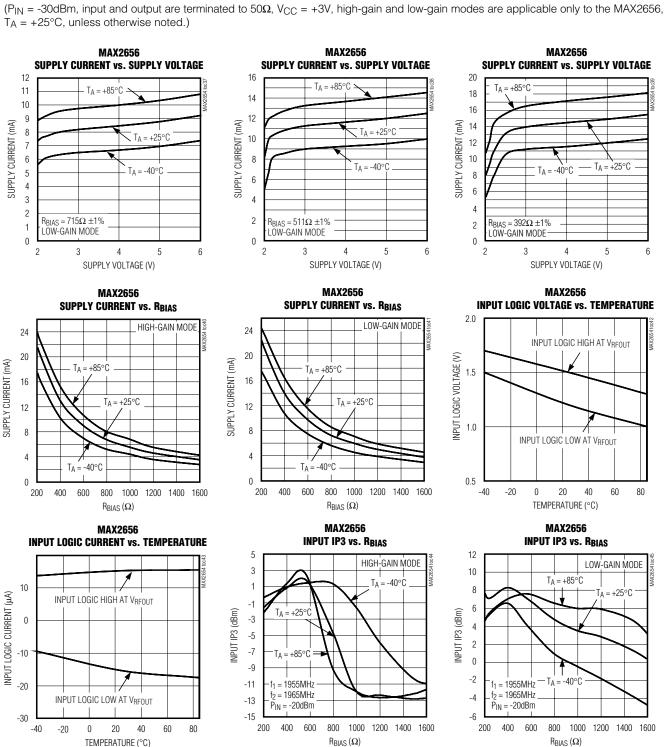
MAX2654/MAX2655/MAX2656

#### **Typical Operating Characteristics (continued)**

(P<sub>IN</sub> = -30dBm, input and output are terminated to  $50\Omega$ , V<sub>CC</sub> = +3V, high-gain and low-gain modes are applicable only to the MAX2656,  $T_A = +25^{\circ}C$ , unless otherwise noted.)



MAX2654/MAX2655/MAX2656

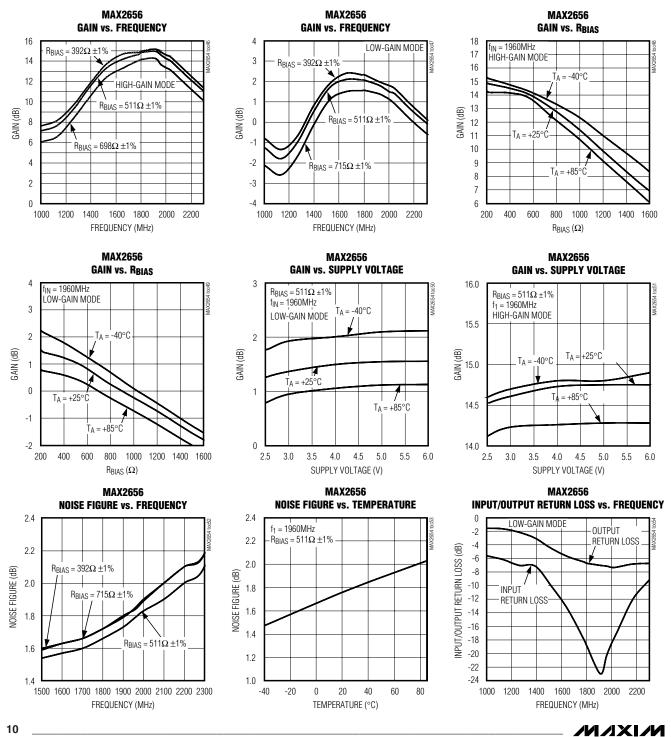


**Typical Operating Characteristics (continued)** n, input and output are terminated to  $50\Omega$ ,  $V_{CC} = +3V$ , high-gain and low-gain modes are applicable only to the MAX2656,

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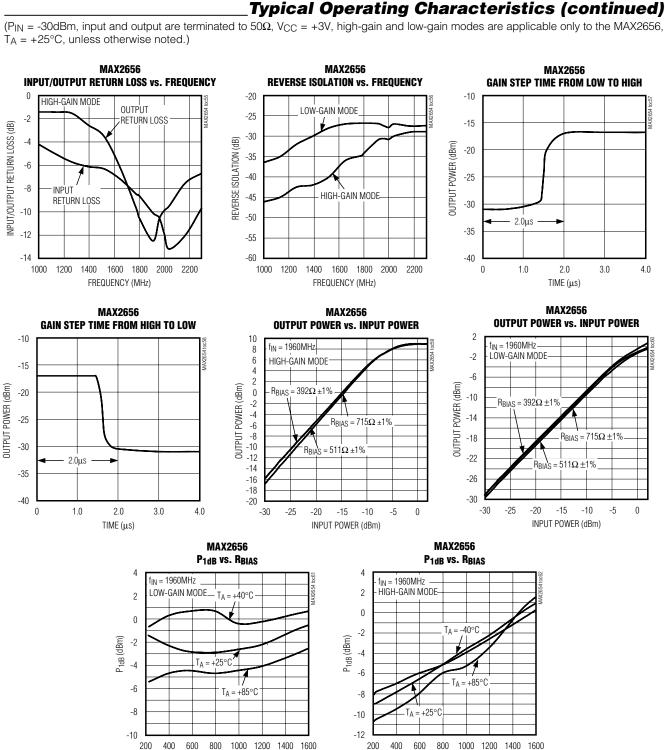
### **Typical Operating Characteristics (continued)**

(PIN = -30dBm, input and output are terminated to  $50\Omega$ , V<sub>CC</sub> = +3V, high-gain and low-gain modes are applicable only to the MAX2656.  $T_A = +25^{\circ}C$ , unless otherwise noted.)



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 $R_{BIAS}(\Omega)$ 



 $R_{BIAS}(\Omega)$ 

/N/IXI/N

### **Pin Description**

P	IN			
MAX2654	MAX2655/ MAX2656	NAME	DESCRIPTION	
1		GND	Ground. Connect to the PC board ground plane through a 0.017in x 0.035in line.	
2, 5	2, 5	GND	Ground. Connect to the PC board ground plane with as low an inductance path as possible.	
	1	BIAS	Bias Control. Connect a resistor $R_{BIAS}$ from $BIAS$ to GND. $R_{BIAS}$ sets IP3 and supply current.	
3	3	RFIN	Amplifier Input. Requires a DC-blocking capacitor and external matching components.	
4	4	V <sub>CC</sub>	Supply Voltage. Bypass to ground with an appropriate capacitor as close to the IC a possible. Refer to MAX2654/MAX2655/MAX2656 EV kits for capacitor values.	
6	6	RFOUT	RF Output. Incorporates an internal DC-blocking capacitor. RFOUT is internally matched to $50\Omega$ . DC bias on this pin selects gain mode (MAX2656) or shutdown mode (MAX2654/MAX2655) (see <i>Applications Information</i> ).	

#### **Detailed Description**

The MAX2654/MAX2655/MAX2656 are low-noise amplifiers designed for applications in GPS receivers, satellite and PCS phones. The MAX2655/MAX2656 feature variable IP3s, adjusted by a single external bias resistor. Another feature of the MAX2654/MAX2655 is a power shutdown control mode, eliminating the need for an external supply switch. The MAX2656 features a high- and low-gain control mode selected by an external logic control.

#### **Input and Output Ports**

The MAX2654/MAX2655/MAX2656 incorporate on-chip matching networks to  $50\Omega$  at the output ports, eliminating the need for external matching components. (For MAX2655, a shunt inductor is recommended for best output return loss. Refer to the MAX2655 EV kit schematic.) The MAX2654/MAX2655/MAX2656 require simple matching networks at the input ports. The values of these matching components are recommended in the *Typical Operating Circuit*.

**Variable IP3** The IP3 of MAX2655 and MAX2656 is adjusted through an external resistor (R<sub>BIAS</sub>). Tables 1 and 2 summarize the values of R<sub>BIAS</sub> for different IP3s for MAX2655 and MAX2656.

#### Table 1. RBIAS vs. IP3 for MAX2655

<b>R</b> BIAS (Ω)	IP3 (dBm)	I <sub>CC</sub> (mA)
698	2.2	5.8
357	3.8	10.1

#### Table 2. RBIAS vs. IP3 for MAX2656

<b>Β</b> ΙΑS (Ω)	IP3 (dBm)	I <sub>CC</sub> (mA)
715	-3	8.5
511	1.5	11.5

#### **Gain-Step Control (MAX2656)**

The DC bias voltage at RFOUT of the MAX2656 serves as a gain-step control input. When the applied DC voltage at RFOUT through a  $10k\Omega$  resistor is less than +0.6V, the device is in high-gain mode. When the applied DC voltage is greater than +2.0V, the device gain is attenuated by 13dB. A standard logic output can be applied as shown in the *Typical Operating Circuit*.

#### Shutdown-Enable (MAX2654/MAX2655)

The DC bias voltage at RFOUT of the MAX2654/ MAX2655 serves as a shutdown enable input. When the applied DC voltage at RFOUT through a 10k $\Omega$  resistor is less than +0.6V, the device is in shutdown mode. When the DC voltage is greater than +2V, the device is enabled.



### **Applications Information**

#### **Input Matching**

For optimum performance, input matching is required. The MAX2654/MAX2655/MAX2656 require a simple LC matching network. The *Typical Operating Circuit* shows the recommended input matching networks. These values are optimized for best simultaneous gain, noise figure, and return loss performance. S-Parameter data can be found on the Maxim website at www.maxim-ic.com.

#### **Layout Issues**

A properly designed PC board is essential to any RFmicrowave circuit. Use controlled impedance lines on all high-frequency inputs and outputs. Bypass  $V_{CC}$  with decoupling capacitors located close to the device. For long V<sub>CC</sub> lines, it may be necessary to add decoupling capacitors. Locate these additional capacitors further away from the device package. Proper grounding of the GND pins is essential. If the PC board uses a topside RF ground, connect it directly to all GND pins. For a board where the ground plane is not on the component layer, connect the GND pins to the board with a multiple vias close to the package.

For MAX2654, connect Pin1 to the PC board ground plane through a 0.017in x 0.035in line.

#### Chip Information

TRANSISTOR COUNT: 135 (Same for MAX2654, MAX2655, MAX2656)