

Evaluates: MAX40263 MAX40263 Evaluation Kit

General Description

The MAX40263 evaluation kit (EV kit) provides a proven design to evaluate the MAX40263 low-noise, low-power, low-bias-current dual operational amplifier with independent shutdown for each channel in a 10-pin QFN package. The EV kit circuit is preconfigured as noninverting amplifiers, but it can be adapted to other topologies by changing a few of components. The EV kit comes with a MAX40263AVB+ installed.

Features

- Accommodates Multiple Op-Amp Configurations
- Component Pads Allow for Sallen-Key Filter
- Accommodates Easy-to-Use Components
- Proven PCB Layout
- **Fully Assembled and Tested**

Quick Start

Required Equipment

- MAX40263 EV kit
- 1.7V to 5.5V, 1A DC power supply
- Precision voltage source
- **Digital multimeter**

MAX40263 EV Kit Photo MAX40263 EV Kit Files

Procedure

The EV kit is fully assembled and tested.

Caution: Do not turn on the power supply until all connections are completed.

Take channel A as an example, follow the steps below to verify board operation:

- 1) Verify that all jumpers (JU1–JU7) are in their default positions, as shown in [Table 1.](#page-1-0)
- 2) Set the power supply to +5V, set the current limit to 1A. Connect the positive terminal of the power supply to V_{DD} and the negative terminal to GND.
- 3) Connect the positive terminal of the precision voltage source to INAP. Connect the negative terminal of the precision voltage source to GND. INAN is already connected to GND through jumper JU1.
- 4) Connect the DMM to monitor the voltage on OUTA. With the 10kΩ feedback resistors and 1kΩ series resistors, the gain of the noninverting amplifier is 11V/V.
- 5) Turn on the power supply.
- 6) Apply 100mV from the precision voltage sources. Observe the output at OUTA on the DMM that reads approximately 1.1V.

[Ordering Information](#page-3-0) appears at end of data sheet.

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Detailed Description of Hardware

The MAX40263 EV kit provides a proven layout for the MAX40263 low-power op amp. The device is a singlesupply op amp that is ideal for sensor interfaces, looppowered systems, and various types of medical and data-acquisition instruments.

The default configuration for the device in the EV kit is noninverting configuration.

Op-Amp Configurations

The device is a single-supply op amp ideal for differential sensing, noninverting amplification, buffering, and filtering. A few common configurations are shown in the next few sections.

The following sections explain how to configure the op amp.

Power up requirement

Table 1. Jumper Descriptions

MAX40263 has a built-in offset auto-calibration function during chip power-up, and an undesired offset can be obtained if the chip power-up speed is slow.

To achieve targeted low offset values, it is suggested to either (a) use a power supply with a fast slew rate (power

supply settles in <10ms), or (b) power up the chip in the shutdown mode $(\overline{\text{SHDNA}} = \overline{\text{SHDNA}} = \text{low})$ and enable the chip $(\overline{\text{SHDNA}} = \overline{\text{SHDNB}} = \text{high})$ after the supply settles.

Noninverting Configuration

The EV kit comes preconfigured as a noninverting amplifier. The gain is set by the ratio of R5 and R1 for channel A (R13 and R20 for channel B). The EV kit comes preconfigured for a gain of +11V/V. The output voltage for the noninverting configuration is given by the equation below:

$$
V_{OUTA} = \left(1 + \frac{RS}{R1}\right) (V_{INAP} \pm V_{OS})
$$

$$
V_{OUTB} = \left(1 + \frac{R20}{R13}\right) (V_{INBP} \pm V_{OS})
$$

Inverting Configuration

To configure the EV kit as an inverting amplifier, remove the shunt on jumper JU1 and install a shunt on jumper JU2 and feed an input signal on the INAN PCB pad for channel A, or remove the shunt on jumper JU4 and install a shunt on jumper JU5 and feed an input signal on the INBN PCB pad for channel B.

**Default position.*

Differential Amplifier

To configure the channel A of the EV kit as a differential amplifier, replace R1, R2, R3, and R5 with appropriate resistors. When $R1 = R2$ and $R3 = R5$, the CMRR of the differential amplifier is determined by the matching of the resistor ratios R1/R2 and R3/R5.

$$
V_{\text{OUTA}} = \text{GAINA} (V_{\text{INAP}} - V_{\text{INAN}})
$$

where:

$$
GAINA = \frac{R5}{R1} = \frac{R3}{R2}
$$

In the same way, to configure the channel B of the EV kit as a differential amplifier, replace R13, R17, R18, and R20 with appropriate resistors. When R13 = R17 and R18 = R20, the CMRR of the differential amplifier is determined by the matching of the resistor ratios R13/R17 and R18/R20.

$$
V_{OUTB} = GAMNB(V_{INBP} - V_{INBN})
$$

where:

$$
GAINB = \frac{R20}{R13} = \frac{R18}{R17}
$$

Sallen-Key Configuration

The Sallen-Key topology is ideal for filtering sensor signals with a second-order filter and acting as a buffer. Schematic complexity is reduced by combining the filter and buffer operations. The EV kit can be configured in a Sallen-Key topology by replacing and populating a few components. For channel A, the Sallen-Key topology can be configured as a unity-gain buffer by replacing R5 with a 0Ω resistor and removing resistor R1. The signal is noninverting and applied to INAP. The filter component pads are R2-R3 and R7-R8, where some must be populated with resistors and others with capacitors, and it is similar for channel B.

Lowpass Sallen-Key Filter: To configure the channel A as a lowpass Sallen-Key filter, remove the shunt from jumper JU1, populate the R2 and R8 pads with resistors, and populate the R3 and R7 pads with capacitors. The corner frequency and Q are then given by:

$$
f_C = \frac{1}{2\pi\sqrt{R_{R2}R_{R8}C_{R3}C_{R7}}}
$$

$$
Q = \frac{\sqrt{R_{R2}R_{R8}C_{R3}C_{R7}}}{C_{R3}(R_{R2} + R_{R8})}
$$

To configure the channel B as a lowpass Sallen-Key filter, remove the shunt from jumper JU4, populate the R17 and R23 pads with resistors, and populate the R18 and R22 pads with capacitors. The corner frequency and Q are then given by:

$$
f_{C} = \frac{1}{2\pi\sqrt{R_{R23}R_{R17}C_{R18}C_{R22}}}
$$

$$
Q = \frac{\sqrt{R_{R23}R_{R17}C_{R18}C_{R22}}}{C_{R18}(R_{R23} + R_{R17})}
$$

Highpass Sallen-Key Filter: To configure the channel A as a highpass Sallen-Key filter, remove the shunt from jumper JU1, populate the R3 and R7 pads with resistors, and populate the R2 and R8 pads with capacitors. The corner frequency and Q are then given by:

$$
f_C = \frac{1}{2\pi\sqrt{R_{R3}R_{R7}C_{R2}C_{R8}}}
$$

$$
Q = \frac{\sqrt{R_{R3}R_{R7}C_{R2}C_{R8}}}{R_{R7}(C_{R2} + C_{R8})}
$$

To configure the channel B as a highpass Sallen-Key filter, remove the shunt from jumper JU4, populate the R18 and R22 pads with resistors, and populate the R23 and R17 pads with capacitors. The corner frequency and Q are then given by:

$$
f_{C} = \frac{1}{2\pi\sqrt{R_{R18}R_{R22}C_{R23}C_{R17}}}
$$

$$
Q = \frac{\sqrt{R_{R18}R_{R22}C_{R23}C_{R17}}}{R_{R22}(C_{R23} + C_{R17})}
$$

Bandpass Sallen-Key Filter: To configure the channel A as bandpass Sallen-Key filter, remove the shunt from jumper JU1, replace R8, populate the R3 and R7 pads with resistors, and populate the C8 and R2 pads with capacitors. The corner frequency and Q are then given by:

$$
f_C = \frac{\sqrt{R_{R7} + R_{R8}}}{2\pi\sqrt{C_{C8}C_{R2}R_{R3}R_{R7}R_{R8}}}
$$

$$
Q = \frac{\sqrt{(R_{R7} + R_{R8})C_{C8}C_{R2}R_{R3}R_{R7}R_{R8}}}{R_{R7}R_{R8}(C_{R2} + C_{C8}) + R_{R3}C_{R2}(R_{R7} - R_{R8}\frac{R_{R5}}{R_{R1}})}
$$

To configure the channel B as Bandpass Sallen-Key filter, remove the shunt from jumper JU4, replace R23, populate the R18 and R22 pads with resistors, and populate the C15 and R17 pads with capacitors. The corner frequency and Q are then given by:

$$
f_C = \frac{\sqrt{R_{R22} + R_{R23}}}{2\pi\sqrt{C_{C15}C_{R17}R_{R18}R_{R22}R_{R23}}}
$$

$$
Q = \frac{\sqrt{(R_{R22} + R_{R23})C_{C15}C_{R17}R_{R18}R_{R22}R_{R23}}}{R_{R22}R_{R23}(C_{R17} + C_{C15}) + R_{R18}C_{R17}\left(R_{R22} - R_{R23}\frac{R_{R20}}{R_{R13}}\right)}
$$

Transimpedance Amplifier (TIA)

To configure the EV kit as a TIA, take channel A for example, place a shunt on jumper JU2 and replace R1 with 0Ω resistors. The output voltage of the TIA is the input current multiplied by the feedback resistor:

$$
V_{\text{OUTA}} = -(I_{\text{INA}} + I_{\text{BIAS}}) \times R_{\text{R5}} \pm V_{\text{OS}}
$$

where:

 I_{INA} is the input current source applied at the INAN test point.

IBIAS is the input bias current.

 $V_{\Omega S}$ is the input offset voltage of the op amp.

Use a capacitor and 0Ω resistor at location R4 or R10 (and C8, if applicable) to stabilize the op amp by rolling off high-frequency gain due to a large cable capacitance.

Capacitive Loads

Some applications require driving large capacitive loads. Take channel A for example, the EV kit provides C8 and R6 pads for an optional capacitive-load driving circuit. C8 simulates the capacitive load while R6 acts as an isolation resistor to improve the op amp's stability at higher capacitive loads. To improve the stability of the amplifier in such cases, replace R6 with a suitable resistor value to improve amplifier phase margin.

Note: Indicate that you are using the MAX40263 when contacting these component suppliers.

Ordering Information

#Denotes RoHS compliance.

Component Suppliers

MAX40263 EV Kit Bill of Materials

MAX40263 EV Kit Schematics

MAX40263 EV Kit PCB Layouts

MAX40263 EV Kit Component Placement Guide—Top Silkscreen

MAX40263 EV Kit PCB Layout—Top

MAX40263 EV Kit PCB Layouts (continued)

