

### DESCRIPTION

The MP6922A is a dual, fast-turn-off, intelligent rectifier for synchronous rectification in LLC resonant converters.

The IC drives two N-channel MOSFETs and regulates their forward voltage drop to about 30mV and turns it off before the switching current goes negative.

MP6922A has a light-load function to latch off the gate driver under light-load conditions, thus limiting the current to below 600µA.

The MP6922A's fast turn-off enables both continuous-conduction mode and discontinuous-conduction mode. An internal reverse-current protection function ensures safe MOSFET operation under high-frequency continuous-current conditions.

MP6922A requires a minimal number of readily-available standard external components and is available in SOIC8E, SOIC8, and SOIC14 packages.

### FEATURES

- Works with Both Standard and Logic-Level MOSFETs
- Compatible with Energy Star's 0.5W Standby Requirements
- $V_{DD}$  Range from 8V to 24V
- 30mV  $V_{DS}$  Regulation Function <sup>(1)</sup>
- Fast Turn-Off Total Delay of 20ns
- Reverse-Current-Protection Function
- Maximum Switching Frequency of 300kHz
- Light Load Mode Function <sup>(1)</sup> with <600µA Quiescent Current
- Supports CCM, CrCM and DCM Operation
- Available in SOIC8E, SOIC8, and SOIC14 packages
- Halogen-free

### APPLICATIONS

- AC-DC Adapters
- LCDs & PDP TVs
- Telecom SMPS

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#### Notes:

- 1) Related issued patent: US Patent US8,067,973; US8,400,790. CN Patent ZL201010504140.4; ZL200910059751.X. Other patents pending.

### TYPICAL APPLICATION



SOIC8



SOIC8E



SOIC14

## ORDERING INFORMATION

Part Number	Package	Top Marking
MP6922AGSE*	SOIC8	MP6922A
MP6922AGN**	SOIC8E	MP6922A
MP6922AGS***	SOIC14	MP6922A

\*For Tape & Reel, add suffix -Z (e.g. MP6922AGSE-Z);

\*\*For Tape & Reel, add suffix -Z (e.g. MP6922AGN-Z);

\*\*\*For Tape & Reel, add suffix -Z (e.g. MP6922AGS-Z);

## PACKAGE REFERENCE



### ABSOLUTE MAXIMUM RATINGS <sup>(2)</sup>

$V_{DD}$ to $V_{S1}, V_{S2}, V_{SS}$ .....	-0.3V to +26V
PGND to $V_{S1}, V_{S2}, V_{SS}$ .....	-0.3V to +0.3V
$V_{G1}$ to $V_{S1}, V_{SS}$ .....	-0.3V to $V_{DD}$
$V_{G2}$ to $V_{S2}, V_{SS}$ .....	-0.3V to $V_{DD}$
$V_{D1}$ to $V_{S1}, V_{SS}$ .....	-0.7V to +180V
$V_{D2}$ to $V_{S2}, V_{SS}$ .....	-0.7V to +180V
LL, EN to $V_{S1}, V_{S2}, V_{SS}$ .....	-0.3V to +6.5V
Maximum Operating Frequency .....	300 kHz
Continuous Power Dissipation.. ( $T_A = +25^\circ\text{C}$ ) <sup>(3)</sup>	
SOIC8 .....	1.4W
SOIC8E .....	2.5W
SOIC14 .....	1.5W
Junction Temperature .....	150°C
Lead Temperature (Solder) .....	260°C
Storage Temperature .....	-55°C to +150°C

### Recommended Operation Conditions <sup>(4)</sup>

$V_{DD}$ to $V_{S1}, V_{S2}, V_{SS}$ .....	8V to 24V
Operating Junction Temp. ( $T_J$ ) .....	-40°C to +125°C

Thermal Resistance <sup>(5)</sup>	$\theta_{JA}$	$\theta_{JC}$
SOIC8 .....	90	45
SOIC8E .....	50	10
SOIC14 .....	86	38

#### Notes:

- Exceeding these ratings may damage the device.
- $T_A = +25^\circ\text{C}$ . The maximum allowable power dissipation is a function of the maximum junction temperature  $T_J(\text{MAX})$ , the junction-to-ambient thermal resistance  $\theta_{JA}$ , and the ambient temperature  $T_A$ . The maximum allowable continuous power dissipation at any ambient temperature is calculated by  $P_D(\text{MAX}) = (T_J(\text{MAX}) - T_A) / \theta_{JA}$ . Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- The device is not guaranteed to function outside of its operating conditions.
- Measured on JESD51-7, 4-layer PCB. Without heatsink.

## ELECTRICAL CHARACTERISTICS

$V_{DD} = 12V$ ,  $-40^{\circ}C \leq T_J \leq 125^{\circ}C$ , unless otherwise noted.

Parameter	Symbol	Conditions	Min	Typ	Max	Units
$V_{DD}$ Voltage Range			8		24	V
$V_{DD}$ UVLO Threshold		Rising	4.8	6.0	7.0	V
		Hysteresis	0.5	1	1.5	V
Operating Current	$I_{CC}$	$C_{LOAD}=5nF$ , $F_{SW}=100kHz$	12	18	23	mA
		$C_{LOAD}=10nF$ , $F_{SW}=100kHz$	24	27	31	mA
Shutdown Current		$V_{DD}=20V$ , $EN=0V$			600	$\mu A$
Light-Load Mode Current					600	$\mu A$
Thermal Shutdown				150		$^{\circ}C$
Thermal Shutdown hysteresis				30		$^{\circ}C$
Enable Shutdown Threshold		Rising	1.1	1.5	2.0	V
		Hysteresis		0.2	0.45	V
Enable UVLO Threshold		Rising	2.3	3	3.6	V
		Hysteresis		0.2	0.45	V
Internal Pull-up Current on EN				10	16	$\mu A$
<b>CONTROL CIRCUITRY SECTION</b>						
$V_{S1,2} - V_{D1,2}$ Forward Voltage	$V_{fwd}$		15	30	45	mV
Turn-on delay	$T_{Don}$	$C_{LOAD}=5nF$	$-20^{\circ}C \leq T_J \leq 125^{\circ}C$	150	200	ns
			$-40^{\circ}C \leq T_J < -20^{\circ}C$	250		
	$T_{Don}$	$C_{LOAD}=10nF$	$-20^{\circ}C \leq T_J \leq 125^{\circ}C$	250	300	ns
			$-40^{\circ}C \leq T_J < -20^{\circ}C$	350		
Input Bias Current on $V_{D1,2}$ pin		$V_{D1,2} = 180V$			1.5	$\mu A$
Minimum ON-Time	$T_{MIN}$	$C_{LOAD} = 5nF$		1		$\mu s$
Minimum OFF-Time	$T_{OFF}$			1.6		$\mu s$
Light-Load–Enter Delay	$T_{LL-Delay}$	$R_{LL}=100k\Omega$	80	160	260	$\mu s$
Light-Load–Enter Pulse Width	$T_{LL}$	$R_{LL}=100k\Omega$		2.2		$\mu s$
Light-Load Turn-On Pulse Width Hysteresis	$T_{LL-H}$	$R_{LL}=100k\Omega$		0.2		$\mu s$
Light-Load–Enter OFF Period Width	$T_{LL-OFF}$	$R_{LL}=100k\Omega$		50		$\mu s$
Light-Load Exit-Pulse Width Threshold ( $V_{D1,2}-V_{S1,2}$ )	$V_{LL-DS}$		-450	-300	-140	mV
Light-load Enter-Pulse Width Threshold ( $V_{G1,2}-V_{S1,2}$ )	$V_{LL-GS}$			1.0		V
Reverse-Current–Protection Threshold	$V_{RCP}$		2.5	3	3.5	V
Reverse-Current–Protection Latch Time	$T_{RCP}$		80	150	230	$\mu s$
<b>GATE DRIVER SECTION</b>						
$V_{G1,2}$ (Low)		$I_{LOAD}=1mA$			0.1	V
$V_{G1,2}$ (High)		$V_{DD} > 16V$	13	14.5	16	V
		$V_{DD} < 16V$		$V_{DD}-2.2$		
Turn-Off Threshold ( $V_{S1,2}-V_{D1,2}$ )				-30		mV
Turn-Off Propagation Delay		$V_{D1,2}=V_{SS}$		15		ns

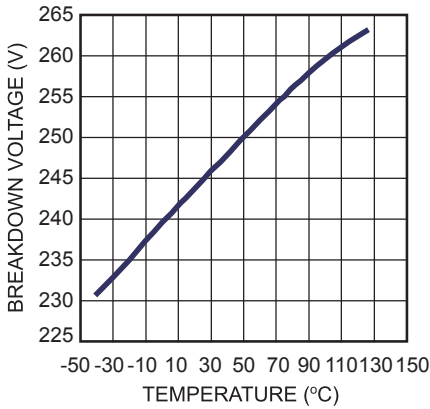
**ELECTRICAL CHARACTERISTICS (continued)**
 $V_{DD} = 12V, -40^{\circ}C \leq T_J \leq 125^{\circ}C$ , unless otherwise noted.

Parameter	Symbol	Conditions	Min	Typ	Max	Units
Turn-Off Total Delay	$T_{Doff}$	$V_{D1,2} = V_{SS}, C_{LOAD} = 5nF,$ $R_{GATE} = 0\Omega$		50	75	ns
Pull-Down Impedance Pull-Down Current		$V_{D1,2} = V_{SS}, C_{LOAD} = 10nF,$ $R_{GATE} = 0\Omega$		50	75	ns
				1	2	$\Omega$
		$3V < V_{G1,2} < 10V$		3		A

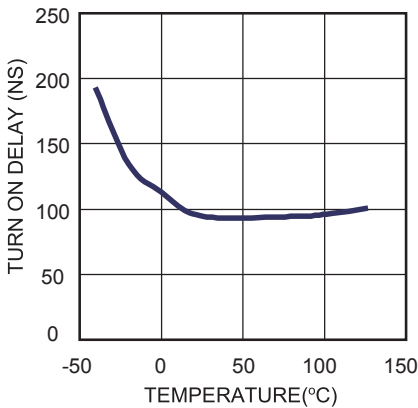
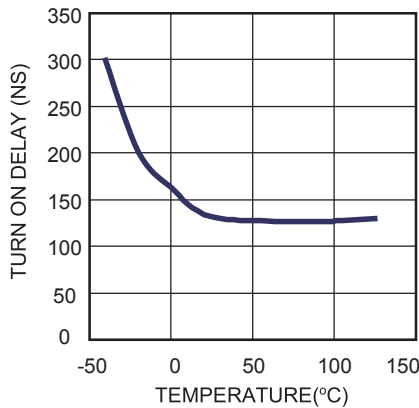
**PIN FUNCTIONS**

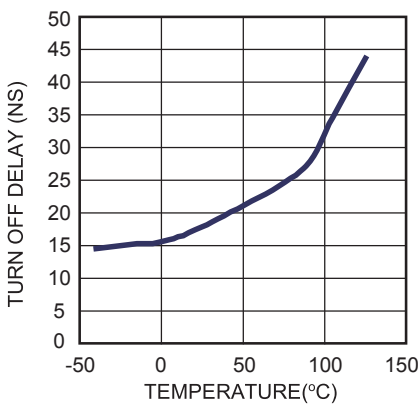
Pin # (SOIC8)	Pin # (SOIC8E)	Pin # (SOIC14)	Name	Description
1	1	2	V <sub>G2</sub>	MOSFET 2 Gate Driver Output.
3	2	3	EN	Enable Pin. Enables the internal IC logic when the EN pin voltage exceeds the EN shutdown threshold; the gate driver remains latched until the EN voltage exceeds the EN UVLO threshold.
4	3	6	V <sub>D2</sub>	MOSFET 2 Drain Voltage Sense.
-	4	7	V <sub>S2</sub>	Source pin used as reference for V <sub>D2</sub> .
-	5	8	V <sub>S1</sub>	Source pin used as reference for V <sub>D1</sub> .
6	6	9	V <sub>D1</sub>	MOSFET 1 Drain Voltage Sense.
7	7	12	V <sub>DD</sub>	Supply Voltage.
8	8	13	V <sub>G1</sub>	MOSFET 1 Gate Driver Output.
2	EXPOSED PAD	1,14	PGND	Power Ground. Power switch return.
-	-	5,10	NC	No Connection.
-	-	4	LL	Light-Load–Timing Set. Connect a resistor to set the light-load timing.
-	-	11	RCP	Reverse Current Protection. Internal 5V reference.
5	-	-	V <sub>SS</sub>	Common Source. Used as reference for both channels.

## TYPICAL CHARACTERISTICS

 **$V_{D1,2}$  Breakdown Voltage vs. Temperature**

 **$V_{DD}$  UVLO Rising vs. Temperature**

**Shutdown Current ( $V_{DD}=20V$ ) vs. Temperature**

**Turn On Delay (Load=5nF) vs. Temperature**

**Turn On Delay (Load=10nF) vs. Temperature**

**Turn Off Delay (Load=5nF) vs. Temperature**

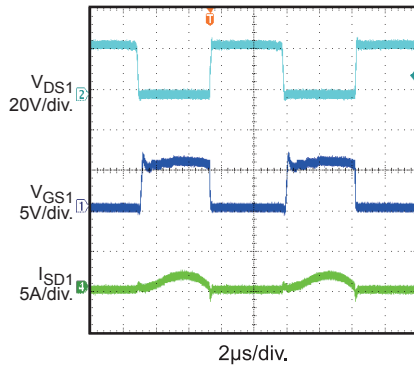
**Turn Off Delay (Load=10nF) vs. Temperature**

 **$V_{S1,2}-V_{D1,2}$  Forward Voltage vs. Temperature**


**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

$V_{DD} = 12V$ , unless otherwise noted.

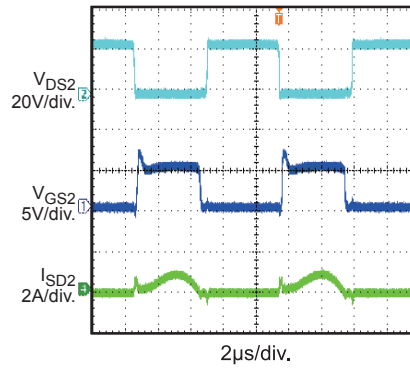
**Operation in 90W  
LLC Converter**

$V_{IN} = 240V_{AC}$ ,  $V_{OUT} = 12V$ ,  $I_{OUT} = 0.75A$



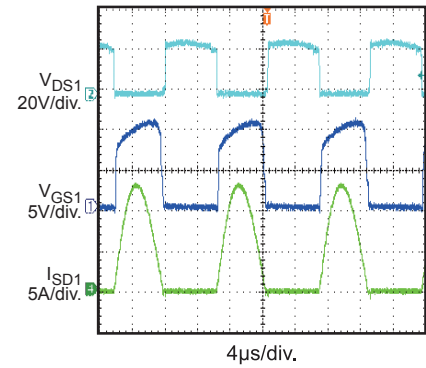
**Operation in 90W  
LLC Converter**

$V_{IN} = 240V_{AC}$ ,  $V_{OUT} = 12V$ ,  $I_{OUT} = 0.75A$



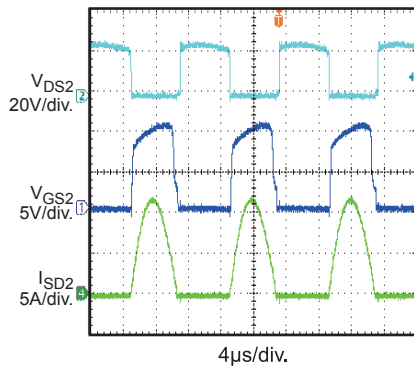
**Operation in 90W  
LLC Converter**

$V_{IN} = 240V_{AC}$ ,  $V_{OUT} = 12V$ ,  $I_{OUT} = 7.5A$



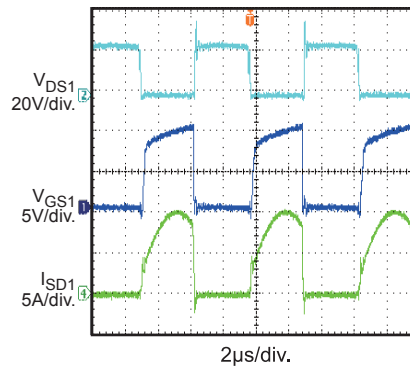
**Operation in 90W  
LLC Converter**

$V_{IN} = 240V_{AC}$ ,  $V_{OUT} = 12V$ ,  $I_{OUT} = 7.5A$



**Operation in 90W  
LLC Converter**

$V_{IN} = 265V_{AC}$ ,  $V_{OUT} = 12V$ ,  $I_{OUT} = 7.5A$





BLOCK DIAGRAM



Figure 1: Functional Block Diagram

## OPERATION

The MP6922A operates in discontinuous-conduction mode (DCM), continuous-conduction mode (CCM), and critical conduction mode (CrCM) condition. Operating in either DCM or CrCM, the control circuitry controls the gate in forward mode; it turns the gate off when the MOSFET current is low. In CCM, the control circuitry turns off the gate during very fast transients.

### Blanking

The control circuitry contains a blanking function that ensure that when the MOSFET turns ON/OFF, the MOSFET remains in that state for  $\sim 1\mu\text{s}$ , which determines the minimum ON-time. During the turn-on blanking period, the turn-off threshold is not totally blanked, but changes to  $\sim +100\text{mV}$  (instead of  $+30\text{mV}$ ). This ensures that the part can always turn OFF even during the turn-on blanking period (albeit slower, so avoid setting the synchronous period to less than  $1\mu\text{s}$  at CCM condition in the LLC converter to eliminate shoot-through).

### VD Clamp

The MP6922A uses a high-voltage JFET at its input because  $V_{D1,2}$  can go as high as  $180\text{V}$ . Connect a small resistor between the  $V_{D1,2}$  pin and the external MOSFET drain to avoid excessive currents when  $V_G$  goes below  $-0.7\text{V}$ .

### Under-Voltage Lockout (UVLO)

When  $V_{DD}$  goes below the UVLO threshold, the part enters sleep mode and a  $10\text{k}\Omega$  resistor pulls down  $V_G$ .

### Enable Pin

If EN is pulled low, the part enters sleep mode.

### Thermal Shutdown

If the junction temperature of the IC exceeds  $150^\circ\text{C}$ ,  $V_G$  is pulled low and the part stops switching. The part resumes normal operation after the junction temperature has dropped to  $120^\circ\text{C}$ .

### Turn-On Phase

$V_{DS}$  ( $V_D - V_{SS}$ ) goes negative ( $< -500\text{mV}$ ) when the switch current flows through the MOSFET's body diode. If  $V_{DS}$  is much lower than the turn-on threshold of the control circuitry ( $-30\text{mV}$ ), then

the MOSFET turns on after about  $200\text{ns}$  turn-on delay (Figure 2).

Triggering the turn-on threshold ( $-30\text{mV}$ ) causes the circuit to add a blanking time (minimum on-time  $1\mu\text{s}$ ), during which the turn-off threshold changes from  $+30\text{V}$  to  $+100\text{mV}$ . This blanking time avoids erroneous triggering caused by ringing on the synchronous power switch.

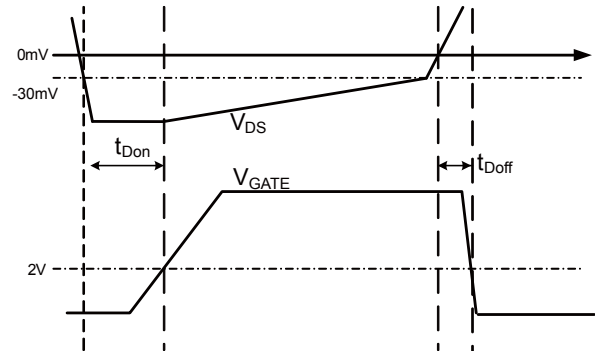


Figure 2: Turn-On and Turn-Off delay

### Conducting Phase

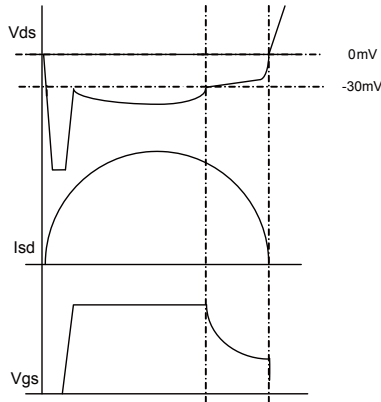
When the MOSFET turns ON,  $V_{DS}$  ( $-I_{SD} \times R_{DS(ON)}$ ) rise relative to the switch current ( $I_{SD}$ ) drop. When  $V_{DS}$  rises above the turn-on threshold ( $-30\text{mV}$ ), the control circuitry stops pulling up the gate driver and the MOSFET driver voltage drops, which increases the MOSFET's  $R_{DS(ON)}$ . This adjusts  $V_{DS}$  ( $-I_{SD} \times R_{DS(ON)}$ ) to around  $-30\text{mV}$  even when the switch current  $I_{SD}$  is fairly small, and can prevent the internal driver from triggering until the current through the MOSFET has dropped to near zero.

### Turn-Off Phase

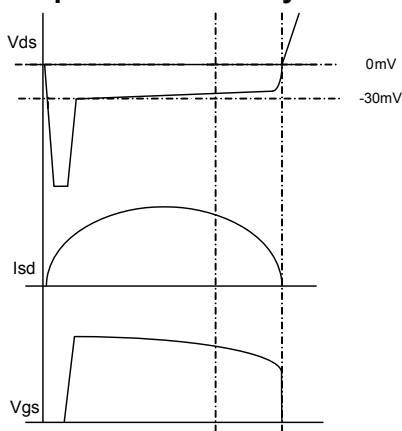
When  $V_{DS}$  rises to trigger the turn-off threshold ( $30\text{mV}$ ), the control circuitry pulls down the driver switch voltage after a  $20\text{ns}$  turn-off delay (shown in Figure 2). Similarly, a  $1.6\mu\text{s}$  blanking time occurs after the switch turns off, during which the MOSFET does not turn on to avoid erroneous triggering.

Figure 3 shows the MP6922A operating under a heavy-load. The high current initially saturates the driver voltage. After  $V_{DS}$  goes above  $-30\text{mV}$ , the driver voltage decreases to adjust the  $V_{DS}$  to around  $-30\text{mV}$ .

Figure 4 shows the MP6922A operating at light-load. The low current prevents the driver voltage from saturating but decreases when the synchronous power switch turns on and adjusts  $V_{DS}$ .



**Figure 3: Synchronous Rectification Operation at Heavy Load**



**Figure 4: Synchronous Rectification Operation at Light Load**

#### Light-Load Latch-Off Function

The gate driver of MP6922A is latched to limit driver losses under light-load condition to improve light-load efficiency.

#### Normal Operation Latch Off

When the MOSFET's switching-cycle conducting period falls below  $2.2\mu\text{s}$  ( $\tau_{LL}$ ), the MP6922A enters light-load mode and latches off the MOSFET after a  $160\mu\text{s}$  delay (light-load-enter delay,  $\tau_{LL-Delay}$ )

After entering light-load mode, the MP6922A monitors the MOSFET's body diode conducting period by sensing  $V_{DS}$  (when  $V_{DS}$  exceeds  $-300\text{mV}$  ( $V_{LL-DS}$ ), MP6922A treats the the MOSFET's body-diode conducting period as finished). If the MOSFET's body diode conducting period exceeds  $2.4\mu\text{s}$  ( $\tau_{LL} + \tau_{LL-H}$ ), light-load mode ends and the MOSFET unlatches to restart the synchronous rectification.

For the SOIC14 package, the MP6922A has an LL pin that allows  $\tau_{LL}$  to be adjusted by an external resistor:

$$\tau_{LL} = 2.2\mu\text{s} \cdot \frac{R_{LL}}{100\text{k}\Omega}$$

#### Latch Off during Burst Operation

The IC also monitors the synchronous MOSFET OFF period. If the OFF period exceeds the light-load-enter OFF period width ( $\tau_{LL-OFF}$ ), the MP6922A enters light-load mode and latches off the gate driver.

The gate driver is unlatched when the drain-source voltage of the synchronous MOSFET  $V_{DS}$  drops below  $-30\text{mV}$ .

#### Reverse Current-Protection Function

When the LLC system operates in CCM at a very high frequency, the synchronous current may reverse before the IC turns off the gate driver, which can lead to shoot-through (in center-tapped outputs with full-wave rectification topologies). The MP6922A has a protection function to latch off the gate driver when the current reverses before the driver signal is pulled low.

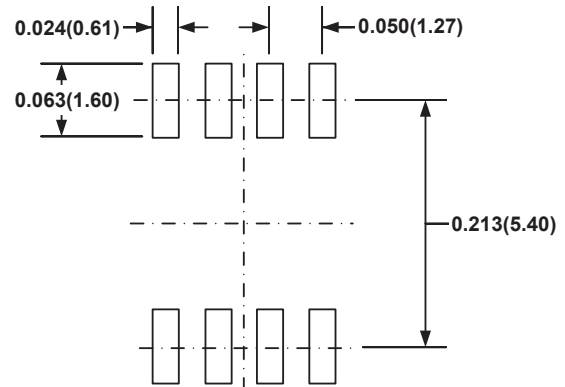
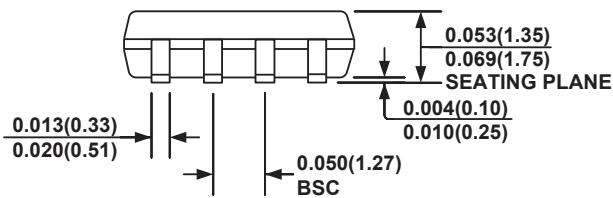
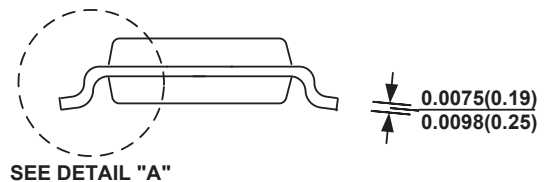
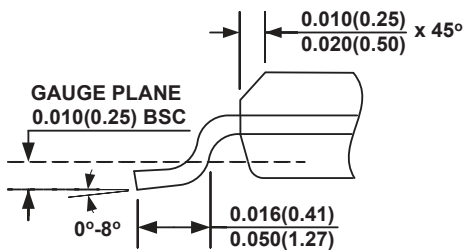
When the synchronous current reverses, the high spike can be observed between the MOSFET's drain/source. The MP6922A monitors the voltage through the RCP pin using a voltage divider. When the voltage of RCP pin exceeds  $V_{RCP}$ , MP6922A latches the driver signal of both channels for  $\sim 150\mu\text{s}$  ( $\tau_{RCP}$ ) to protect the synchronous MOSFET. At the end of  $\tau_{RCP}$ , MP6922A restarts the synchronous rectification.

TYPICAL APPLICATION CIRCUIT



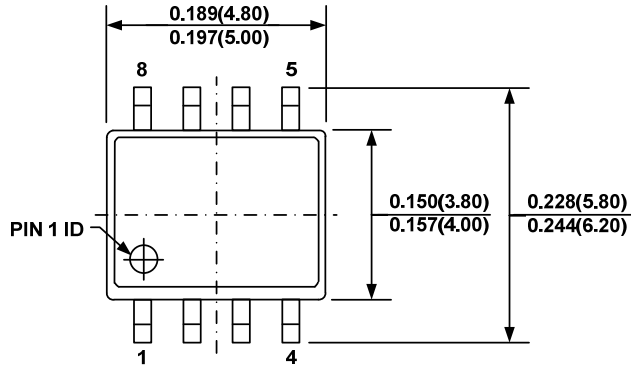
Figure 5—Synchronous Rectification in LLC with MP6922A

**PACKAGE INFORMATION**
**SOIC8**

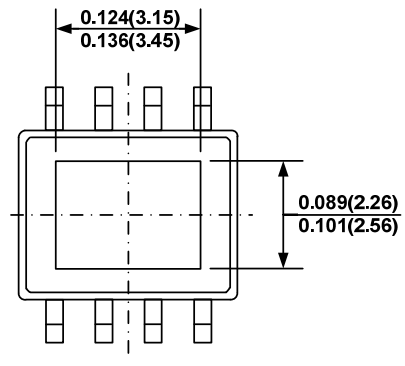
**TOP VIEW**

**RECOMMENDED LAND PATTERN**

**FRONT VIEW**

**SIDE VIEW**

**DETAIL "A"**
**NOTE:**

- 1) CONTROL DIMENSION IS IN INCHES. DIMENSION IN BRACKET IS IN MILLIMETERS.
- 2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
- 3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
- 4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.004" INCHES MAX.
- 5) DRAWING CONFORMS TO JEDEC MS-012, VARIATION AA.
- 6) DRAWING IS NOT TO SCALE.

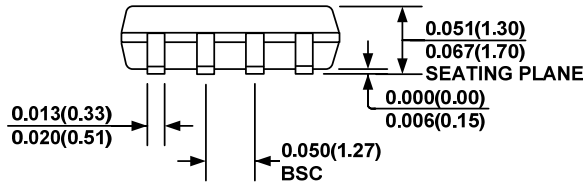
**SOIC8E (EXPOSED PAD)**



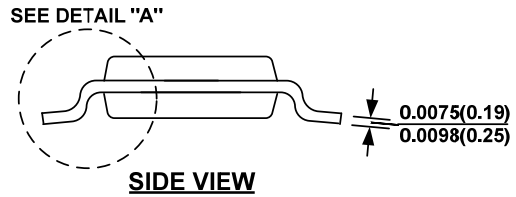
**TOP VIEW**



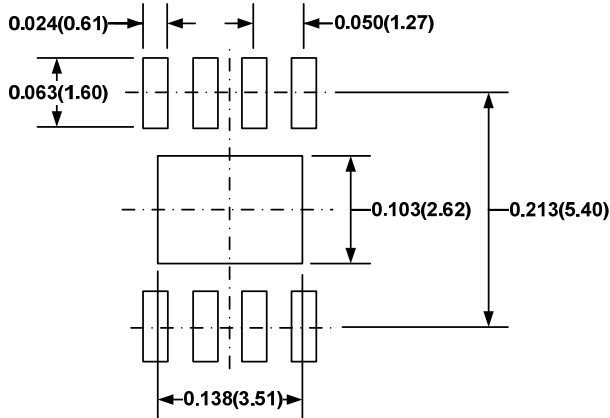
**BOTTOM VIEW**



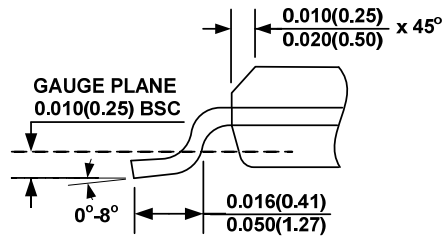
**FRONT VIEW**



**SIDE VIEW**



**RECOMMENDED LAND PATTERN**



**DETAIL "A"**

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