**Document Category: Product Specification**

## *UltraCMOS® RF Digital Step Attenuator, 9 kHz–55 GHz*

# **Features**

- Wideband support up to 55 GHz
- Glitch-safe attenuation state transitions
- Flexible attenuation steps of 0.5 dB and 1 dB up to 31.5 dB
- Extended +105 °C operating temperature
- Parallel and serial programming interfaces with serial addressability
- Flip-chip die

# **Applications**

- Test and measurement (T&M)
- Point-to-point communication systems
- Very small aperture terminals (VSAT)



#### *Figure 1 • PE43508 Functional Diagram*



## **Product Description**

The PE43508 is a 50Ω, HaRP™ technology-enhanced, 6-bit RF digital step attenuator (DSA) that supports a wide frequency range from 9k to 55 GHz. The PE43508 features glitch-safe attenuation state transitions, supports 1.8V control voltage and optional  $V_{SS,ext}$  bypass mode to improve spurious performance, making this device ideal for test and measurement, point-to-point communication systems, and very small aperture terminals (VSAT).

The PE43508 provides an integrated digital control interface that supports both serial addressable and parallel programming of the attenuation. The PE43508 covers a 31.5 dB attenuation range in 0.5 dB and 1 dB steps. It is capable of maintaining 0.5 dB and 1 dB monotonicity through 55 GHz. In addition, no external blocking capacitors are required if 0 VDC is present on the RF ports.

The PE43508 is manufactured on pSemi's UltraCMOS® process, a patented variation of silicon-on-insulator (SOI) technology.



pSemi's HaRP technology enhancements deliver high linearity and excellent harmonics performance. It is an innovative feature of the UltraCMOS process, offering the performance of GaAs with the economy and integration of conventional CMOS.

## **Optional External V<sub>ss</sub> Control**

For proper operation, the  $V_{SS,EXT}$  control pin must be grounded or tied to the  $V_{SS}$  voltage specified in [Table 2](#page-2-0). When the  $V_{SS,EXT}$  control pin is grounded, FETs in the switch are biased with an internal negative voltage generator. For applications that require the lowest possible spur performance,  $V_{SS,EXT}$  can be applied externally to bypass the internal negative voltage generator.

## **Absolute Maximum Ratings**

Exceeding absolute maximum ratings listed in **[Table 1](#page-1-1)** may cause permanent damage. Operation should be restricted to the limits in **[Table 2](#page-2-0)**. Operation between operating range maximum and absolute maximum for extended periods may reduce reliability.

#### **ESD Precautions**

When handling this UltraCMOS device, observe the same precautions as with any other ESD-sensitive devices. Although this device contains circuitry to protect it from damage due to ESD, precautions should be taken to avoid exceeding the rating specified in **[Table 1](#page-1-1)**.

#### **Latch-up Immunity**

Unlike conventional CMOS devices, UltraCMOS devices are immune to latch-up.

<span id="page-1-1"></span>*Table 1 • Absolute Maximum Ratings for the PE43508*

<span id="page-1-0"></span>

<b>Parameter/Condition</b>	<b>Min</b>	<b>Max</b>	<b>Unit</b>
Positive supply voltage, V <sub>DD</sub>	$-0.3$	5.5	V
Negative supply voltage, V <sub>SS EXT</sub>	$-3.6$	0.3	V
Digital input voltage	$-0.3$	3.6	v
Peak RF input power, $50\Omega$		Fig. 7	dBm
Maximum junction temperature		$+150$	$^{\circ}C$
Storage temperature range	$-65$	$+150$	$^{\circ}C$
ESD voltage HBM, all pins $(*)$		1000	v
Note: * Human body model (MIL-STD 883 Method 3015)			



# **Recommended Operating Conditions**

**[Table 2](#page-2-0)** lists the recommending operating condition for the PE43508. Devices should not be operated outside the recommended operating conditions listed below.

<span id="page-2-0"></span>*Table 2 • Recommended Operating Condition for the PE43508* 

<b>Parameter</b>	<b>Min</b>	<b>Typ</b>	<b>Max</b>	<b>Unit</b>
Normal mode, $V_{SS\_EXT} = 0V^{(1)}$				
Positive supply voltage, V <sub>DD</sub>	2.3	3.3	5.5	$\vee$
Positive supply current, $I_{DD}^{(3)}$		170	250	μA
Bypass mode, $V_{SS\_EXT} = -3.0V^{(2)}$				
Positive supply voltage, $V_{DD}$ ( $V_{DD} \geq 3.4V$ . See Table 3 for full spec compliance.)	3.1	3.4	5.5	$\vee$
Positive supply current, $I_{DD}^{(3)}$		122	166	μA
Negative supply voltage, V <sub>SS_EXT</sub>	$-3.3$	$-3.0$	$-2.7$	$\vee$
Negative supply current, I <sub>SS</sub>	$-40$	$-16$		μA
Normal or bypass mode				
Digital input high	1.17		3.60	$\vee$
Digital input low	$-0.3$		0.6	$\vee$
Digital input current <sup>(4)</sup>		10	20	μA
RF input power, CW <sup>(5)</sup> (7)			Fig. 7	dBm
RF input power, pulsed <sup>(6) (7)</sup>			Fig. 7	dB <sub>m</sub>
Operating temperature range	$-40$	$+25$	$+105$	$^{\circ}C$

<span id="page-2-2"></span>**Notes:** 

1) Normal mode: Connect V<sub>SS</sub>  $_{\text{EXT}}$  (pin 15) to GND (V<sub>SS EXT</sub> = 0V) to enable internal negative voltage generator.

<span id="page-2-3"></span>2) Bypass mode: Use  $V_{SS}$   $EXT$  (pin 15) to bypass and disable internal negative voltage generator.

<span id="page-2-6"></span>3) Due to startup inrush current, a minimum current limit of 600 µA is allowed for normal operation of the DSA.

<span id="page-2-1"></span>4) Applies to all pins except pins 10, 12, 19 and 20. P/S (pin 10), A0/D4 (pin 12), A2/D6 (pin 19) and A1/D5 (pin 20) have internal 1.5 MΩ pull-up resistor to internal 1.8V  $V_{DD}$ .

<span id="page-2-7"></span>5) 100% duty cycle, all bands, 50Ω.

<span id="page-2-4"></span>6) ≤ 5% duty cycle, 50Ω.

<span id="page-2-5"></span>7) The maximum peak envelope of any OFDM complex waveform signal, such as CP-OFDM, should not exceed the maximum peak RF input power in **[Table 1](#page-1-1)**. The maximum average power of any complex waveform should not exceed the operating maximum RF input power, CW.



# **Electrical Specifications**

[Table 3](#page-3-0) provides the PE43508 key electrical specifications at 25 °C,  $Z_S = Z_L = 50\Omega$ , unless otherwise specified. Normal mode<sup>[\(1\)](#page-4-0)</sup> is at  $V_{DD}$  = 3.3V and  $V_{SS\_EXT}$  = 0V. Bypass mode<sup>[\(2\)](#page-2-3)</sup> is at  $V_{DD}$  = 3.4V and  $V_{SS\_EXT}$  = –3.0V.

#### <span id="page-3-0"></span>*Table 3 • PE43508 Electrical Specifications*





#### *Table 3 • PE43508 Electrical Specifications (Cont.)*



**Notes:** 

<span id="page-4-0"></span>1) Normal mode: Connect  $V_{SS\_EXT}$  (pin 15) to GND ( $V_{SS\_EXT}$  = 0V) to enable internal negative voltage generator.

2) Bypass mode: Use  $V_{SS\_EXT}$  (pin 15) to bypass and disable internal negative voltage generator.

<span id="page-4-1"></span>3) The input 1 dB compression point is a linearity figure of merit. Refer to **[Table 2](#page-2-0)** for the operating RF input power (50Ω).

<span id="page-4-2"></span>4) Limited by the frequency range capabilities of the test systems.

![](_page_5_Picture_1.jpeg)

## **Switching Frequency**

The PE43508 has a maximum 400 kHz switching rate in normal mode (pin 15 tied to ground). A faster switching rate is available in bypass mode (pin 15 tied to  $V_{SS\text{EXT}}$ ). The rate at which the PE43508 can be switched is then limited to the switching time as specified in **[Table 3](#page-3-0)**.

Switching frequency is defined to be the speed at which the DSA can be toggled across attenuation states. Switching time is the time duration between the point the control signal reaches 50% of the final value and the point the output signal reaches within 10% or 90% of its target value.

## **Spur-free Performance**

The PE43508 spur fundamental occurs around 4 MHz. Its typical spurious performance in normal mode is –168 dBm/Hz (pin 15 tied to ground), with 30 kHz bandwidth. If spur-free performance is desired, the internal negative voltage generator can be disabled by applying a negative voltage to  $V_{SS,ext}$  (pin 15).

## **Glitch-safe Attenuation State**

The PE43508 features a novel architecture to provide safe transition behavior when changing attenuation states. When RF input power is applied, positive output power spikes are prevented during attenuation state changes by optimized internal timing control.

## **Truth Tables**

**Table 4**–**[Table 6](#page-5-0)** provide the truth tables for the PE43508.

### *Table 4 • Parallel Truth Table*

![](_page_5_Picture_339.jpeg)

### *Table 4 • Parallel Truth Table (Cont.)*

![](_page_5_Picture_340.jpeg)

### <span id="page-5-1"></span>*Table 5 • Serial Address Word Truth Table*

	<b>Address</b>							
$\overline{A7}$ (MSB)			A6   A5   A4   A3   A2   A1				A <sub>0</sub> (LSB)	<b>Setting</b>
L	$\mathbf{L}$	L	L	L	L	L	L	000
L	L	L	L	L	L	L	н	001
L	L	L	L	L	L	н	L	010
$\mathbf{L}$	L	L	L	L	L	н	н	011
L	L	L	L	L	Н	L	L	100
L	L	L	L	L	н	L	н	101
L	L	L	L	L	н	н	L	110
$\mathbf{L}$	L	L	L	L	н	н	н	111

<span id="page-5-0"></span>*Table 6 • Serial Attenuation Word Truth Table*

![](_page_5_Picture_341.jpeg)

![](_page_6_Picture_0.jpeg)

# **Serial Addressable Register Map**

**[Figure 2](#page-6-0)** provides the serial addressable register map for the PE43508.

#### <span id="page-6-0"></span>*Figure 2 • Serial Addressable Register Map*

![](_page_6_Picture_93.jpeg)

The attenuation word is derived directly from the value of the attenuation state. To find the attenuation word, multiply the value of the state by four, then convert to binary.

For example, to program the 18.5 dB state at address 3:

 $4 \times 18.5 = 74$  $74 \rightarrow 01001010$ 

Address Word: 00000011 Attenuation Word: 01001010 Serial Input: 00000011**01001010** 

# **Programming Options**

### **Parallel/Serial Selection**

Either a parallel or serial addressable interface can be used to control the PE43508. The  $\overline{P}/S$  bit provides this selection, with  $\overline{P/S}$  = LOW selecting the parallel interface and  $\overline{P/S}$  = HIGH selecting the serial interface. The P/S pin has an internal tie HIGH (namely, the pin is internally tied to the 1.8V  $V_{DD}$ ), so if this is left floating, the part defaults to serial mode. If there is a need to put this part in parallel mode, a LOW logic should be applied to this pin.

### **Parallel Mode Interface**

The parallel interface consists of six CMOScompatible control lines that select the desired attenuation state, as shown in **Table 4**.

The parallel interface timing requirements are defined by **[Figure 4](#page-10-0)** (Latched-Parallel/Direct-Parallel Timing Diagram), **[Table 10](#page-10-1)** (Parallel and Direct Interface AC Characteristics) and switching time (**[Table 3](#page-3-0)**).

For latched parallel programming, the latch enable (LE) should be held LOW while changing attenuation state control values then pulse LE HIGH to LOW (per **[Figure 4](#page-10-0)**) to latch new attenuation state into the device.

For direct parallel programming, the LE line should be pulled HIGH. Changing attenuation state control values changes the device state to new attenuation. Direct mode is ideal for manual control of the device (using hardwire, switches, or jumpers).

### **Serial-Addressable Interface**

The serial-addressable interface is a 16-bit serial-in, parallel-out shift register buffered by a transparent latch. The 16-bits make up two words comprised of 8 bits each. The first word is the attenuation word, which controls the state of the DSA. The second word is the address word, which is compared to the static (or programmed) logical states of the A0, A1 and A2 digital inputs. If there is an address match, the DSA changes state; otherwise its current state remains unchanged. **[Figure 3](#page-8-0)** illustrates an example timing diagram for programming a state.

The serial-addressable interface is controlled using three CMOS-compatible signals: SDI, CLK, and LE. The SDI and CLK inputs allow data to be serially

![](_page_7_Picture_12.jpeg)

entered into the shift register. Serial data is clocked in LSB first. The serial interface data output, SDO, outputs serial input data delayed by 16 clock cycles to control the cascaded attenuator using a single serial peripheral interface (SPI) bus.

The shift register must be loaded while LE is held LOW to prevent the attenuator value from changing as data is entered. The LE input should then be toggled HIGH and brought LOW again, latching the new data into the DSA. The Address Word truth table is listed in **[Table 5](#page-5-1)**. The address pins A0 (pin 12), A1 (pin 20), and A2 (pin 19) can either be grounded logic LOW or left floating (logic HIGH due to internal pull-up to 1.8V  $V_{DD}$ ) depending upon what fixed address the user wants the DSA to be set at.The Attenuation Word truth table is listed in **[Table 6](#page-5-0)**. A programming example of the serial register is illustrated in **[Figure 2](#page-6-0)**. The serial timing diagram is illustrated in **[Figure 3](#page-8-0)**.

### **Power-up Control Settings**

The PE43508 always initializes to the maximum attenuation setting (31.5 dB) on power-up for both the serial addressable and latched parallel modes of operation (as long as the LE pin is logic LOW during start up) and it remains in this setting until the user latches in the next programming word.

**In direct parallel mode** ( $\overline{P/S}$  = LOW and logic HIGH present on the LE pin during the power-up), the DSA can be preset to any state within the 31.5 dB range by pre-setting the parallel control pins D[6:1] prior to power-up. In this mode, there is a 4 µs delay between the time the DSA is powered-up to the time the desired state is set. If the control pins are left floating in this mode during power-up, the device defaults to the 28dB attenuation setting.

**In latched parallel mode** ( $\overline{P/S}$  = LOW), if the LE pin is kept LOW during power-up, the part should default to maximum attenuation state (31.5dB). Logic LOW should be present on the LE pin during power-up and then logic HIGH should be written on the LE pin when the user wants to program the part. If the LE is kept floating during power-up, the part should default to maximum attenuation state (31.5dB).

**In serial mode** ( $\overline{P/S}$  = HIGH or left floating) logic HIGH on the LE pin during the power up: The part should default to minimum attenuation state

![](_page_8_Picture_0.jpeg)

(reference state). But a logic LOW on the LE pin during the power up, the part should default to maximum attenuation state.

Dynamic operation between serial and parallel programming modes is not supported.

If the DSA powers up in serial mode  $(\overline{P/S} = HIGH)$ , prior to toggling to parallel mode, the user must

![](_page_8_Picture_264.jpeg)

ensure that the pins LE, SDI/D1, CLK/D2, A0/D4, A1/ D5, and A2/D6 are set to logic LOW.

If the DSA powers up in either latched or direct parallel mode, the pins LE, SDI/D1, CLK/D2, A0/D4, A1/D5, and A2/D6 must be set to logic LOW and the pin SDO/D3 set to high impedance prior to toggling to serial addressable mode  $(\overline{P/S} = HIGH)$ .

![](_page_8_Picture_265.jpeg)

<span id="page-8-0"></span>![](_page_8_Figure_9.jpeg)

![](_page_8_Figure_10.jpeg)

Notes:

- 1. SPI mode 0:
	- SDI data is captured on the CLK's rising edge - SDO data is valid on CLK falling edge
- 2. CLK shared pin with 1 dB parallel control bit D2
- 3. SDI shared pin with 0.5 dB parallel control bit D1
- 4. SDO shared pin with 2 dB parallel control bit D3
- 5. A0 shared pin with 4 dB parallel control bit D4
- 6. A1 shared pin with 8 dB parallel control bit D5
- 7. A2 shared pin with 16 dB parallel control bit D6
- 8. Serial data bits D[7], D[0], and A[7:3] must be set to logic low

 $9. X =$  Undefined

![](_page_9_Picture_1.jpeg)

### *Table 8 • Serial Interface AC Characteristics***[\(1\)](#page-9-0)**

![](_page_9_Picture_129.jpeg)

#### <span id="page-9-1"></span><span id="page-9-0"></span>*Table 9 • Latch and Clock Specifications*

![](_page_9_Picture_130.jpeg)

![](_page_10_Picture_0.jpeg)

#### <span id="page-10-0"></span>*Figure 4 • Latched-Parallel/Direct-Parallel Timing Diagram*

![](_page_10_Figure_3.jpeg)

#### <span id="page-10-1"></span>*Table 10 • Parallel and Direct Interface AC Characteristics***[\(\\*\)](#page-10-2)**

<span id="page-10-2"></span>![](_page_10_Picture_150.jpeg)

![](_page_11_Picture_1.jpeg)

The following example shows a scenario where two DSAs are connected in series.

<span id="page-11-0"></span>![](_page_11_Figure_3.jpeg)

![](_page_11_Figure_4.jpeg)

The following table provides the complete SDI and SDO content for the example shown in **[Figure 5](#page-11-0)**.

*Figure 6 • Serial Addressable Cascaded Devices Table*

	D <sub>[0]</sub>		21 I DI 31	DI 41	1   D[51   D[61   D[71	ITAIOI	דו A ונ		1   A[21   A[31   A[41   ,		I AI 511		$I A$ [6] $A$ [7] $I$					D[0] <b> </b> D[1] <b> </b> D[2] <b> </b> D[3] <b> </b> D[4]	DI51	<b>HD</b> I61"	71	LAIOLL	AI1		A[2] A[3]	. LAI4'			
I CLK								10	---	ᅭ	ᅩ	14	15	16	17	18	19	20	$\mathbf{a}$ ᅩ	$\sim$ ᅩ	วว دے	24	$\sim$ 25	۷b	77	28	29	20 υU	$\sim$ ـ ت
SDI																													
SDO			$\sqrt{2}$						$\lambda$	$\lambda$			$\lambda$				U												

![](_page_12_Picture_0.jpeg)

![](_page_12_Picture_1.jpeg)

#### <span id="page-12-0"></span>*Figure 7 • Power De-rating Curve, 9 kHz–55GHz, 50* Ω

![](_page_12_Figure_3.jpeg)

![](_page_13_Picture_1.jpeg)

# **Typical Performance Data**

[Figure 8](#page-13-0)–[Figure 36](#page-17-0) show the typical performance data at 25 °C and V<sub>DD</sub> = 3.3V (Z<sub>S</sub> = Z<sub>L</sub> = 50 $\Omega$ ), unless otherwise specified.

#### <span id="page-13-0"></span>*Figure 8 • Insertion Loss vs. Temperature*

![](_page_13_Figure_5.jpeg)

*Figure 10 • RF1 Return Loss (All Attenuation States) vs. Attenuation Setting*

![](_page_13_Figure_7.jpeg)

*Figure 12 • RF2 Return Loss (All Attenuation States) vs. Attenuation Setting*

![](_page_13_Figure_9.jpeg)

*Figure 9 • RF1 Return Loss (Major Attenuation States) vs. Attenuation Setting*

![](_page_13_Figure_11.jpeg)

#### *Figure 11 • RF2 Return Loss (Major Attenuation States) vs. Attenuation Setting*

![](_page_13_Figure_13.jpeg)

#### *Figure 13 • RF1 Return Loss for Reference State vs. Temperature*

![](_page_13_Figure_15.jpeg)

![](_page_14_Picture_0.jpeg)

#### *Figure 14 • RF1 Return Loss for 31.5 dB Attenuation Setting vs. Temperature*

![](_page_14_Figure_3.jpeg)

*Figure 16 • RF2 Return Loss for 31.5 dB Attenuation Setting vs. Temperature*

![](_page_14_Figure_5.jpeg)

*Figure 18 • Relative Phase Error for 31.5 dB Attenuation Setting vs. Frequency*

![](_page_14_Figure_7.jpeg)

#### *Figure 15 • RF2 Return Loss for Reference State vs. Temperature*

![](_page_14_Figure_9.jpeg)

#### *Figure 17 • Relative Phase Error vs. Attenuation Setting*

![](_page_14_Figure_11.jpeg)

#### *Figure 19 • Attenuation Error @ 6 GHz vs. Temperature*

![](_page_14_Figure_13.jpeg)

![](_page_15_Picture_1.jpeg)

*Figure 20 • Attenuation Error @ 8 GHz vs. Temperature*

![](_page_15_Figure_3.jpeg)

*Figure 22 • Attenuation Error @ 26.5 GHz vs. Temperature*

![](_page_15_Figure_5.jpeg)

*Figure 24 • Attenuation Error @ 50 GHz vs. Temperature*

![](_page_15_Figure_7.jpeg)

*Figure 21 • Attenuation Error @ 13 GHz vs. Temperature*

![](_page_15_Figure_9.jpeg)

*Figure 23 • Attenuation Error @ 45 GHz vs. Temperature*

![](_page_15_Figure_11.jpeg)

![](_page_15_Figure_12.jpeg)

![](_page_15_Figure_13.jpeg)

![](_page_16_Picture_0.jpeg)

#### *Figure 26 • 0.5 dB Step Attenuation vs. Frequency***[\(1\)](#page-16-0)**

![](_page_16_Figure_3.jpeg)

*Figure 28 • 0.5 dB Step, Actual vs. Frequency*

![](_page_16_Figure_5.jpeg)

*Figure 30 • Major State Bit Error vs. Attenuation Setting*

![](_page_16_Figure_7.jpeg)

*Figure 27 • 1 dB Step Attenuation vs. Frequency***[\(2\)](#page-16-1)**

![](_page_16_Figure_9.jpeg)

*Figure 29 • 1 dB Step, Actual vs. Frequency*

![](_page_16_Figure_11.jpeg)

*Figure 31 • 0.5 dB Attenuation Error vs. Frequency*

![](_page_16_Figure_13.jpeg)

<span id="page-16-0"></span>1) Monotonicity is held so long as step attenuation does not cross below -0.5 dB.

<span id="page-16-1"></span>2) Monotonicity is held so long as step attenuation does not cross below -1.0 dB.

![](_page_17_Picture_1.jpeg)

#### <span id="page-17-1"></span>*Figure 32 • 1 dB Attenuation Error vs. Frequency*

![](_page_17_Figure_3.jpeg)

*Figure 34 • IIP3 vs. Attenuation Setting*

![](_page_17_Figure_5.jpeg)

<span id="page-17-0"></span>*Figure 36 • Attenuation Transient (24-23.5 dB)*

![](_page_17_Figure_7.jpeg)

*Figure 33 • IIP2 vs. Attenuation Setting*

![](_page_17_Figure_9.jpeg)

*Figure 35 • Attenuation Transient (23.5-24 dB)*

![](_page_17_Figure_11.jpeg)

![](_page_18_Picture_1.jpeg)

# **Evaluation Setup**

The PE43508 s-parameter data up to 67 GHz (**[Table 3](#page-3-0)** and **[Figure 8](#page-13-0)**–**[Figure 32](#page-17-1)**) were taken using either coplanar waveguide with ground (CPWG) or grounded co-planar waveguide (GCPW) on an alumina substrate (**[Figure 37](#page-18-0)**) and RF probes.

The PE43508 input 1dB compression point below 40 GHz, input IP2 and IP3 measurements up to 16 GHz, settling time and switching time (**[Table 3](#page-3-0)**) were taken on a mmWave PCB using 2.92 mm connectors.

<span id="page-18-0"></span>*Figure 37 • Alumina Substrate Board Used in the PE43508 Evaluation*

![](_page_18_Figure_6.jpeg)

![](_page_19_Picture_1.jpeg)

# **Evaluation Kit**

There are two types of evaluation kits, CPWG or GCPW on an alumina substrate (**[Figure 37](#page-18-0)**) and a mmWave PCB using 2.92 mm connectors (**[Figure 38](#page-19-0)**), designed to ease customer evaluation of the PE43508 digital step attenuator. For more information on the mmWave PCB evaluation kit, see the *PE43508 Evaluation Kit (EVK) User's Manual*. For more information on the alumina substrate evaluation kit, see Application Note 78, *PE43508 55 GHz DSA Substrate Carrier Assembly Guide*.

<span id="page-19-0"></span>![](_page_19_Figure_4.jpeg)

![](_page_19_Figure_5.jpeg)

![](_page_19_Figure_6.jpeg)

![](_page_20_Picture_0.jpeg)

# **Pin Configuration**

This section provides pin information for the PE43508. **[Figure 39](#page-20-0)** shows the pin configuration of this device. **[Table 11](#page-20-1)** provides a description for each pin.

<span id="page-20-0"></span>*Figure 39 • Pin Configuration (Bumps Up) for the PE43508* 

![](_page_20_Figure_5.jpeg)

#### <span id="page-20-1"></span>*Table 11 • Pin Descriptions for the PE43508*

![](_page_20_Picture_360.jpeg)

**Notes:** 

- <span id="page-20-2"></span>1) RF pins 7 and 17 must be at 0 VDC. The RF pins do not require DC blocking capacitors for proper operation if the 0 VDC requirement is met.
- <span id="page-20-4"></span>2) P/S (pin 10), A0/D4 (pin 12), A2/D6 (pin 19) and A1/D5 (pin 20) have internal 1.5 MΩ pull-up resistor to internal 1.8V V<sub>DD</sub>. These pins will have an internal logic HIGH on them if they are left floating by the user. In serial mode, the user can leave the P/S pin floating and the part will default to serial mode.
- <span id="page-20-3"></span>3) Use  $V_{SS-EXT}$  (pin 15) to bypass and disable internal negative voltage generator. Connect  $V_{SS\_EXT}$  (pin 15) to GND ( $V_{SS\_EXT}$  = 0V) to enable internal negative voltage generator.

![](_page_21_Picture_1.jpeg)

# **Die Mechanical Specifications**

This section provides the die mechanical specifications for the PE43508.

#### *Table 12 • Mechanical Specifications forthe PE43508*

![](_page_21_Picture_84.jpeg)

![](_page_22_Picture_0.jpeg)

#### *Table 13 • Pin Coordinates for the PE43508*

![](_page_22_Picture_288.jpeg)

center of the pin.

#### *Figure 40 • Pin Layout for the PE43508***[\(1\)](#page-22-1)[\(2\)](#page-22-0)**

![](_page_22_Figure_6.jpeg)

**Notes:** 

<span id="page-22-1"></span>1) Drawings are not drawn to scale.

<span id="page-22-0"></span>2) Singulated die size shown, bump side up.

![](_page_23_Picture_1.jpeg)

# **Tape and Reel Specification**

This section provides the tape and reel specification for the PE43508.

#### *Figure 41 • Tape and Reel Specifications for the PE43508*

![](_page_23_Figure_5.jpeg)

## **Storage Recommendation**

As a best practice, devices that will be stored for a long period of time (> 1 week) should be kept in a dry nitrogen environment.