



TSH345

Triple video buffer with selectable filter for HD and SD video applications

Features

- Selectable 6th order filtering of 36 MHz, 18 MHz and 9 MHz
- 5 V single-supply operation
- Internal input DC level shifter
- No input capacitor required
- 3 matched 6 dB amplifiers
- AC or DC output-coupled
- Very low harmonic distortion
- Specified for 150 Ω loads
- Data min. and max. are tested during production

Applications

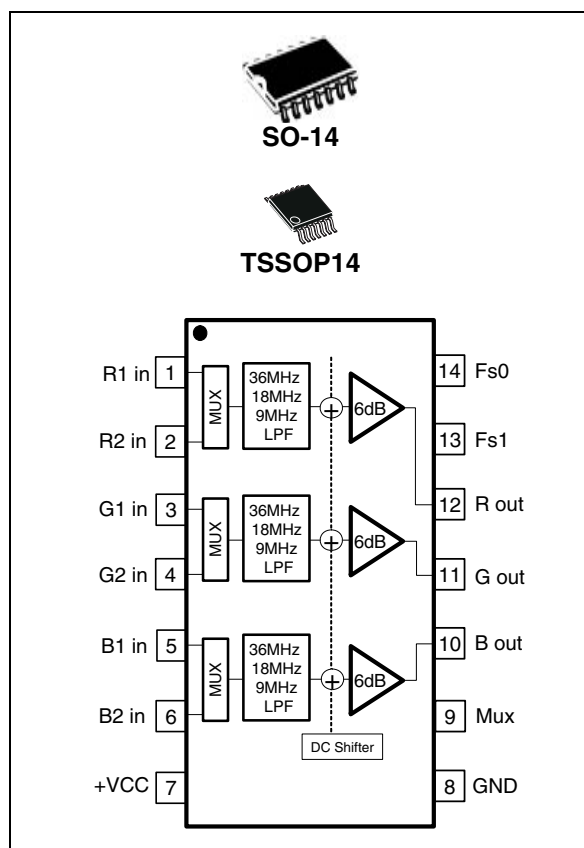
- High-end video systems
- High definition TV (HDTV)
- Broadcast and graphic video
- Multimedia products

Description

The TSH345 is a triple single-supply video buffer featuring an internal gain of 6 dB and selectable filtering for HD and SD video outputs on 75 Ω video lines. The TSH345 is ideal to drive YC, CVBS, YUV, YPbPr or RGB signals from video DAC outputs.

The main advantage of this circuit is its input DC level shifter. It allows driving video signals on 75 Ω video lines without damaging the synchronization tip and without input or output capacitors when using a single 5 V power supply. The DC level shifter is internally fixed and optimized to keep the output video signals between low and high output rails in the best position for the greatest linearity.

The TSH345 is available in SO-14 and TSSOP-14 plastic packages for optimum space saving.



1 Absolute maximum ratings and operating conditions

Table 1. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{CC}	Supply voltage ⁽¹⁾	6	V
V_{in}	Input voltage range	2.5	V
T_{oper}	Operating free air temperature range	-40 to +85	°C
T_{stg}	Storage temperature	-65 to +150	°C
T_j	Maximum junction temperature	150	°C
R_{thjc}	Thermal resistance junction to case		
	SO-14 TSSOP14	22 32	°C/W
R_{thja}	Thermal resistance junction to ambient area		
	SO-14 TSSOP14	125 110	°C/W
P_{max}	Maximum power dissipation (at $T_{amb} = 25^\circ\text{C}$) for $T_j = 150^\circ\text{C}$		
	SO-14 TSSOP14	1 1.1	W
ESD	CDM: charged device model	250	V
	HBM: human body model	2	kV
	MM: machine model	100	V

1. All voltage values, except differential voltage, are with respect to network terminal.

Table 2. Operating conditions

Symbol	Parameter	Value	Unit
V_{CC}	Power supply voltage	4.5 to 5.5 ⁽¹⁾	V

1. Tested in full production with +5 V single power supply.

2 Electrical characteristics

Table 3. Electrical characteristics at $V_{CC} = +5\text{ V}$ single supply, $T_{amb} = 25^\circ\text{C}$ (unless otherwise specified)

Symbol	Test conditions	Min.	Typ.	Max.	Unit
DC performance					
V_{DC}	Output DC shift $R_L = 150\ \Omega$, T_{amb} $-40^\circ\text{C} < T_{amb} < +85^\circ\text{C}$	100	240 310	440	mV
I_{ib}	Input bias current T_{amb} , input to GND $-40^\circ\text{C} < T_{amb} < +85^\circ\text{C}$		1.3 1.4	3.6	μA
R_{in}	Input resistance, T_{amb}		1		$\text{M}\Omega$
C_{in}	Input capacitance, T_{amb}		0.1		pF
I_{CC}	Total supply current (3 x operators) No load, input to GND $-40^\circ\text{C} < T_{amb} < +85^\circ\text{C}$		44.6 45	51.6	mA
G	DC voltage gain $R_L = 150\ \Omega$, $V_{in} = 1.4\text{V}$ $-40^\circ\text{C} < T_{amb} < +85^\circ\text{C}$	1.96	2 1.96	2.05	V/V
Output characteristics					
V_{OH}	High level output voltage $R_L = 150\ \Omega$ $-40^\circ\text{C} < T_{amb} < +85^\circ\text{C}$	3.4	3.9 3.8		V
V_{OL}	Low level output voltage $R_L = 150\ \Omega$		47		mV
I_{out}	I_{source} T_{amb} $-40^\circ\text{C} < T_{amb} < +85^\circ\text{C}$	76	100 91		mA
	I_{sink} $-40^\circ\text{C} < T_{amb} < +85^\circ\text{C}$	106	134 126		mA
Filtering					
Standard definition	Bandwidth F1 selected, small signal, $V_{ICM} = 0.5\text{ V}$, $R_L = 150\ \Omega$ -3 dB bandwidth -1 dB bandwidth		9 5.7		MHz
	Attenuation F1 selected/F=27 MHz, small signal, $V_{ICM} = 0.5\text{ V}$, $R_L = 150\ \Omega$	40	45		dB

Table 3. Electrical characteristics at $V_{CC} = +5\text{ V}$ single supply, $T_{amb} = 25^\circ\text{C}$ (unless otherwise specified) (continued)

Symbol	Test conditions	Min.	Typ.	Max.	Unit
Standard definition with progressive scanning	Bandwidth F2 selected, small signal, $V_{ICM} = 0.5\text{ V}$, $R_L = 150\ \Omega$ -3 dB bandwidth -1 dB bandwidth	13	21 18		MHz
	Attenuation F2 selected/ $F = 54\text{ MHz}$, small signal, $V_{ICM} = 0.5\text{ V}$, $R_L = 150\ \Omega$	32	38		dB
High definition	Bandwidth F3 selected, small signal, $V_{ICM} = 0.5\text{ V}$, $R_L = 150\ \Omega$ -3 dB bandwidth -1 dB bandwidth	25	36 32		MHz
	Attenuation F3 selected/ $F = 74.25\text{ MHz}$, small signal, $V_{ICM} = 0.5\text{ V}$, $R_L = 150\ \Omega$	25	32		dB
D	Delay between each channel		0.5		ns
gd	Group delay variation F1 selected/ $F = 0$ to 6 MHz		11		ns
Δg	Differential gain F1 selected/ $F = 6\text{ MHz}$, $R_L = 150\ \Omega$		0.38		%
$\Delta\Phi$	Differential phase F1 selected/ $F = 6\text{ MHz}$, $R_L = 150\ \Omega$		0.5		°
Noise					
eN	Total input voltage noise in Standard Definition $F = 100\text{ kHz}$, $R_{IN} = 50\ \Omega$		74		nV/ $\sqrt{\text{Hz}}$
	Total input voltage noise in High Definition $F = 100\text{ kHz}$, $R_{IN} = 50\ \Omega$		86		
Standby mode					
I_{STBY}	Total current consumption in standby mode $Fs1 = 1$, $Fs0 = 1$ T_{amb} $-40^\circ\text{ C} < T_{amb} < +85^\circ\text{ C}$		440 480	690	μA
T_{on}	Time from standby to active mode		5		μs
T_{off}	Time from active to standby mode		5		μs
Fs1, Fs0 and Mux features					
V_{high}	High level	0.9			V
V_{low}	Low level			0.3	V

Table 4. Filter and standby settings, $V_{CC} = +5\text{ V}$ single supply, $T_{amb} = 25^\circ\text{C}$

Fs1⁽¹⁾	Fs0⁽¹⁾	Settings	
0	0	F3	Filtering for high definition (HD)
0	1	F2	Filtering for progressive video (PV)
1	0	F3	Filtering for standard definition (SD)
1	1	Standby	TSH345 in standby mode

1. Fs1 and Fs0 pins must never be left floating.

Table 5. Mux settings, $V_{CC} = +5\text{ V}$ single supply, $T_{amb} = 25^\circ\text{C}$

Mux⁽¹⁾	Settings	
0	R1 G1 B1	Video1 selected
1	R2 G2 B2	Video2 selected

1. The MUX pin must never be left floating.

Figure 1. Filtering

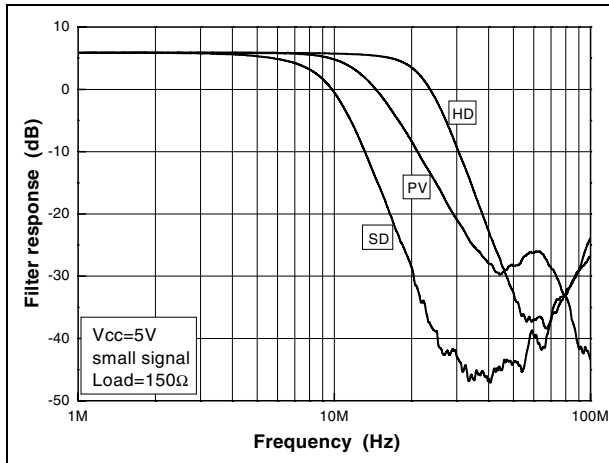


Figure 2. Gain flatness

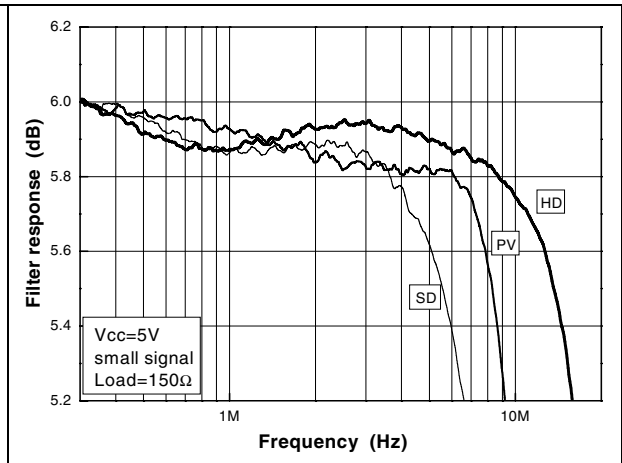


Figure 3. Distortion 1 MHz (HD)

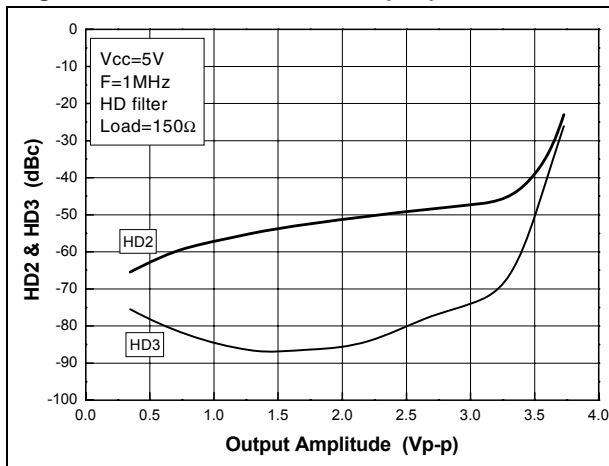


Figure 4. Distortion 10 MHz (HD)

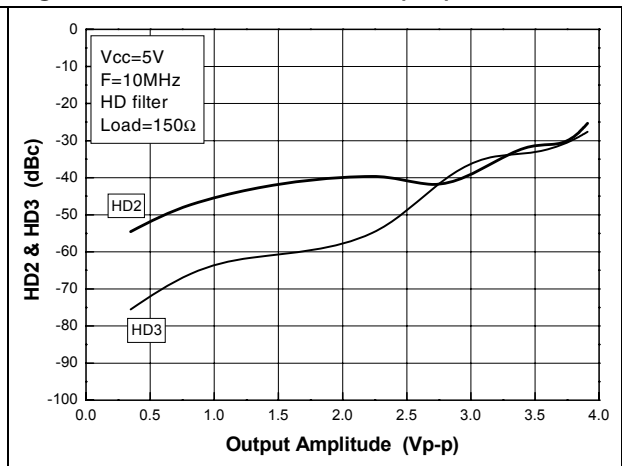


Figure 5. Distortion 1 MHz (PV)

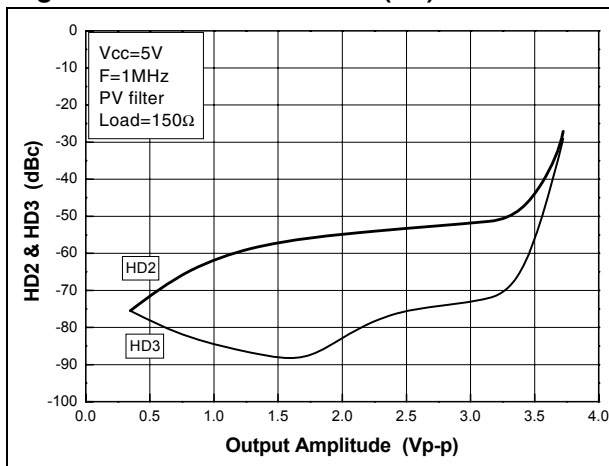


Figure 6. Distortion 10 MHz (PV)

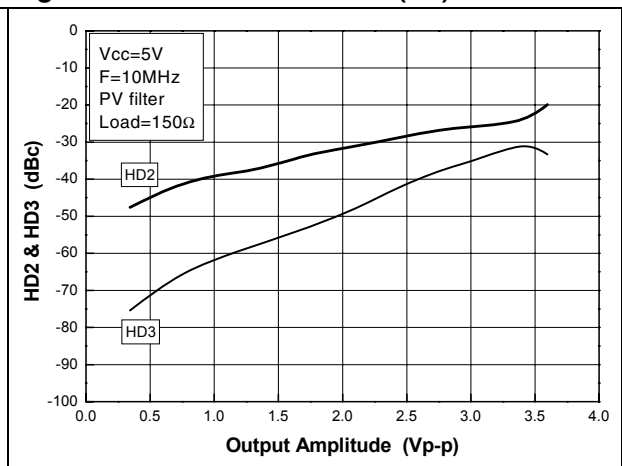


Figure 7. Distortion 1 MHz (SD filter)

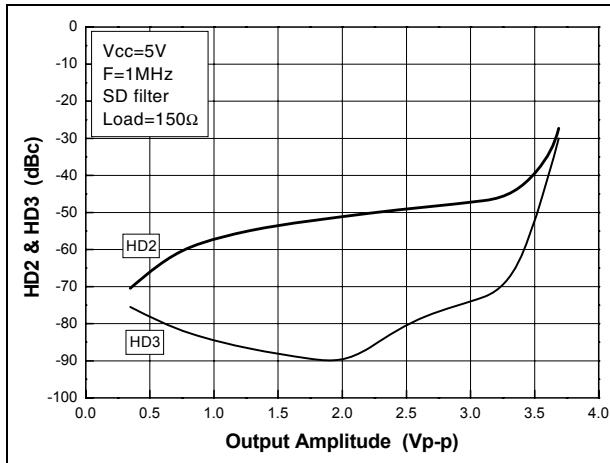


Figure 8. Input noise vs. frequency

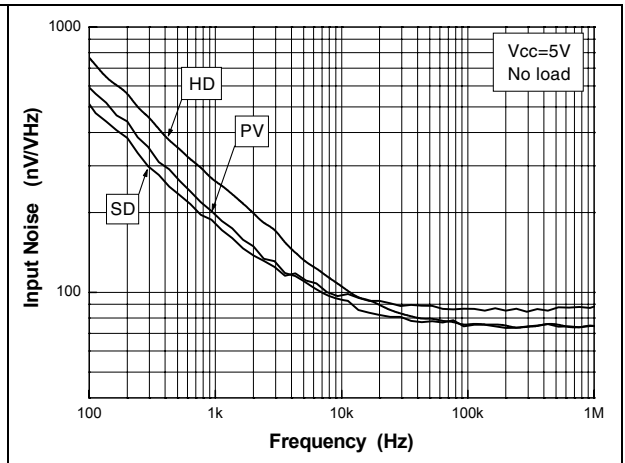


Figure 9. Gain vs. input amplitude

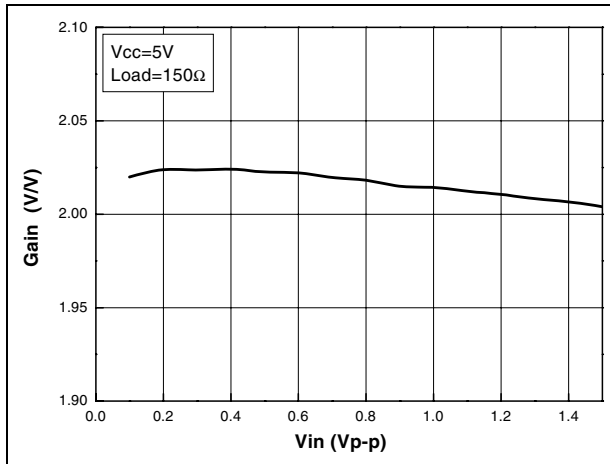


Figure 10. Channel crosstalk vs. frequency

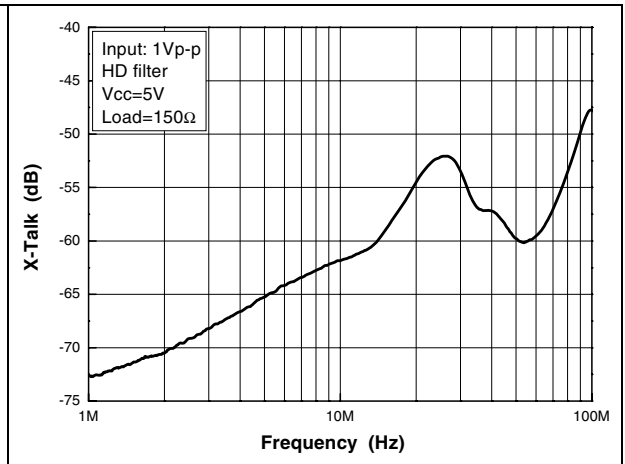


Figure 11. Output vs input amplitude

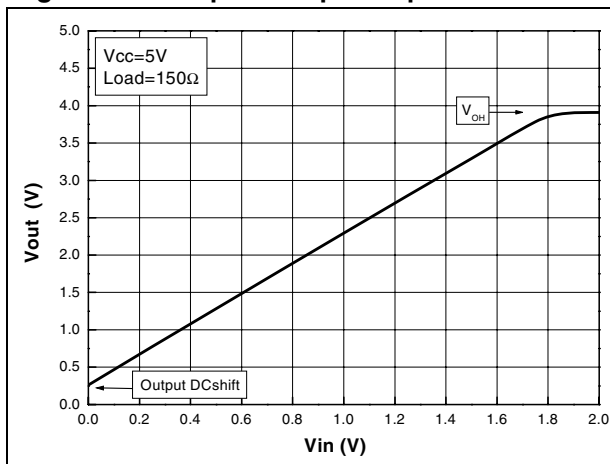


Figure 12. MUX isolation

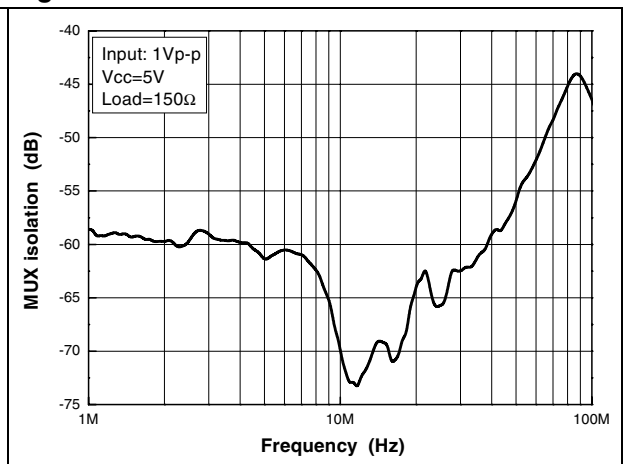


Figure 13. Current consumption vs. supply

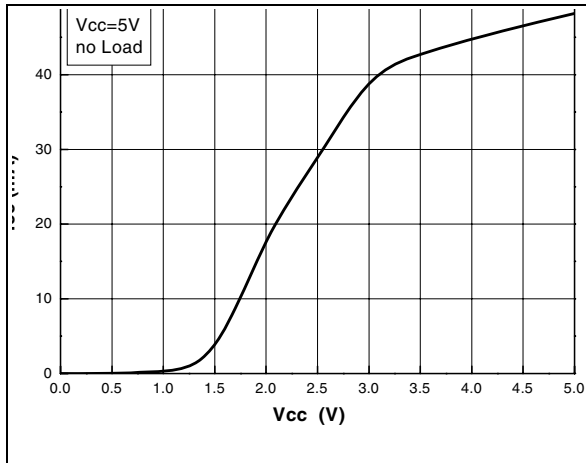


Figure 14. Supply current vs. temperature

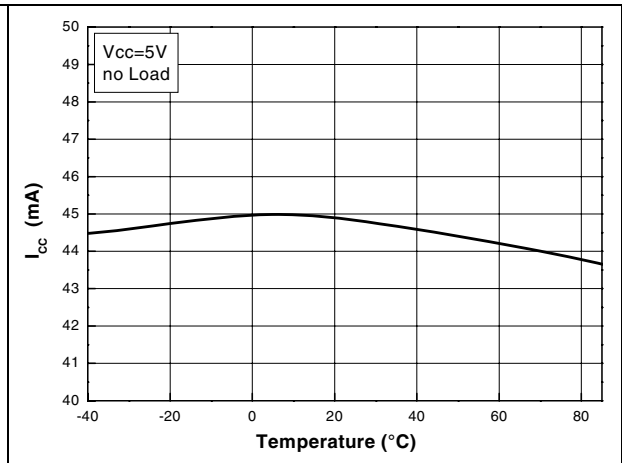


Figure 15. Filtering vs. temperature

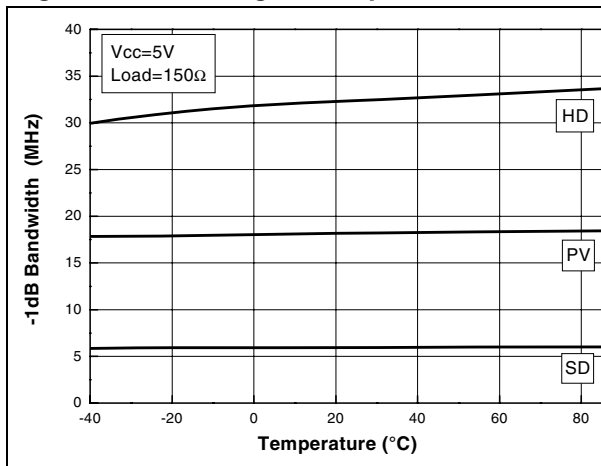


Figure 16. Filter attenuation vs. temperature

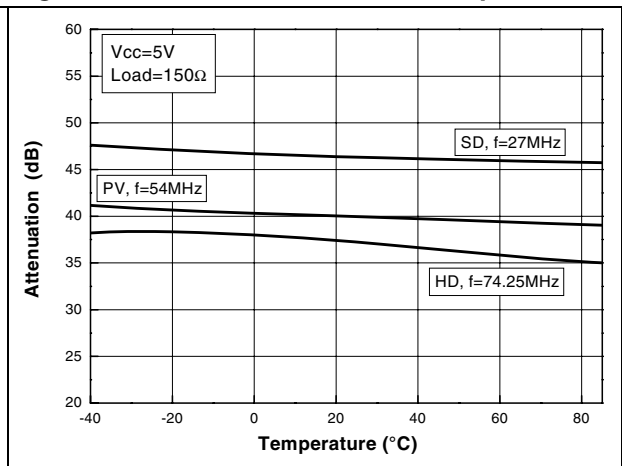


Figure 17. Gain matching vs. temperature

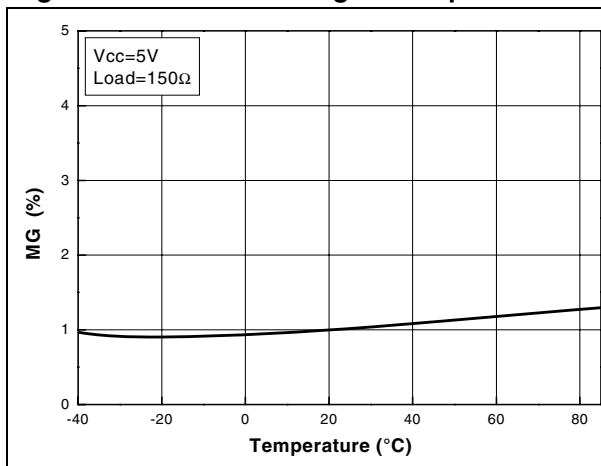


Figure 18. Output DC shift vs. temperature

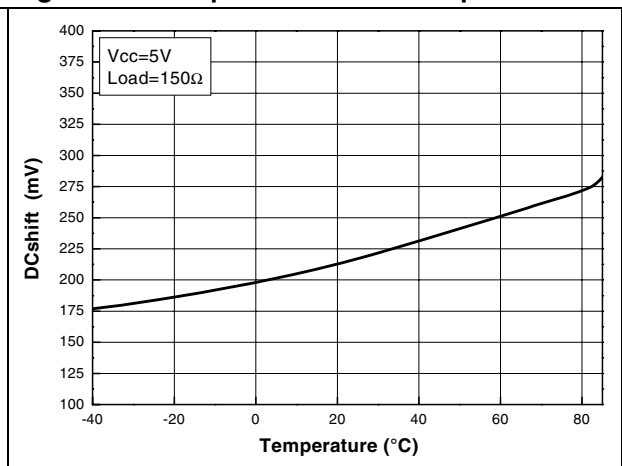


Figure 19. Standby current vs. temperature

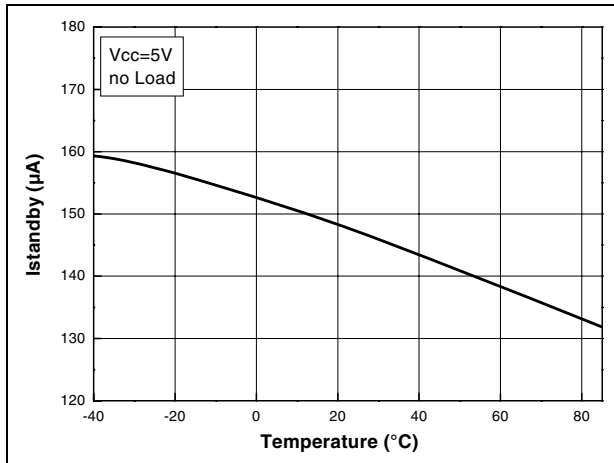


Figure 20. Isink vs. temperature

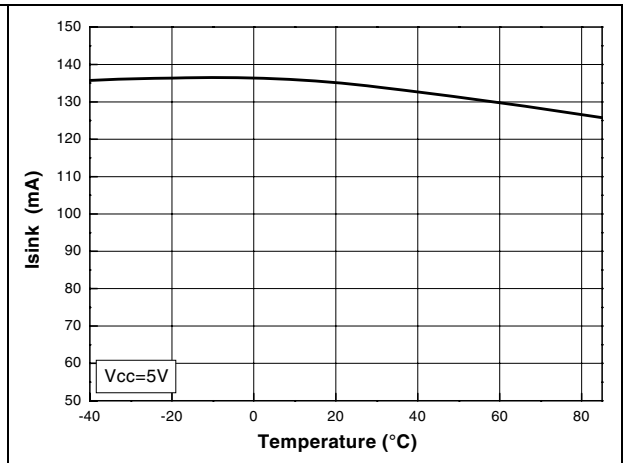


Figure 21. Isource vs. temperature

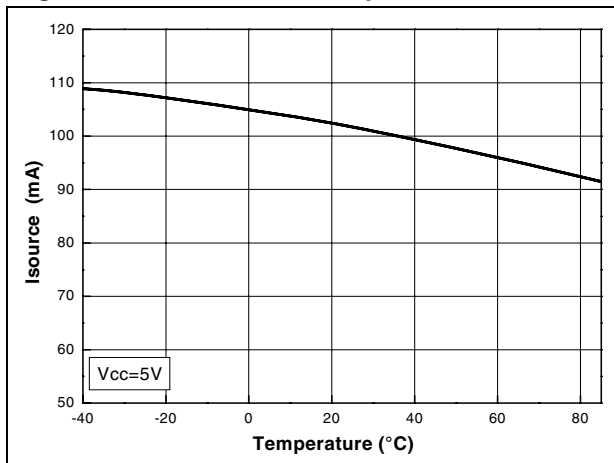


Figure 22. Ibias vs. temperature

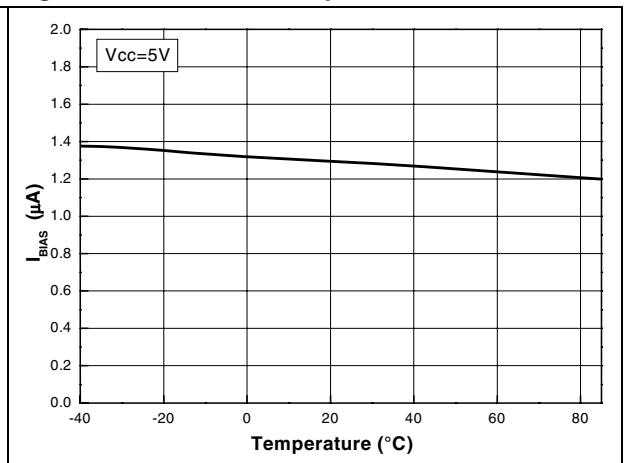


Figure 23. VOL vs. temperature

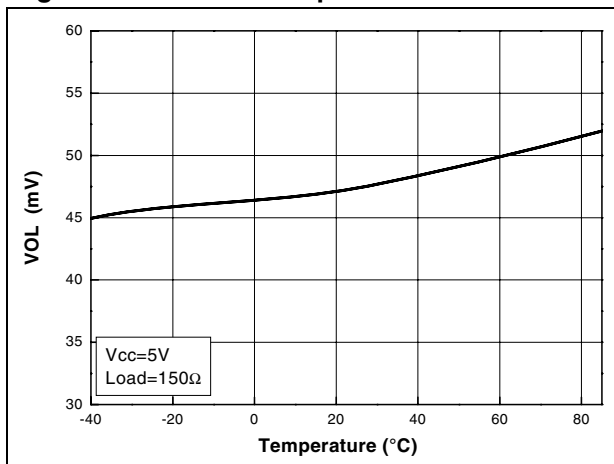


Figure 24. VOH vs. temperature

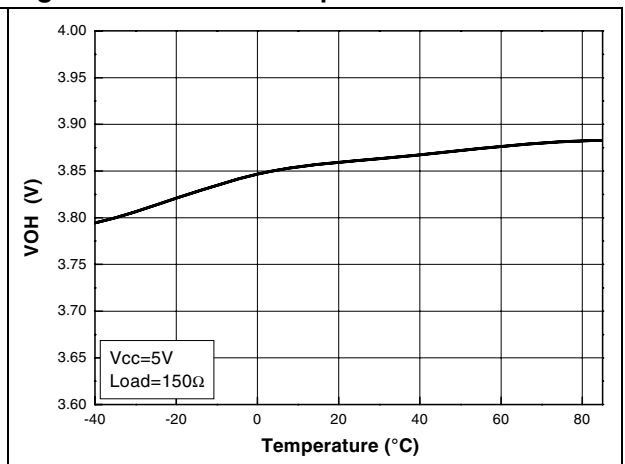
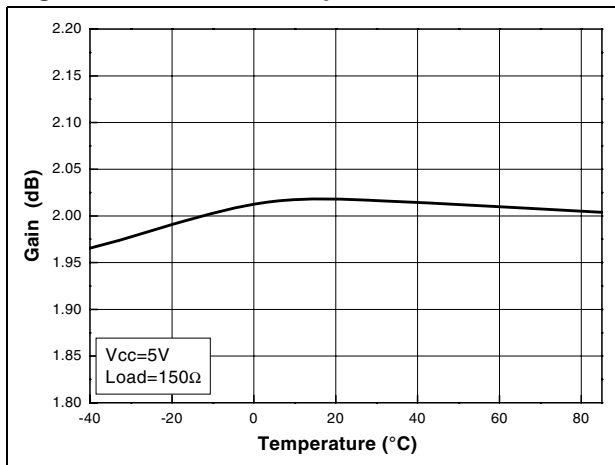


Figure 25. Gain vs. temperature



2.1 Power supply considerations: improving the power supply noise rejection

Correct power supply bypassing is very important to optimize performance in low- and high-frequency ranges. Bypass capacitors should be placed as close as possible to the IC pin (pin 4) to improve high-frequency bypassing. A capacitor (C_{LF}) greater than $10\ \mu\text{F}$ is necessary to improve the PSRR in low frequencies. For better quality bypassing, you can add a capacitor of $100\ \text{nF}$ (C_{HF}). C_{HF} must be placed as close as possible to the IC pin to improve the noise supply rejection in the higher frequencies. A coil can be added in order to better reject the noise from the supply and to prevent current peaks as much as possible.

Figure 26. Circuit for power supply bypassing

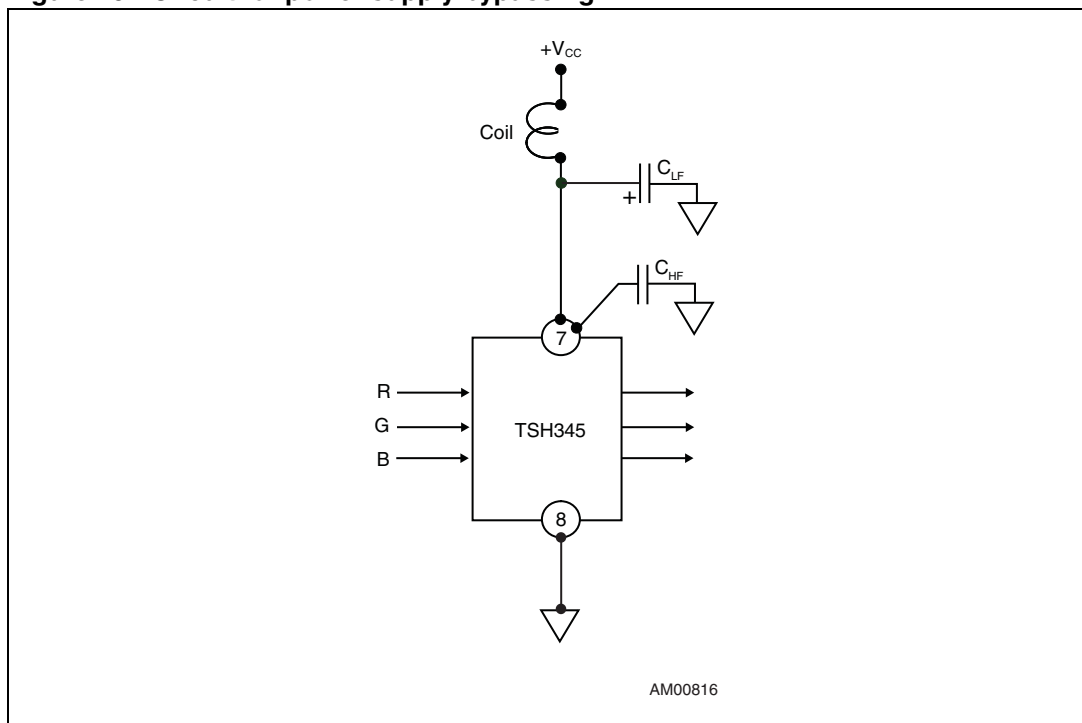


Figure 27. Circuit for noise rejection improvement measurement

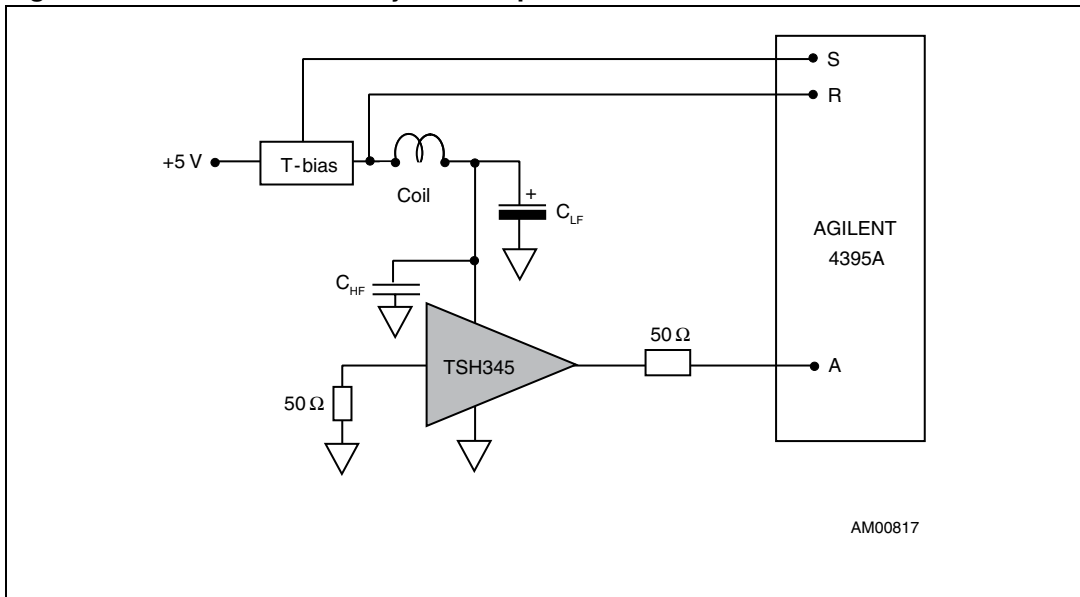
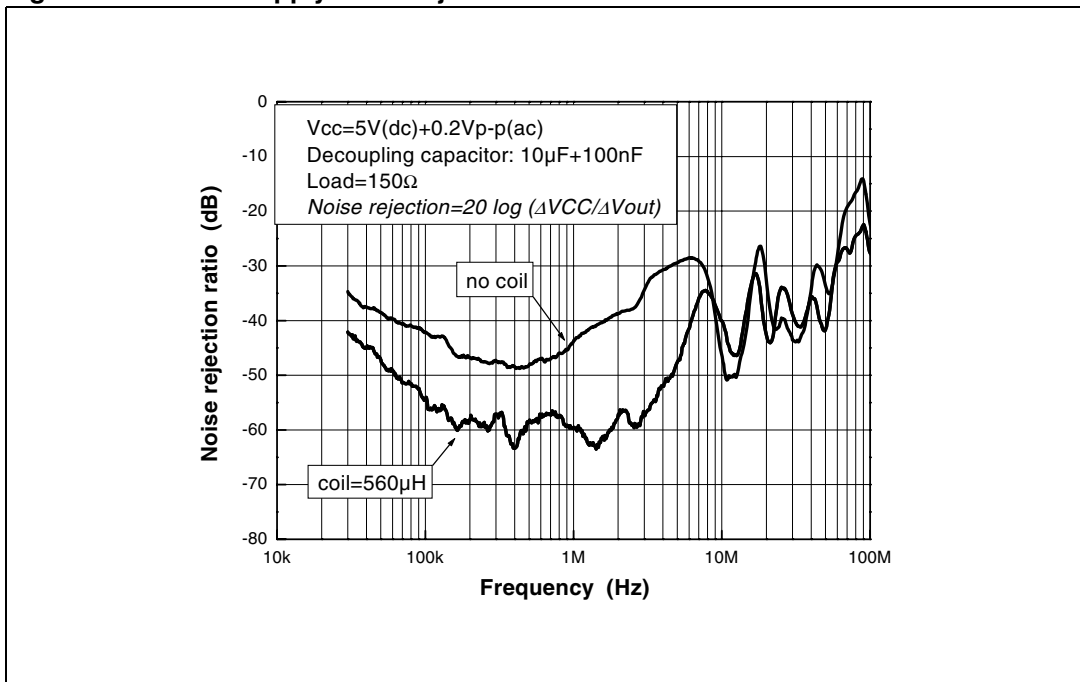


Figure 28 shows how the power supply noise rejection evolves according to the frequency and depending on how carefully power supply decoupling is achieved.

Figure 28. Power supply noise rejection



3 Using the TSH345 to drive YC, CVBS, YUV, YPbPr and RGB video components

Figure 29. Implementation of the video driver on output video DACs

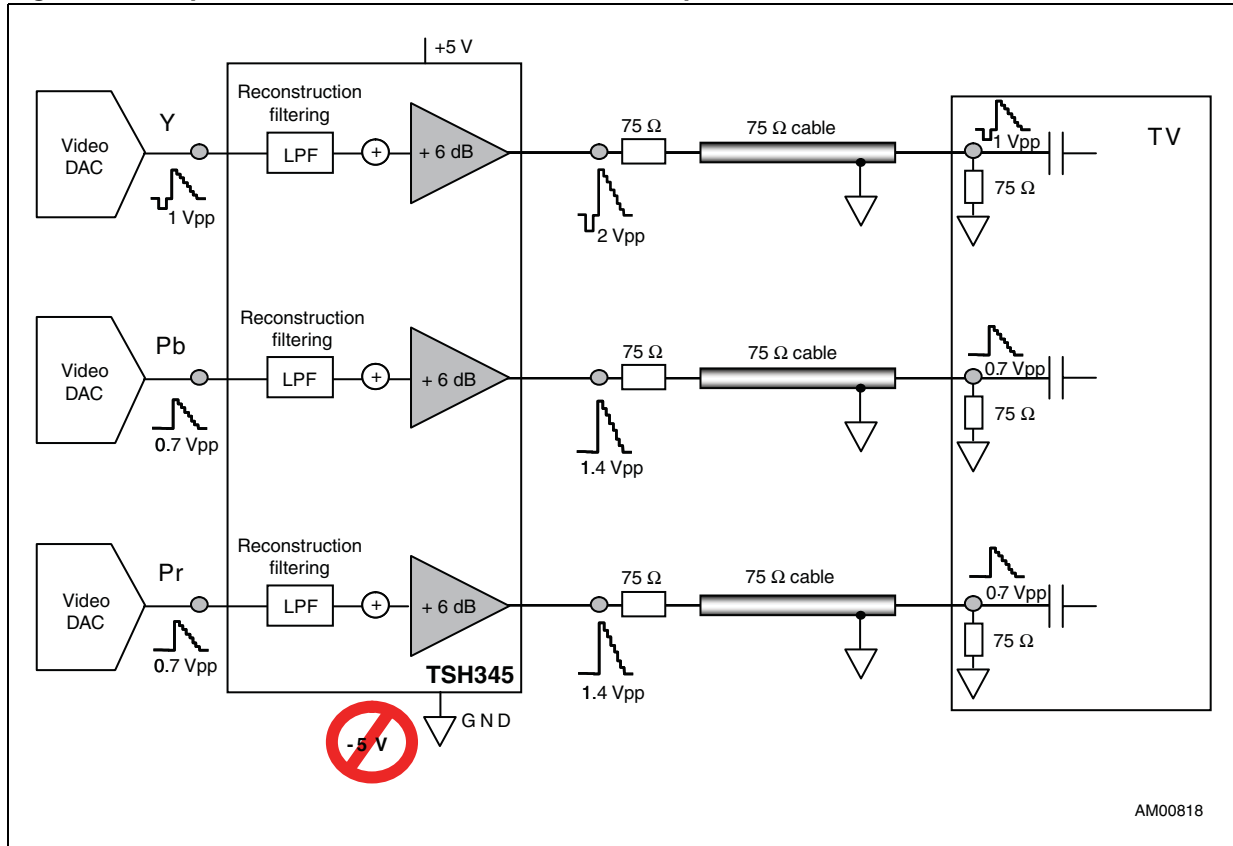


Figure 30. Synchronization details (example for a black picture)

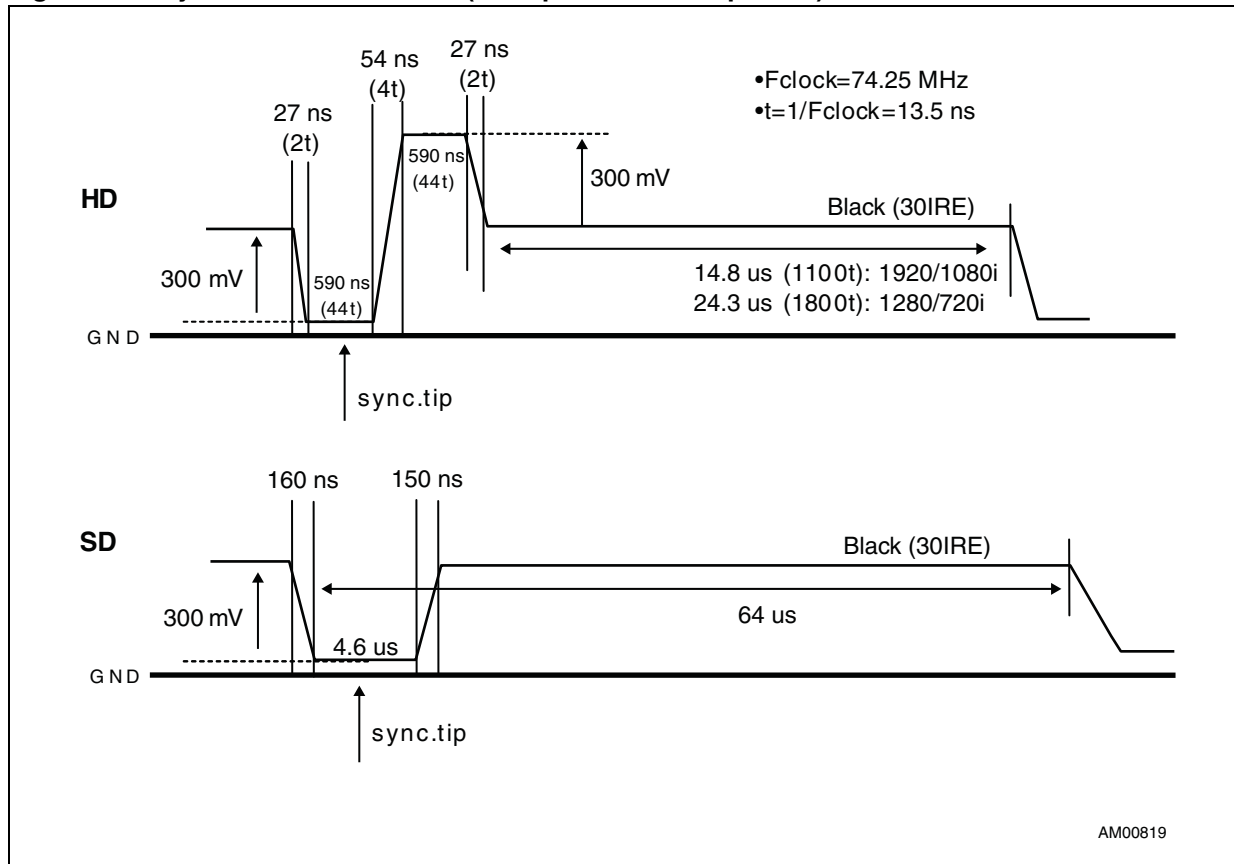


Figure 31. HD video signal

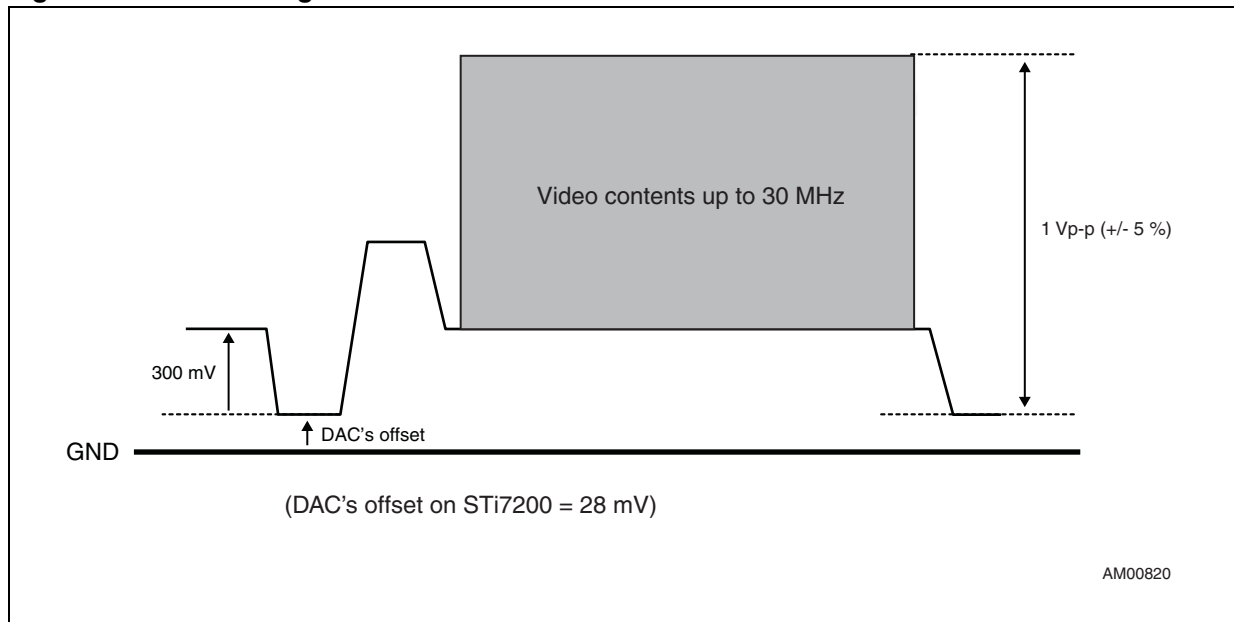


Figure 32. Standard video signal

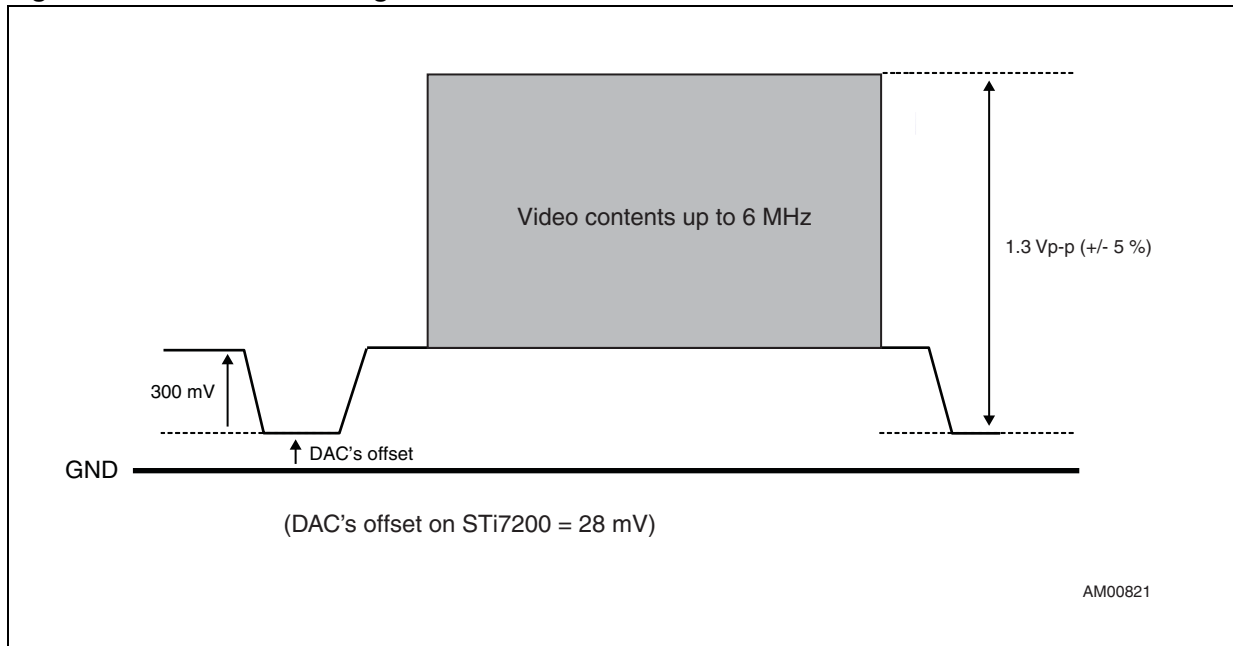
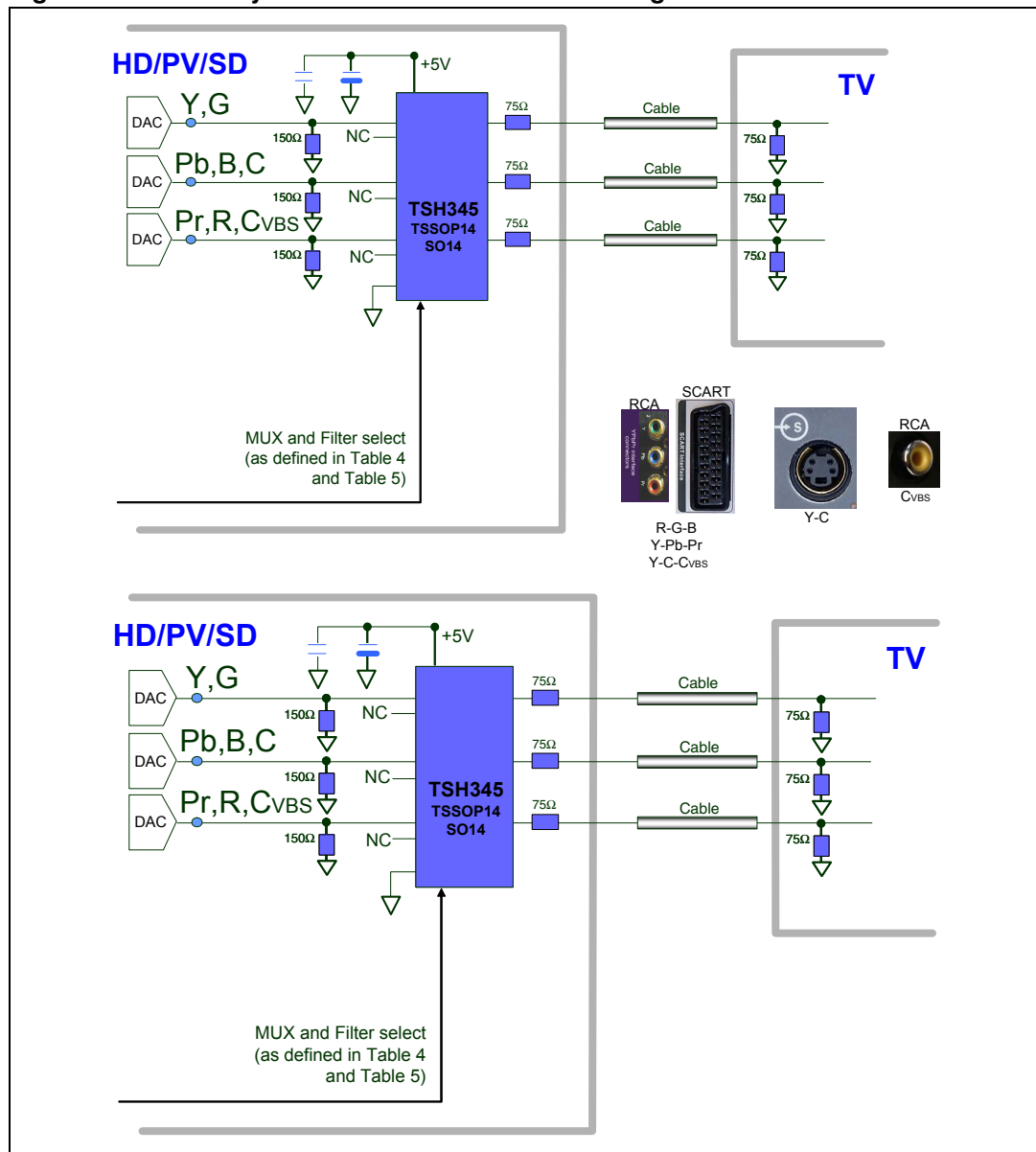


Figure 33. Flexibility of the TSH345 for SD and HD signals



The TSH345 is used to drive either high-definition video signals up to 30 MHz or progressive and interlaced standard definition video signals on 75-Ω video lines. It can drive a large panel of signals such as YC and CVBS, YUV, YPbPr and RGB, where the bottom of the signal (the synchronization tip in the case of Y and CVBS signals) is close to zero volts. An internal input DC value is added to the video signal in order to shift the bottom from GND.

The shift is not based on the average of the signal, but is an analog summation of a DC component to the video signal. Therefore, no input capacitors are required, which provides a real advantage in terms of cost and board space.

Under these conditions, it is possible to drive the signal in single supply without any saturation of the driver against the lower rail.

Since half of the signal is lost through output impedance matching, in order to properly drive the video line the shifted signal is multiplied by a gain of 2 or +6 dB.

3.1 Output capacitor

The output can be either DC-coupled or AC-coupled. The output can be directly connected to the line via a 75-Ω resistor (see [Figure 34](#)), or an output capacitor can be used to remove any DC components in the load. Assuming the load is 150 Ω, a coupling capacitor of 220 μF can be used to provide a very low cut-off frequency close to 5 Hz (see [Figure 35](#)).

Figure 34. DC output coupling for SD, PV and HD

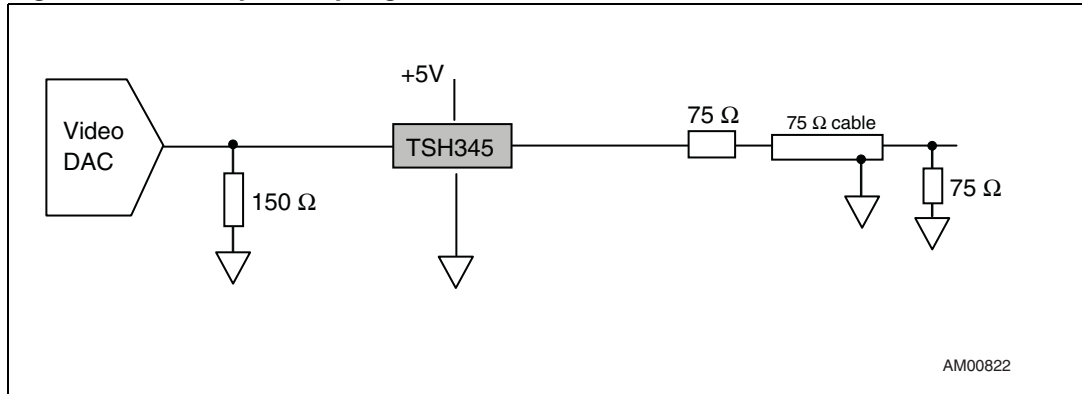
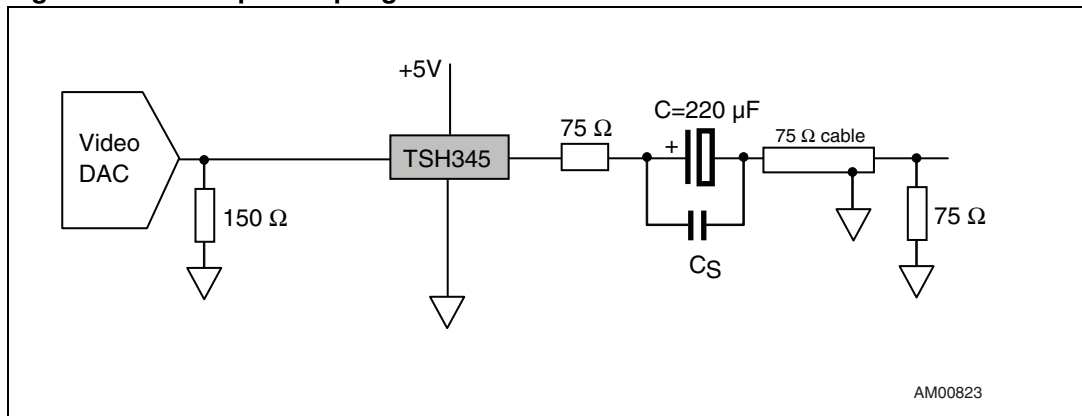


Figure 35. AC output coupling



1. C_S is 100 nF used to decrease the parasitic components of C in high frequencies. It is preferable to limit the use of this output AC-coupling to the standard definition only.
2. The 75-Ω resistor must be as close as possible to the output of the driver to minimize the effect of parasitic capacitance.

4 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK[®] packages, depending on their level of environmental compliance. ECOPACK[®] specifications, grade definitions and product status are available at: www.st.com. ECOPACK[®] is an ST trademark.

4.1 SO-14 package information

Figure 36. SO-14 package mechanical drawing

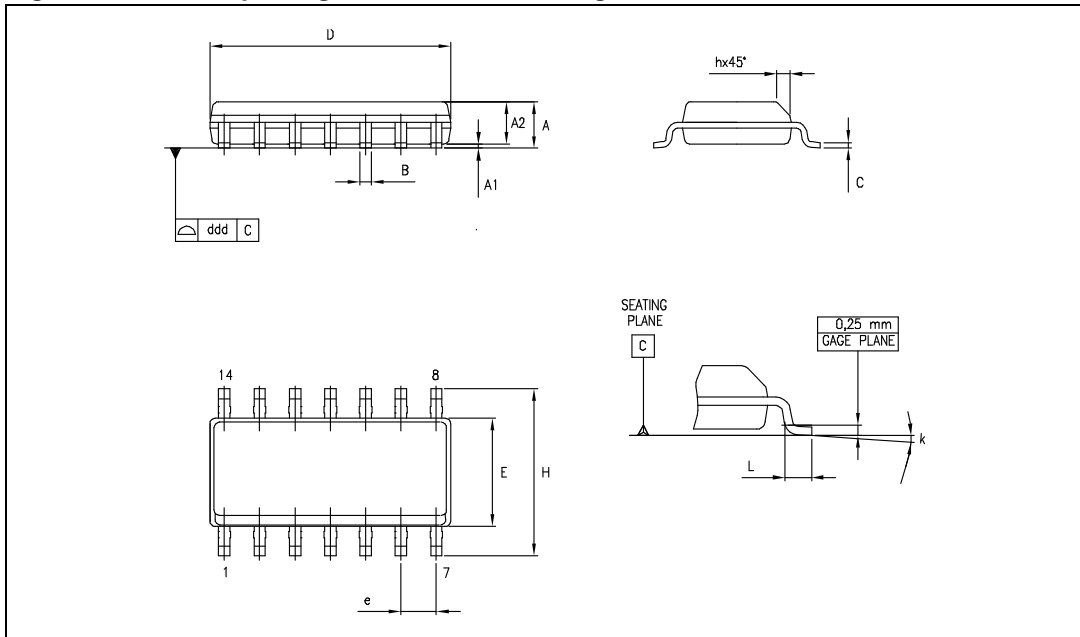


Table 6. SO-14 package mechanical data

Ref.	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	1.35		1.75	0.05		0.068
A1	0.10		0.25	0.004		0.009
A2	1.10		1.65	0.04		0.06
B	0.33		0.51	0.01		0.02
C	0.19		0.25	0.007		0.009
D	8.55		8.75	0.33		0.34
E	3.80		4.0	0.15		0.15
e		1.27			0.05	
H	5.80		6.20	0.22		0.24
h	0.25		0.50	0.009		0.02
L	0.40		1.27	0.015		0.05
k	8° (max.)					
ddd			0.10			0.004

Note: D and F dimensions do not include mold flash or protrusions. Mold flash or protrusions must not exceed 0.15 mm.

5 Ordering information

Table 8. Order codes

Part number	Temperature range	Package	Packing	Marking
TSH345ID	-40°C to +85°C	SO-14	Tube	TSH345I
TSH345IDT			Tape & reel	TSH345I
TSH345IPT		TSSOP14	Tape & reel	TSH345I

6 Revision history

Table 9. Document revision history

Date	Revision	Changes
29-May-2007	1	Initial release.
18-Dec-2008	2	Added curves in Chapter 2: Electrical characteristics . Added all test limits in Chapter Table 3 .