GaAs, MMIC, Fundamental Mixer, 3 GHz to 10 GHz

Data Sheet **HMC787A**

FEATURES

Conversion loss: 9 dB typical at 3 GHz to 9 GHz
Local oscillator (LO) to radio frequency (IF) isolation: 43 dB typical at 3 GHz to 9 GHz
RF to intermediate frequency (IF) isolation: 26 dB typical at 3 GHz to 9 GHz
Input third-order intercept (IP3): 24 dBm typical at 3 GHz to 9 GHz
Input 1 dB compression point (P1dB): 17 dBm typical at 3 GHz to 9 GHz
Input second-order intercept (IP2): 67 dBm typical at 3 GHz to 9 GHz
Passive double-balanced topology
Wide IF frequency range: dc to 4 GHz
12-terminal, ceramic, leadless chip carrier (LCC) package

APPLICATIONS

Microwave radio
Industrial, scientific, and medical (ISM) band and ultrawide band (UWB) radio
Test equipment and sensors
Military end use

FUNCTIONAL BLOCK DIAGRAM

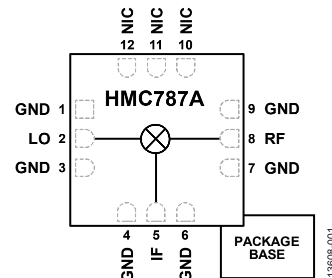

Figure 1.

GENERAL DESCRIPTION

The HMC787A is a general-purpose, double balanced mixer in a 12-terminal, RoHS compliant, ceramic leadless chip carrier (LCC) package that can be used as an upconverter or down-converter from 3 GHz to 10 GHz. This mixer is fabricated in a gallium arsenide (GaAs), metal semiconductor field effect transistor (MESFET) process and requires no external components or matching circuitry.

The HMC787A provides excellent local oscillator (LO) to radio frequency (RF) and LO to intermediate frequency (IF) isolation due to optimized balun structures and operates with a LO drive level of 17 dBm. The ceramic LCC package eliminates the need for wire bonding and is compatible with high volume, surface-mount manufacturing techniques.

Rev. C **Document Feedback**

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REVISION HISTORY

9/2017—Rev. B to Rev. C
Changes to Table 4 ... 5

5/2017—Rev. A to Rev. B
Changed E-12-1 to E-12-4 ... Throughout
Change to Figure 1 ... 1
Change to Figure 2 ... 5
Updated Outline Dimensions ... 17
Changes to Ordering Guide ... 17

2/2017—Rev. 0 to Rev. A
Changes to Storage Temperature Range Parameter, Table 2 ... 4

10/2016—Revision 0: Initial Version

SPECIFICATIONS

T_A = 25°C, IF = 100 MHz, LO = 17 dBm, and all measurements performed as downconverter, unless otherwise noted.

Table 1.

Parameter	Min	Typ	Max	Unit
FREQUENCY RANGE				
RF	3		10	GHz
LO	3		10	GHz
IF	DC		4	GHz
LO DRIVE LEVEL		17		dBm
PERFORMANCE AT RF = 3 GHz to 9 GHz				
Conversion Loss		9	11	dB
Single Sideband (SSB) Noise Figure		9		dB
Input Third-Order Intercept (IP3)	15	24		dBm
Input 1 dB Compression Point (P1dB)		17		dBm
Input Second-Order Intercept (IP2)		67		dB
RF to IF Isolation	15	26		dB
LO to RF Isolation		48		dB
LO to IF Isolation	35	43		dB
PERFORMANCE AT RF = 9 GHz to 10 GHz				
Conversion Loss		9	11	dB
SSB Noise Figure		9		dB
Input IP3	15	24		dBm
Input P1dB		15		dBm
Input IP2		66		dB
RF to IF Isolation	15	26		dB
LO to RF Isolation		47		dB
LO to IF Isolation	25	42		dB

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
RF Input Power	28 dBm
LO Input Power	28 dBm
IF Input Power	28 dBm
IF Source and Sink Current	12 mA
Continuous Power Dissipation, P_{DISS} (T_A = 85°C, Derate 11.6 mW/°C Above 85°C)	1044 mW
Maximum Junction Temperature	175°C
Maximum Peak Reflow Temperature (MSL3)[1]	260°C
Operating Temperature Range	−40°C to +85°C
Storage Temperature Range	−65°C to +150°C
Electrostatic Discharge (ESD) Sensitivity	
Human Body Model (HBM)	1500 V (Class 1C)
Field Induced Charged Device Model (FICDM)	1000 V (Class C5)

[1] See the Ordering Guide section.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

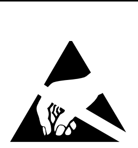
Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Careful attention to PCB thermal design is required.

Table 3. Thermal Resistance

Package Type	θ_{JA}	θ_{JC}	Unit
E-12-4[1]	120	86	°C/W

[1] See JEDEC standard JESD51-2 for additional information on optimizing the thermal impedance (PCB with 3 × 3 vias).

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

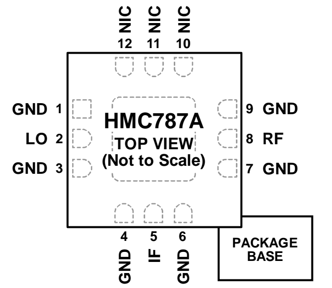


Figure 2. Pin Configuration

Table 4. Pin Function Descriptions

Pin No.	Mnemonic	Description
1, 3, 4, 6, 7, 9	GND	Ground. Connect the package bottom to RF/dc ground. See Figure 3 for the GND interface schematic.
2	LO	Local Oscillator. This pin is ac-coupled and matched to 50 Ω. See Figure 4 for the LO interface schematic.
5	IF	Intermediate Frequency. This pin is dc-coupled. For applications not requiring operation to dc, externally block this pin using a series capacitor whose value is chosen to pass the necessary IF frequency range. For operation to dc, this pin must not source or sink more than 12 mA of current or device nonfunction and possible device failure results. See for Figure 5 the IF interface schematic.
8	RF	Radio Frequency. This pin is ac-coupled and matched to 50 Ω. See Figure 6 for the RF interface schematic.
10 to 12	NIC	Not Internally Connected.
	EPAD	Exposed Pad. Exposed pad must be connected to RF/dc ground.

INTERFACE SCHEMATICS

Figure 3. GND Interface Schematic

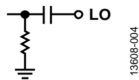


Figure 4. LO Interface Schematic

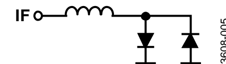


Figure 5. IF Interface Schematic

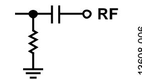


Figure 6. RF Interface Schematic

TYPICAL PERFORMANCE CHARACTERISTICS

DOWNCONVERTER PERFORMANCE

Data taken as downconverter, lower sideband (high-side LO), T_A = 25°C, and LO drive level = 17 dBm, unless otherwise noted. Measurements taken with LO amplifier in line with lab bench LO source.

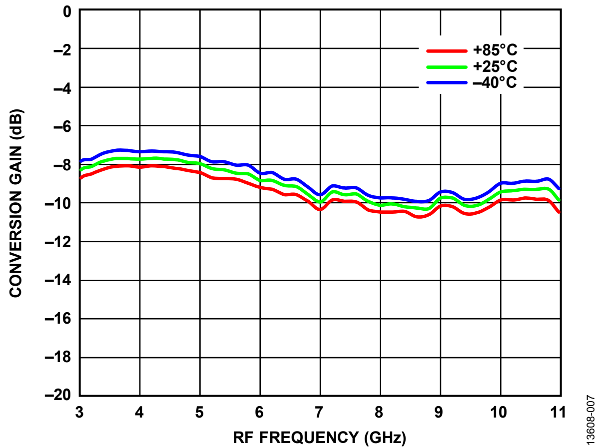


Figure 7. Conversion Gain vs. RF Frequency for Various Temperatures, IF = 100 MHz, LO Power = 17 dBm

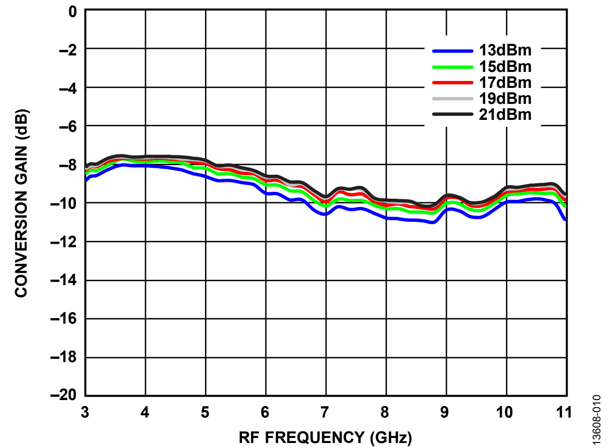


Figure 10. Conversion Gain vs. RF Frequency for Various LO Powers, IF = 100 MHz, Temperature = 25°C

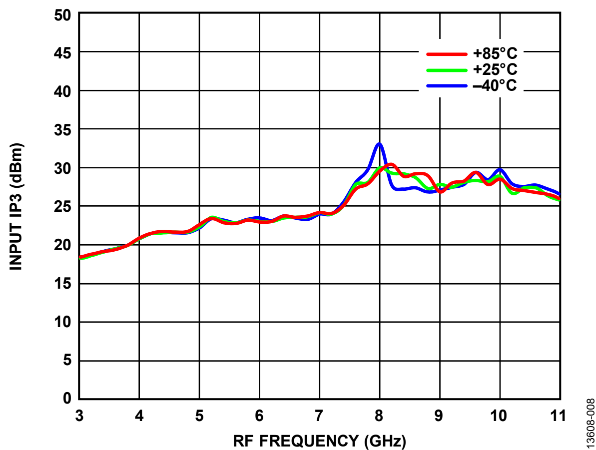


Figure 8. Input Third-Order Intercept (IP3) vs. RF Frequency for Various Temperatures, IF = 100 MHz, LO Power = 17 dBm

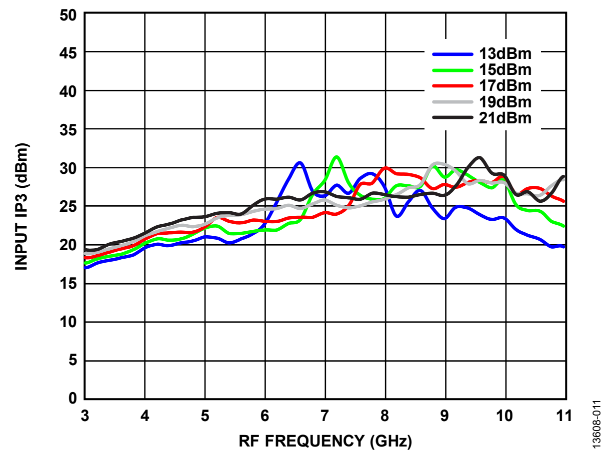


Figure 11. Input Third-Order Intercept (IP3) vs. RF Frequency for Various LO Powers, IF = 100 MHz, Temperature = 25°C)

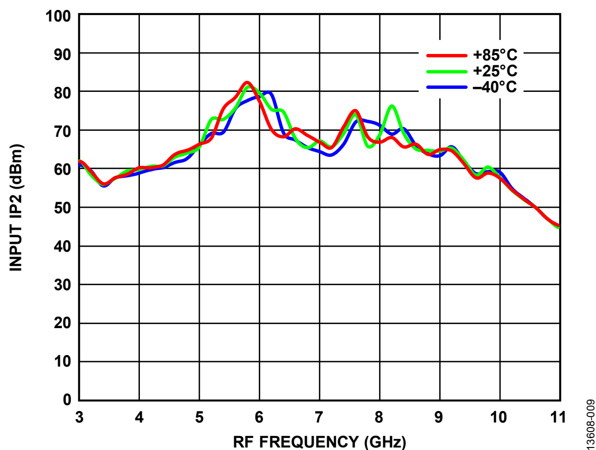


Figure 9. Input Second-Order Intercept (IP2) vs. RF Frequency for Various Temperatures, IF = 100 MHz, LO Power = 17 dBm

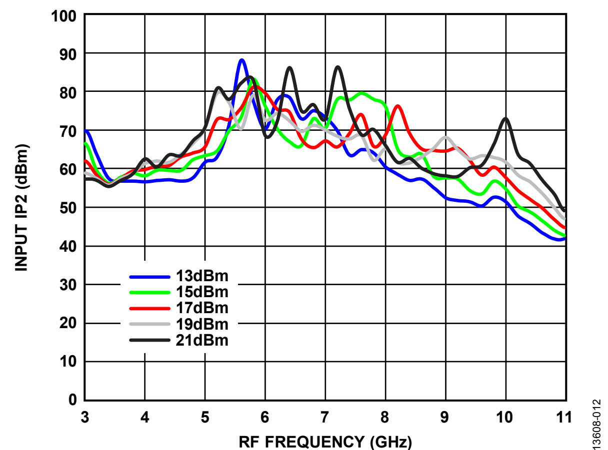


Figure 12. Input Second-Order Intercept (IP2) vs. RF Frequency for Various LO Powers, IF = 100 MHz, Temperature = 25°C

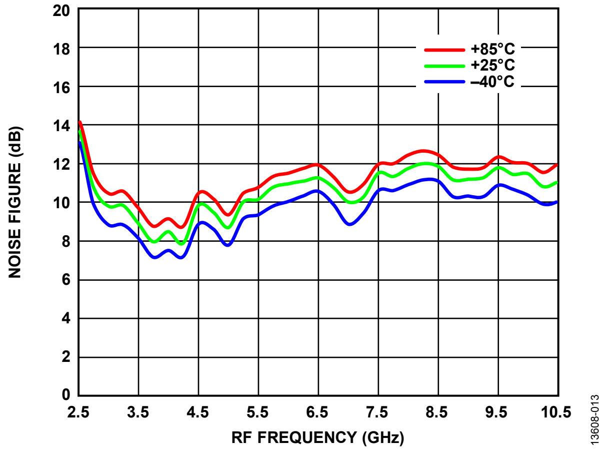


Figure 13. Single Sidband Noise Figure vs. RF Frequency for Various Temperatures, IF = 100 MHz, LO Power = 17 dBm

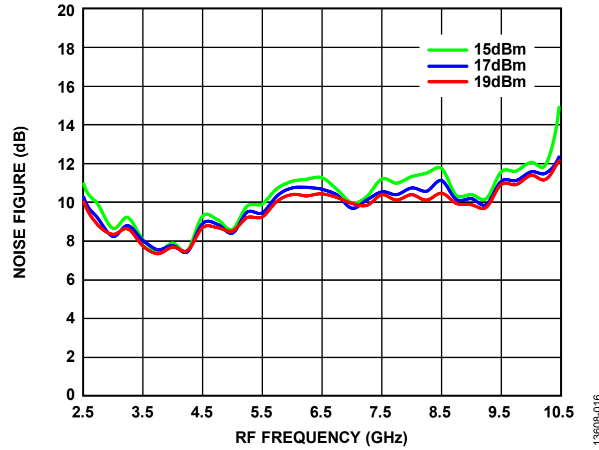


Figure 16. Single Sideband Noise Figure vs. RF Frequency for Various LO Powers, IF = 100 MHz, Temperature = 25°C

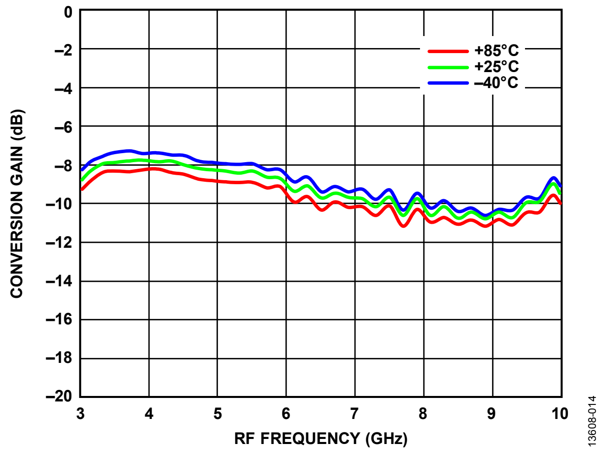


Figure 14. Conversion Gain vs. RF Frequency for Various Temperatures, IF = 1100 MHz, LO Power = 17 dBm

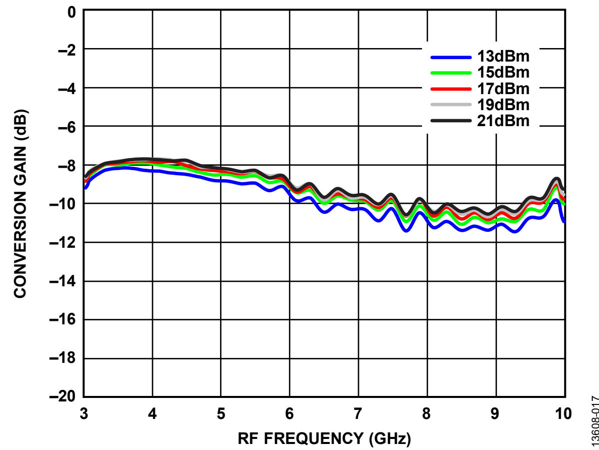


Figure 17. Conversion Gain vs. RF Frequency for Various LO Powers, IF = 1100 MHz, Temperature = 25°C

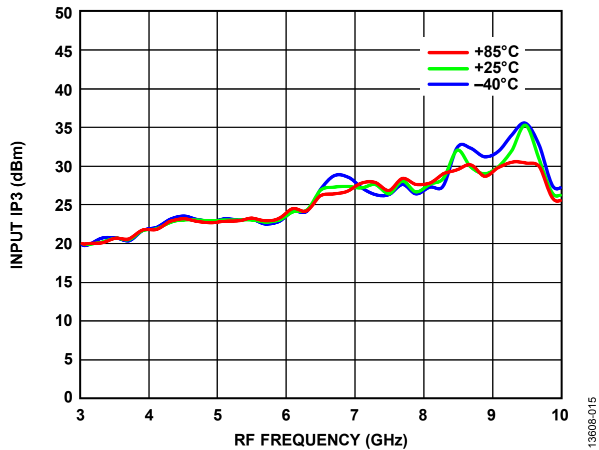


Figure 15. Input Third-Order Intercept (IP3) vs. RF Frequency for Various Temperatures, IF = 1100 MHz, LO Power = 17 dBm

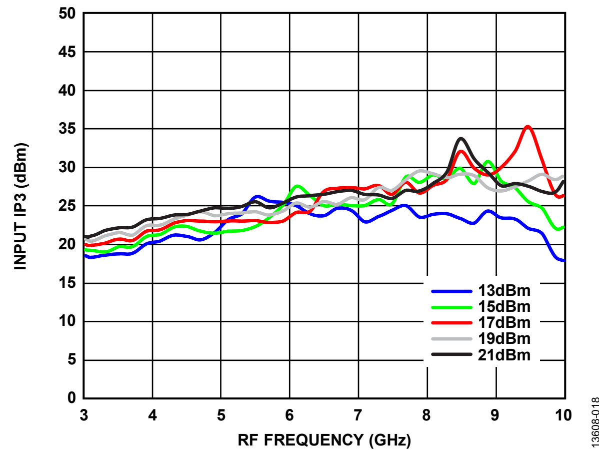


Figure 18. Input Third-Order Intercept (IP3) vs. RF Frequency for Various LO Powers, IF = 1100 MHz, Temperature = 25°C

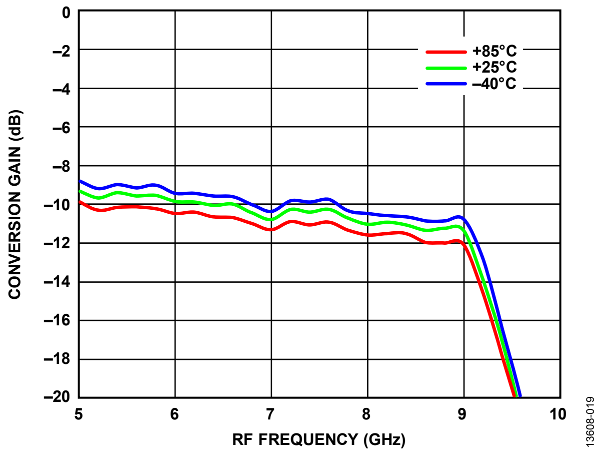


Figure 19. Conversion Gain vs. RF Frequency for Various Temperatures, IF = 3000 MHz, LO Power = 17 dBm

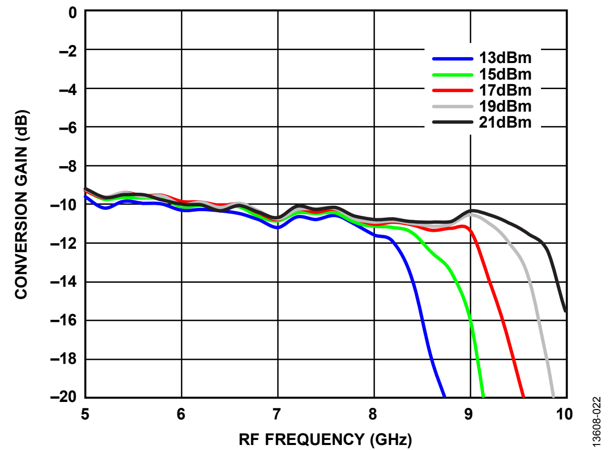


Figure 22. Conversion Gain vs. RF Frequency for Various LO Powers, IF = 3000 MHz, Temperature = 25°C

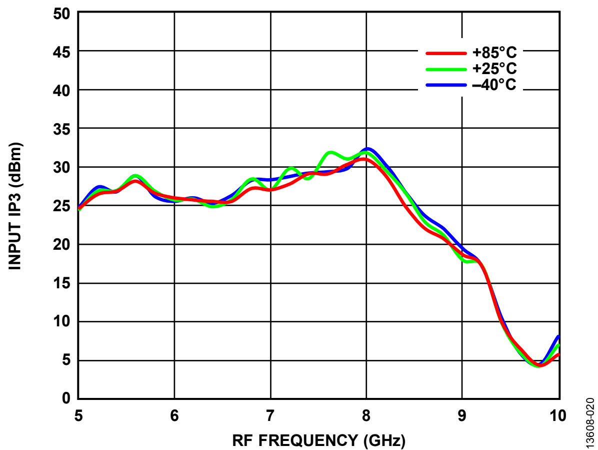


Figure 20. Input Third-Order Intercept (IP3) vs. RF Frequency for Various Temperatures, IF = 3000 MHz, LO Power = 17 dBm

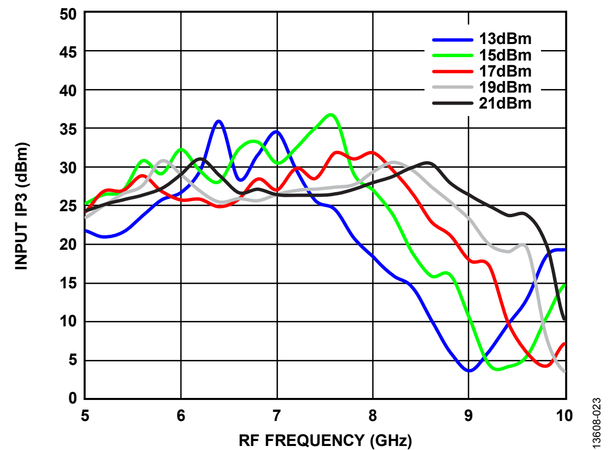


Figure 23. Input Third-Order Intercept (IP3) vs. RF Frequency for Various LO Powers, IF = 3000 MHz, Temperature = 25°C

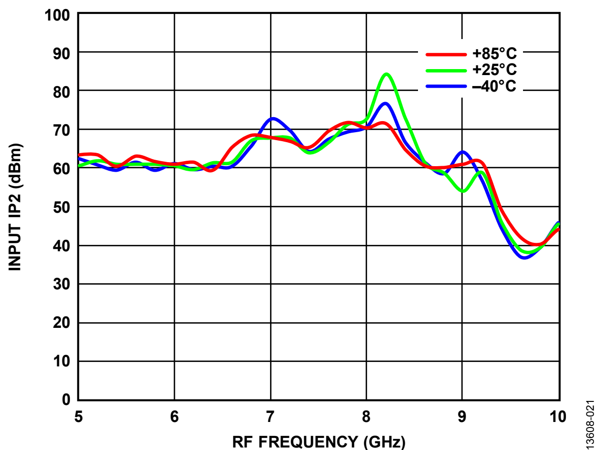


Figure 21. Input Second-Order Intercept (IP2) vs. RF Frequency at Various Temperatures, IF = 3000 MHz, LO Power = 17 dBm

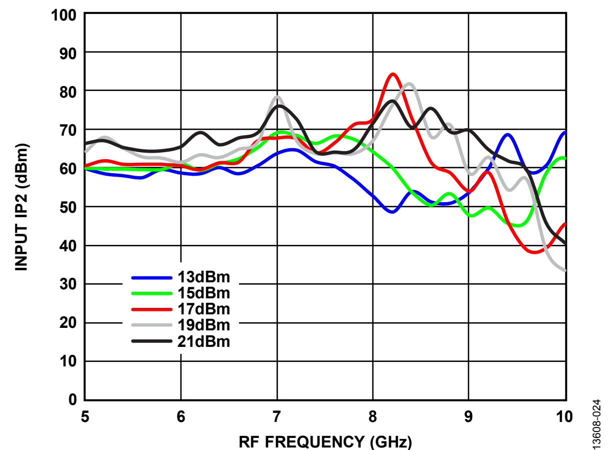


Figure 24. Input Second-Order Intercept (IP2) vs. RF Frequency for Various LO Powers, IF = 3000 MHz, Temperature = 25°C

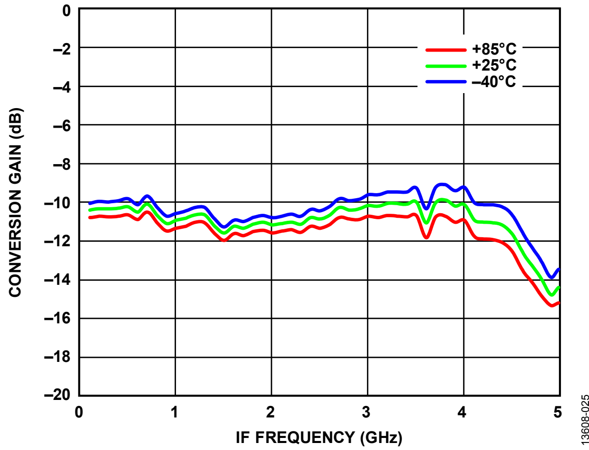


Figure 25. Conversion Gain vs. IF Frequency for Various Temperatures, LO = 9510 MHz, LO Power = 17 dBm

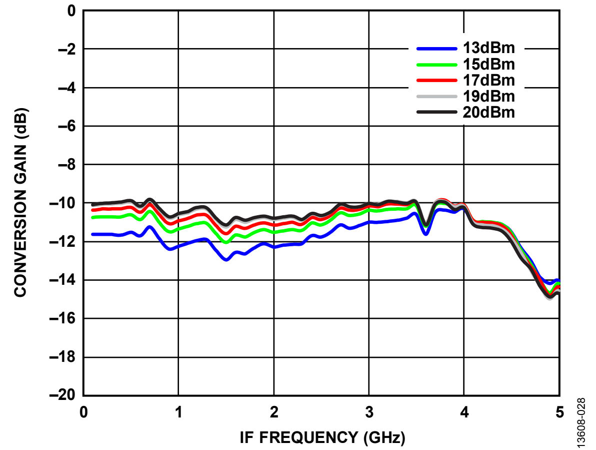


Figure 28. Conversion Gain vs. IF Frequency for Various LO Powers, LO = 9510 MHz, LO Power = 17 dBm

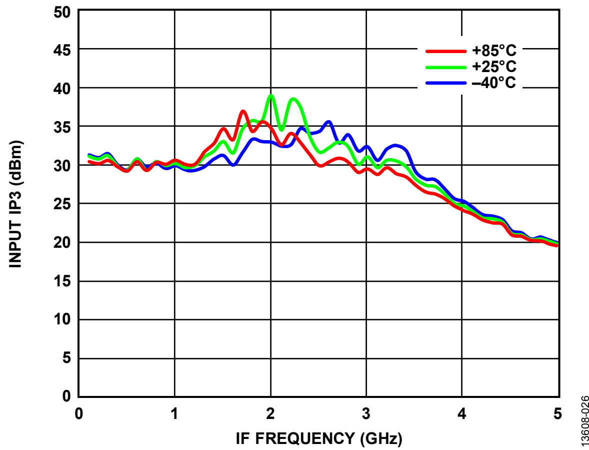


Figure 26. Input Third-Order Intercept (IP3) vs. IF Frequency for Various Temperatures, LO = 9510 MHz, LO Power = 17 dBm

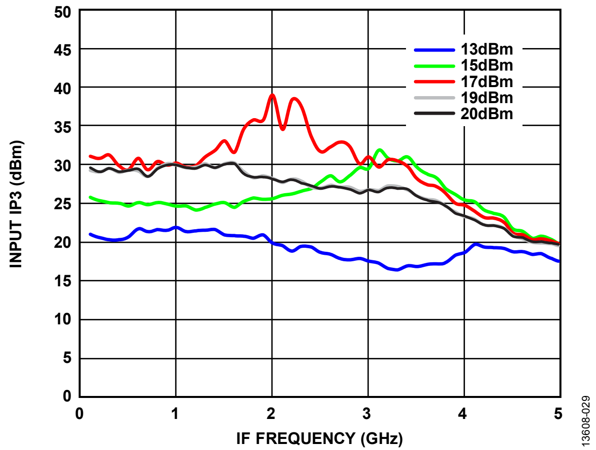


Figure 29. Input Third-Order Intercept (IP3) vs. IF Frequency for Various LO Powers, LO = 9510 MHz, LO Power = 17 dBm

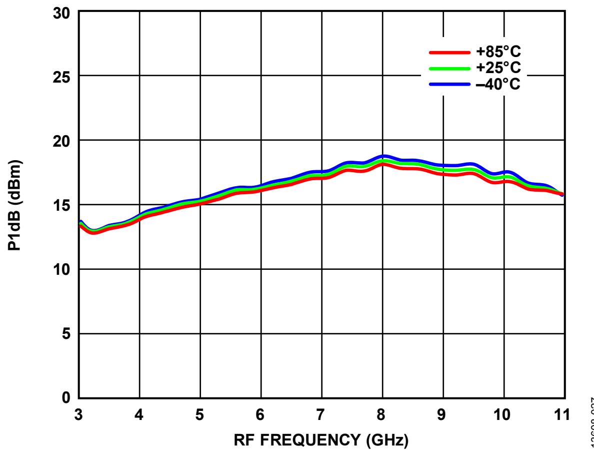


Figure 27. Input 1 dB Gain Compression (P1dB) vs. RF Frequency for Various Temperatures, IF = 100 MHz, LO = 17 dBm

UPCONVERTER PERFORMANCE

Data taken as upconverter, lower sideband (high-side LO), T_A = 25°C, and LO drive level = 17 dBm, unless otherwise noted. Measurements taken with LO amplifier in line with lab bench LO source.

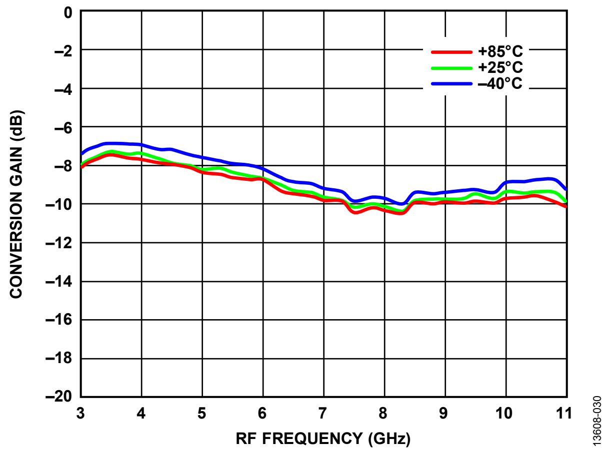


Figure 30. Conversion Gain vs. RF Frequency for Various Temperatures, IF = 100 MHz, LO Power = 17 dBm

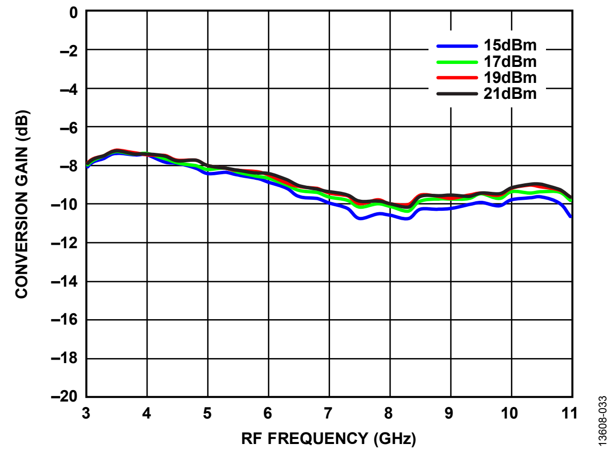


Figure 33. Conversion Gain vs. RF Frequency for Various LO Powers, IF = 100 MHz, Temperature = 25°C

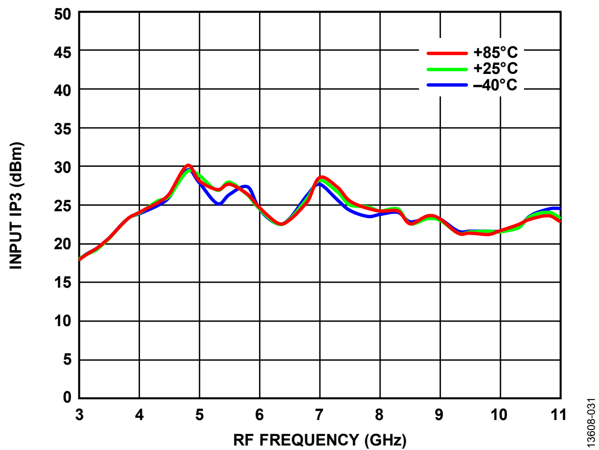


Figure 31. Input Third-Order Intercept (IP3) vs. RF Frequency for Various Temperatures, IF = 100 MHz, LO Power = 17 dBm

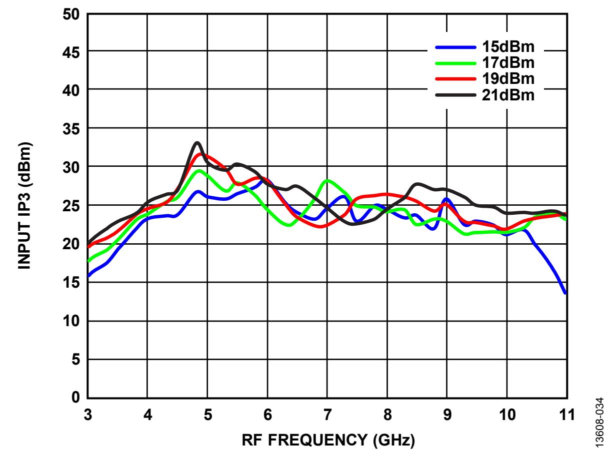


Figure 34. Input Third-Order Intercept (IP3) vs. RF Frequency for Various LO Powers, IF = 100 MHz, Temperature = 25°C

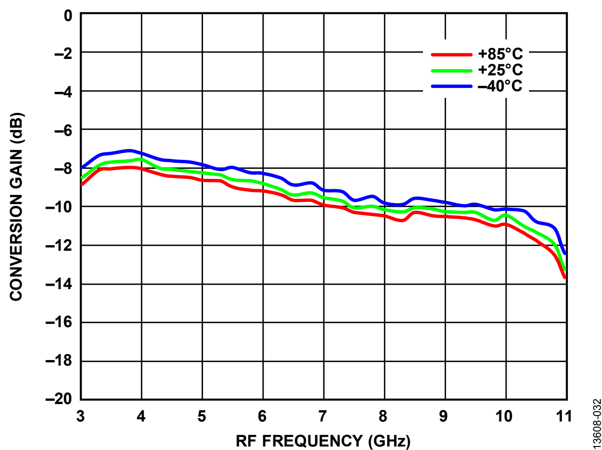


Figure 32. Conversion Gain vs. RF Frequency for Various Temperatures, IF = 1100 MHz, LO Power = 17 dBm

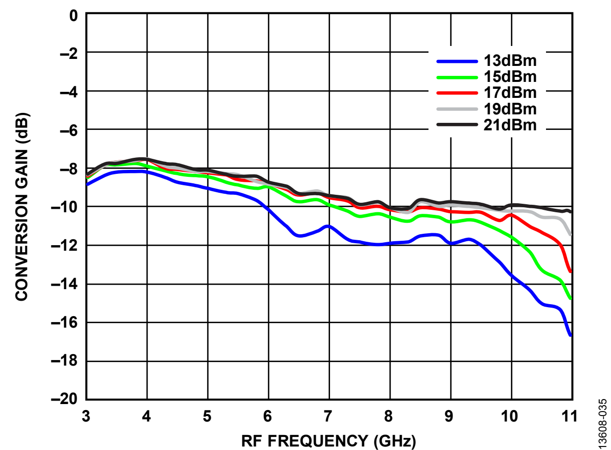


Figure 35. Conversion Gain vs. RF Frequency for Various LO Powers, IF = 1100 MHz, Temperature = 25°C

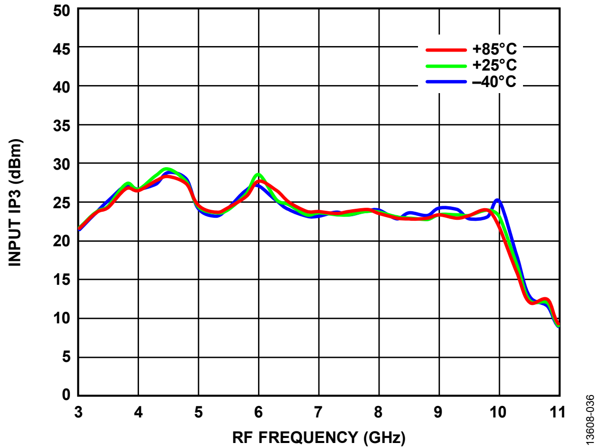


Figure 36. Input Third-Order Intercept (IP3) vs. RF Frequency for Various Temperatures at IF = 1100 MHz, LO Power = 17 dBm

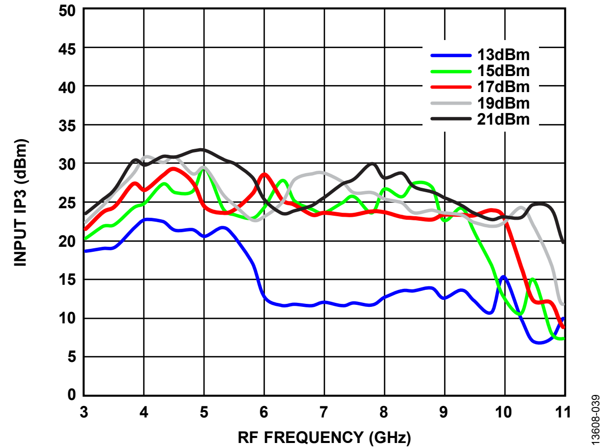


Figure 39. Input Third-Order Intercept (IP3) vs. RF Frequency for Various LO Powers at IF = 1100 MHz, Temperature = 25°C

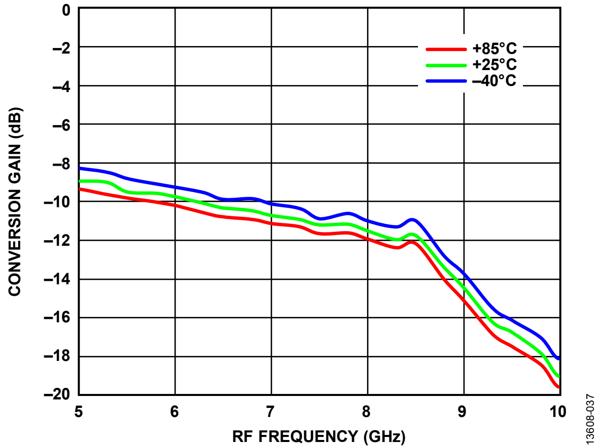


Figure 37. Conversion Gain vs. RF Frequency for Various Temperatures, IF = 3000 MHz, LO Power = 17 dBm

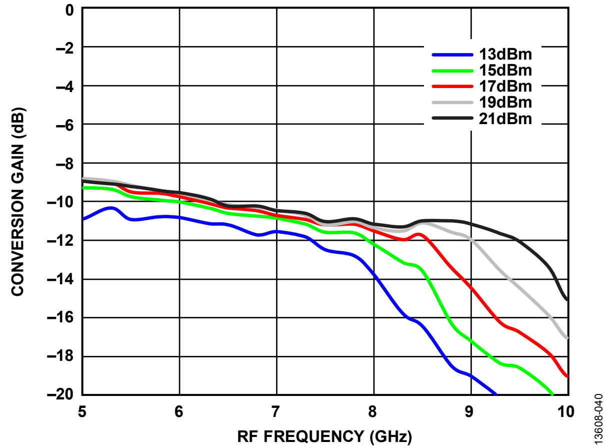


Figure 40. Conversion Gain vs. RF Frequency for Various LO Powers, IF = 3000 MHz, Temperature = 25°C

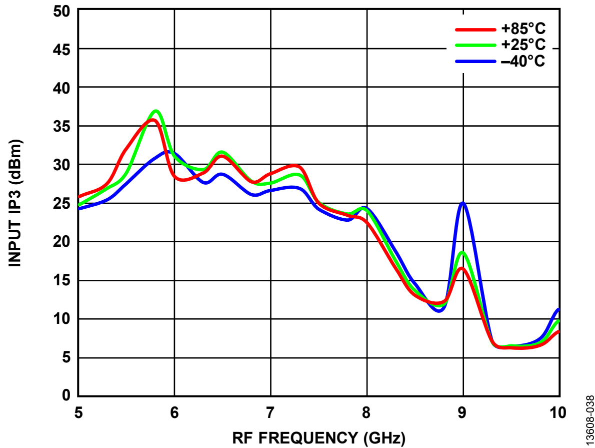


Figure 38. Input Third-Order Intercept (IP3) vs. RF Frequency for Various Temperatures at IF = 3000 MHz, LO Power = 17 dBm

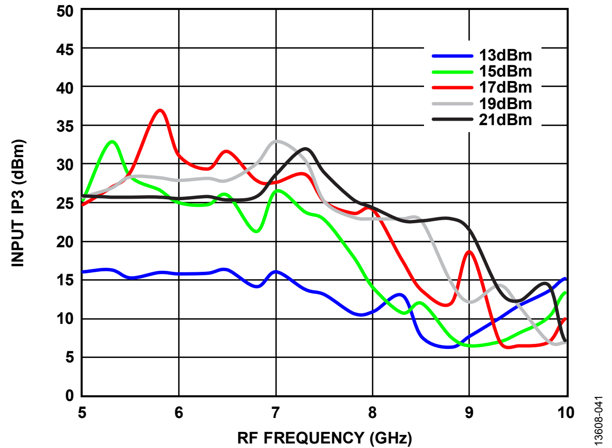


Figure 41. Input Third-Order Intercept (IP3) vs. RF Frequency for Various LO Powers at IF = 3000 MHz, Temperature = 25°C

ISOLATION AND RETURN LOSS PERFORMANCE

Data taken as downconverter, lower sideband (high-side LO), IF = 100 MHz, T_A = 25°C, and LO drive level = 17 dBm, unless otherwise noted. Measurements taken with LO amplifier in line with lab bench LO source.

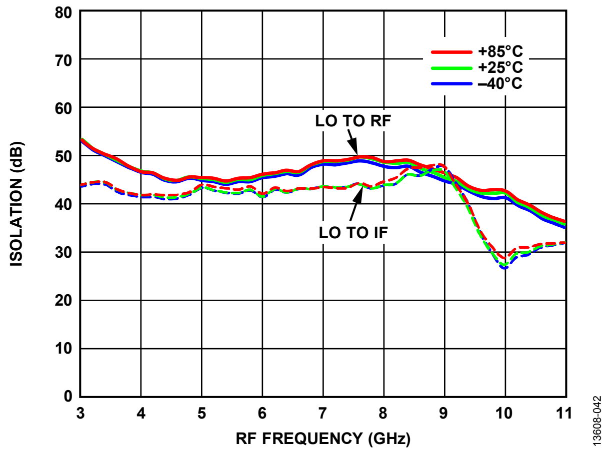


Figure 42. LO to RF and LO to IF Isolation vs. RF Frequency for Various Temperatures, LO Power = 17 dBm

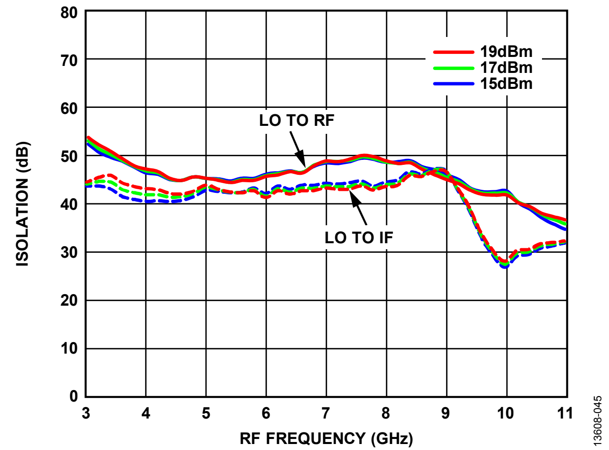


Figure 45. LO to RF and LO to IF Isolation vs. RF Frequency for Various LO Powers, Temperature = 25°C

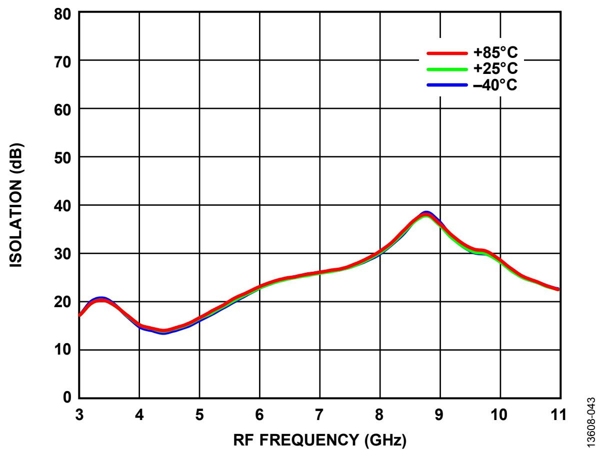


Figure 43. RF to IF Isolation vs. RF Frequency for Various Temperatures, LO Frequency = 7000 MHz, LO Power = 17 dBm

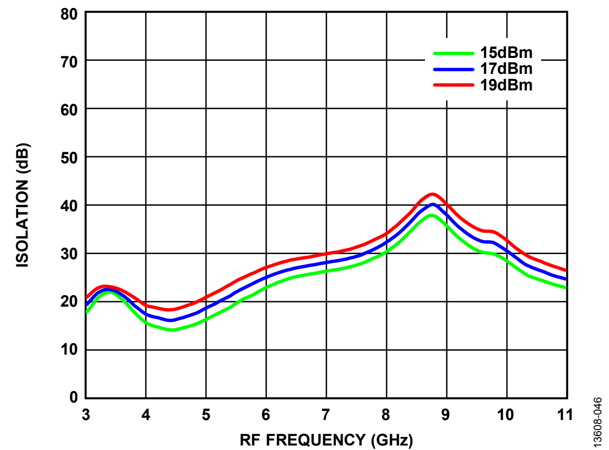


Figure 46. RF to IF Isolation vs. RF Frequency for Various LO Powers, Temperature = 25°C, LO Frequency = 7000 MHz

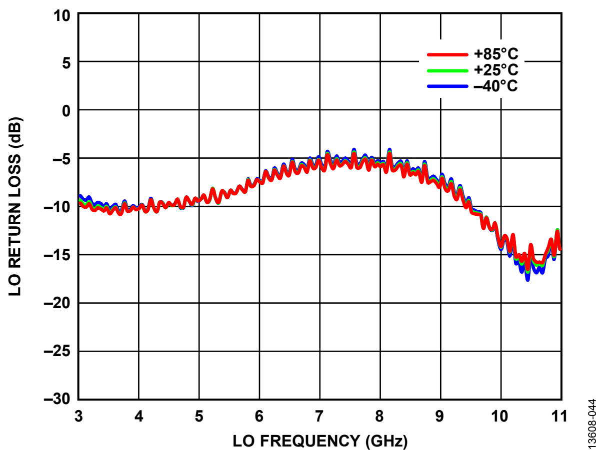


Figure 44. LO Return Loss vs. LO Frequency at Various Temperatures, LO Power = 17 dBm

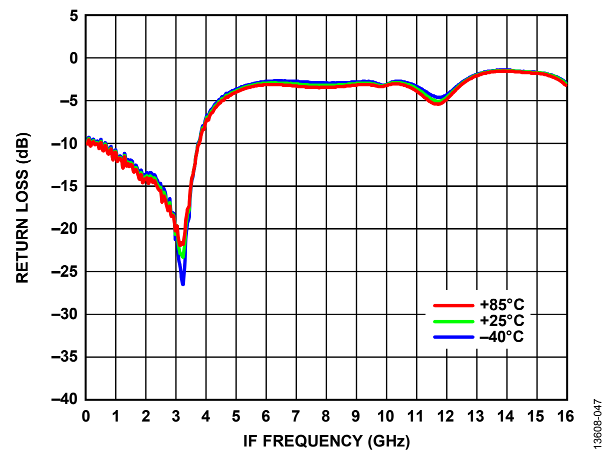


Figure 47. IF Return Loss vs. IF Frequency at Various Temperatures, LO Power = 17 dBm, LO Frequency = 6 GHz

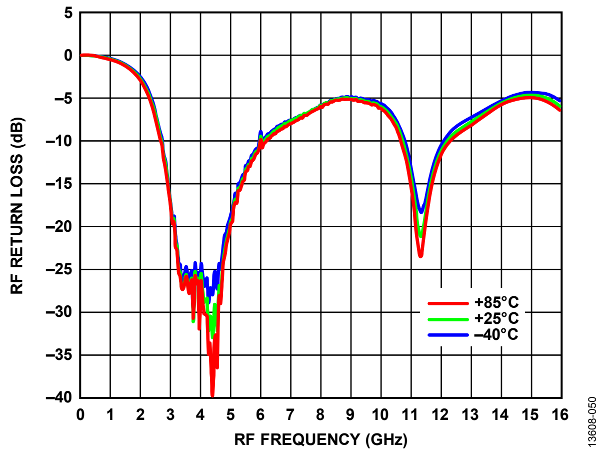


Figure 48. RF Return Loss vs. RF Frequency at Various Temperatures, LO Power = 17 dBm, LO Frequency = 6 GHz

SPURIOUS AND HARMONICS PERFORMANCE

Mixer spurious products are measured in dBc from the IF output power level, unless otherwise noted. Spur values are (M × RF) − (N × LO).

LO Harmonics

LO = 17 dBm, and all values in dBc below input LO level measured at RF port.

Table 5. Harmonics of LO

LO Frequency (GHz)	N_{LO} Spur at RF Port			
	1	2	3	4
3	55	53	58	59
6	46	45	64	76
10	41	57	61	61

M × N Spurious Outputs, IF = 100 MHz

RF = 3.1 GHz, LO = 3 GHz, RF power = −5 dBm, and LO power = 17 dBm.

		N × LO				
		0	1	2	3	4
M × RF	0	Not applicable	14	38	32	54
	1	12.3	0	25	27	60
	2	90.1	93	98	94	90
	3	91.1	90	92	98	94
	4	89.1	88	91	94	98

RF = 6.1 GHz, LO = 6 GHz, RF power = −5 dBm, and LO power = 17 dBm.

		N × LO				
		0	1	2	3	4
M × RF	0	Not applicable	12	26	42	42
	1	16.3	0	36	35	50
	2	89.1	91	97	90	88
	3	86.2	88	88	95	87
	4	83.1	87	89	90	96

RF =10.1 GHz, LO = 10 GHz, RF power = −5 dBm, and LO power = 17 dBm.

		N × LO				
		0	1	2	3	4
M × RF	0	Not applicable	4	39	36	43
	1	19.9	0	87	83	79
	2	84.3	84	97	87	83
	3	78.4	82	87	97	87
	4	72.4	78	83	89	95

M × N Spurious Outputs, IF = 3000 MHz

RF = 3.1 GHz, LO = 6.1 GHz, RF power = −5 dBm, and LO power = 17 dBm.

		N × LO				
		0	1	2	3	4
M × RF	0	Not applicable	13	28	44	44
	1	13.6	0	47	45	64
	2	90.7	99	89	89	86
	3	91	93	94	87	87
	4	90.2	89	99	92	90

RF = 6.1 GHz, LO = 3.1 GHz, RF power = −5 dBm, and LO power = 17 dBm.

		N × LO				
		0	1	2	3	4
M × RF	0	Not applicable	13	39	28	50
	1	16	0	24	18	90
	2	89	88	89	92	97
	3	86	86	89	87	89
	4	82	86	85	87	89

RF = 10.1 GHz, LO = 7.1 GHz, RF power = −5 dBm, and LO power = 17 dBm.

		N × LO				
		0	1	2	3	4
M × RF	0	Not applicable	11	28	41	54
	1	17.7	0	30	35	52
	2	82.9	86	84	94	86
	3	75	81	86	87	92
	4	68	76	79	83	87

THEORY OF OPERATION

The HMC787A is a general-purpose, double balanced mixer in a 12-terminal, RoHS compliant, ceramic leadless chip carrier (LCC) package that can be used as an upconverter or downconverter from 3 GHz to 10 GHz. This mixer is fabricated in a gallium arsenide (GaAs), metal semiconductor field effect transistor (MESFET) process and requires no external components or matching circuitry. The HMC787A provides excellent local oscillator (LO) to radio frequency (RF) and LO to intermediate frequency (IF) isolation due to optimized balun structures and operates with a LO drive level of 17 dBm. The ceramic LCC package eliminates the need for wire bonding and is compatible with high volume, surface-mount manufacturing techniques.

APPLICATIONS INFORMATION

TYPICAL APPLICATION CIRCUIT

Figure 49 shows the typical application circuit for the HMC787A. The LO and RF pins are internally ac-coupled. When IF operation is not required until dc, it is recommended to use an ac-coupled capacitor at the IF port. When IF operation to dc is required, do not exceed the IF source and sink currents specified in the Absolute Maximum Ratings section.

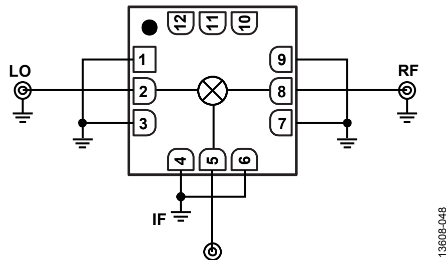


Figure 49. Typical Applications Circuit

EVALUATION PCB INFORMATION

The circuit board used in the application must use RF circuit design techniques. Signal lines must have 50 Ω impedance, and the package ground leads and exposed pad must be connected directly to the ground plane similarly to that shown in Figure 50. Use a sufficient number of via holes to connect the top and bottom ground planes. The evaluation circuit board shown in Figure 50 is available from Analog Devices, Inc., upon request.

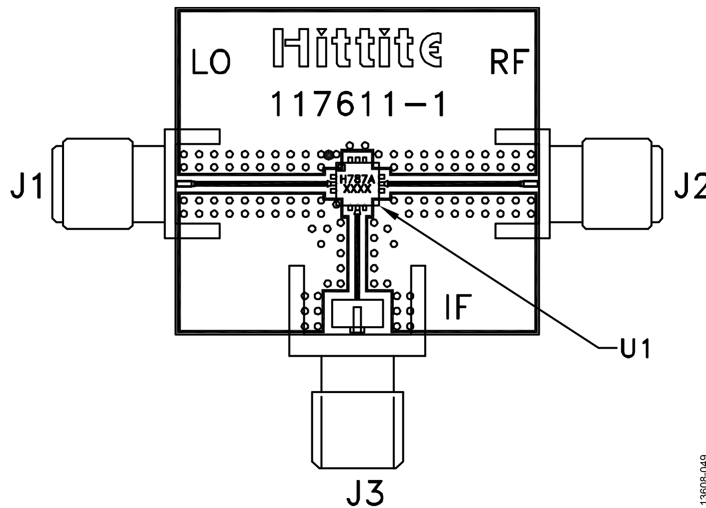


Figure 50. Evaluation PCB Top Layer

Table 6. Bill of Materials for the EV1HMC787ALC3B Evaluation PCB

Quantity	Reference Designator	Part Number	Description
1		117611-1	PCB, evaluation board
2	J1 to J2	104935	2.92 mm connectors, SRI
1	J3	105192	SMA connector, Johnson
1	U1	HMC787ALCB	Device under test (DUT)